

Fermilab Main Injector Beam Position Monitor Upgrade

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An upgrade of the Beam Position Monitor (BPM) signal processing and data acquisition system for the Fermilab Main Injector is described. The Main Injector is a fast cycling synchrotron that accelerates protons or antiprotons from 8 to 120 (or 150) GeV/c. Each Main Injector cycle can have a totally different magnet ramp, RF frequency configuration, beam bunch structure, and injection/extraction pattern from the previous cycle. The new BPM system provides the capabilities and flexibility required by the dynamic and complex machine operations. The system offers measurement capability in 2.5 MHz and 53 MHz channels to detect the range of bunch structures for protons and antiprotons in both wideband (turn-by-turn) and narrowband (closed-orbit) modes. The new BPM read-out system is based on the digital receiver concept and is highly configurable, allowing the signal processing of nearly all Main Injector beam conditions, including the detection of individual batches. An overview of the BPM system in the Main Injector operating environment, some technology details and first beam measurements are presented.

INTRODUCTION

A project to upgrade the readout electronics of the Fermilab Main Injector (MI) Beam Position Monitors (BPMs) was begun in 2005. This project is one of many improvements to the Fermilab accelerator complex aimed at providing increased luminosity and reliability for the Run 2 collider physics program. The design builds upon many of the technologies and techniques used to upgrade the BPM electronics in the Fermilab Recycler Ring¹, the Fermilab Tevatron², as well as some of the transfer lines.

OPERATIONAL REQUIREMENTS

The Fermilab Main Injector (MI) is designed to accelerate either protons or antiprotons with an injection energy of 8 GeV and an extraction energy as high as 150 GeV. At any time, however, only one of the two beam species may be present. The time structure of the beam may be one of the following:

- **53 MHz Protons.** From 1 bunch up to a full batch of 84 bunches in successive 53 MHz buckets (19 ns apart). Up to 6 batches, each one of 84 bunches, can be loaded in the MI.
- **53 MHz Anti-protons.** Four consecutive groups of antiprotons, spaced by 396 ns, of typically 5 (up to 9) 53 MHz bunches, are present in MI.
- **2.5 MHz - Protons or anti-protons.** Four 2.5 MHz bunches in successive 2.5 MHz RF buckets (396 ns spacing).

It is required that the system be able to measure either the 2.5 or the 53 MHz components of the beam and, in both cases, be capable of either narrowband or wideband measurements³. In the system described here, the narrowband measurements are the same for both the 2.5 and 53 MHz channels; they have a resolution bandwidth of about 1 kHz and represent an ensemble average over all bunches in the machine over approximately 40 turns. Wideband 2.5 MHz measurements have a resolution bandwidth of about 250 kHz and represent an average over all bunches within a window of about 1/28 of the circumference of the MI. Wideband 53 MHz measurements have a resolution bandwidth of about 500 kHz and represents an average over all bunches within a window of about 1/14 of the circumference of the MI.

The narrowband mode is also referred to as Closed Orbit mode. When wideband measurements are made they are triggered to sample the same section of the stored beam on consecutive turns; so it is also referred to as Turn-by-turn mode.

A cartoon description of the MI function is that it must accept beam from one of several injection lines, accelerate the beam, perhaps change its RF structure, and deliver it to one or more of the extraction lines. Each distinct combination of these steps is referred to as an MI cycle. Some cycles have multiple injections, all from the same injection line. Including test and development cycles, about 20 cycles are currently defined but one can imagine that new cycles will be defined for future operations. In typical operations the MI repeats a supercycle that contains many cycles, possibly all the same cycle, but possibly different cycles mixed together. A typical cycle length is 1 to 2 seconds but a few take up to 10 seconds.

To operate the MI effectively, the BPM system needs to be highly programmable with a different program for each MI cycle. The following discusses the measurements that may be a part of the program for a particular cycle; some, but not all, of these options are mutually exclusive. Perform continuous narrowband measurements at a rate of at least 500 Hz; the system must be able to report these measurements in real time at 500 Hz. For each injection in the cycle, perform a

wideband measurement of the newly injected beam for its first 512 turns; report the position on the first turn and the average position over the first 16 turns. For one of the extractions in a cycle, perform a wideband measurement on the portion of the beam to be extracted, capturing the last turns (up to 512) before extraction; report the position on the last turn. At a specified time within the cycle, perform a wideband measurement on a specified part of the beam recording the position for 2048 turns. For all of the wideband measurements, the system makes the data available at the conclusion of the cycle and retains the data until it is overwritten by another instance of the same cycle. The BPM system must be able to report synchronized wideband measurements from around the ring, with each BPM measuring the same portion of the beam.

Finally, the MI control system periodically commands the BPM to retain the most recent narrowband measurement and to hold this measurement in a buffer for future inspection. Measurements from each cycle are kept in local memory by the front-end software system⁴.

The quantities that serve as the basis for the BPM measurement specifications are defined as follows:

- a) Measurement range. The range of positions, relative to the BPM center, over which the BPM measurement must be valid and meet the accuracy requirements.
- b) Absolute position accuracy. This determines how well the position of the beam is measured with respect to the survey center-line.
- c) Relative position accuracy. This determines how well the displacement of the beam is measured over the measurement range. This requirement does not include offset errors, but does set limits on the scale errors, non-linearities and random errors.
- d) Position resolution. The smallest change in beam position that the BPM can reliably measure.
- e) Position linearity. The linearity of the BPM response to orbit changes over the measurement range. The linearity is defined as the difference between the measured BPM position and the slope of the BPM response at the center of the BPM.
- f) Long-term position stability. The BPM systems ability to give the same position value for the same beam position and intensity over time.
- g) Relative intensity accuracy. The relative intensity measurements between BPMs.

Table 1 summarizes measurement specifications for 53 MHz and 2.5 MHz bunch structures, respectively.

Measurement specifications	2.5 MHz bunch structure (anti-protons)	53 MHz bunch structure
Measurement range	± 25 mm	± 25 mm
Absolute position accuracy	1 mm + 10% of actual beam position	1 mm + 5% of actual beam position
Relative position accuracy (3σ)	10% of actual beam position	5% of actual beam position
Position resolution (3σ)	0.3 mm (0.5 mm)	within 1.5% of the slope of the BPM response at the center of the BPM over ± 15 mm
Long-term position stability	0.5 mm	0.2 mm
Relative intensity accuracy	20% (30%)	10% (20%)

Table 1: Measurement specifications for 2.5 MHz and 53 MHz bunch structure, narrow bandwidth mode. Values in parenthesis are for high bandwidth mode, with single turn resolution.

THE MI-BPM READ-OUT ELECTRONICS HARDWARE

Figure 1 gives an overview of the hardware components of the upgraded Main Injector BPM system. The figure shows 2 signal channels which produce one position channel. All 201 stripline BPM pickups and their tunnel cabling remains unchanged⁵. Upgraded components are the Combiner Boxes, located in the accelerator tunnel and all read-out electronics hardware, concentrated in 7 locations around the ring. The analog electronics – several 8-channel so-called Transition Modules and a Controller Module - are located in a 6U high crate frame, supplied by an external, linear power supply. A core component of the BPM system is the commercial Echotek GC814 digital signal receiver. Several 8-channel modules, together with a processor module (Motorola MVME5500) and an in-house developed Timing & Fan-out Generator (TFG-II) are arranged in a VME crate. All the cabling within a rack and between crates is new.

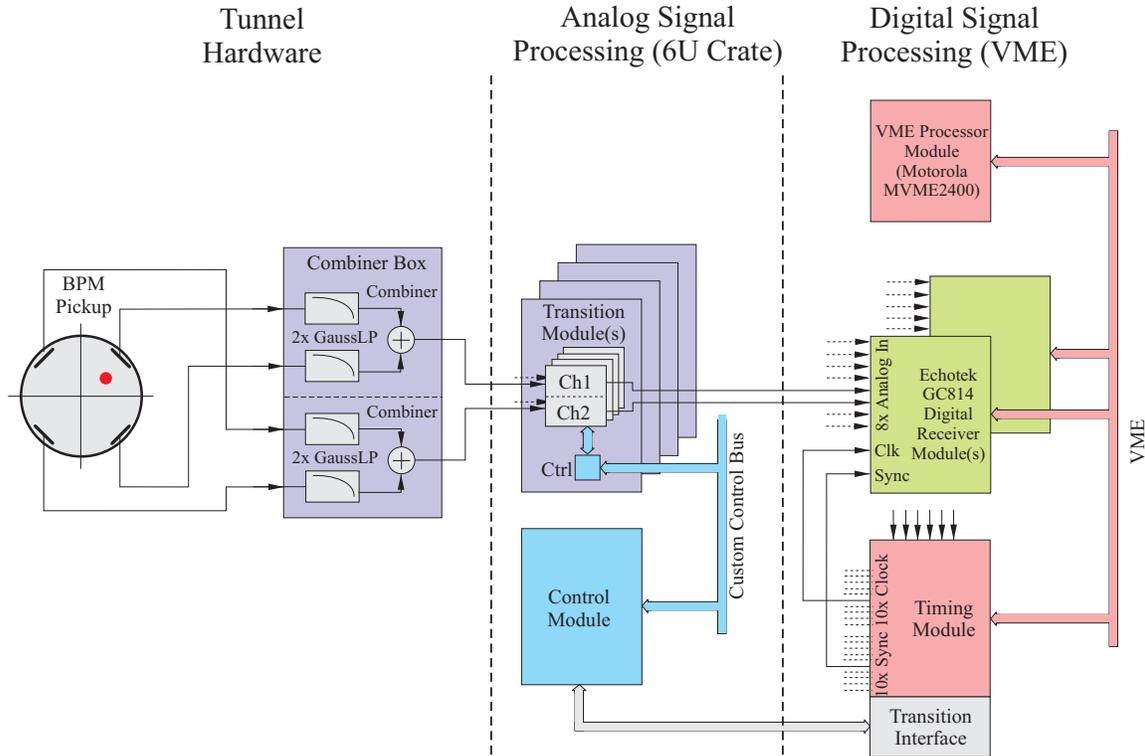


Figure 1. Overview of the Main Injector BPM hardware, shown 2 of 430 channels total.

Timing and Trigger Distribution

The Timing Generator Fanout (TGF-II) is a double width, 6U VME card that is an evolution of the Tevatron TGF card. The TGF-II receives cycle timing and control signals from the accelerator control system and beam synchronous information from the MI, the Recycler and the Booster accelerators. These are translated into the timing and control signals required for the operation of the BPM system, including those signals required by the VME processor, the Transition Modules, and the digital receivers. TGF-II is also the source for diagnostic signals that can be injected onto the RF signal path at the transition boards.

A fundamental timing signal utilized by the TGF-II is the MI RF signal, the master beam synchronous timing reference for the MI. The TGF-II card uses this reference to derive the digitization clock, $10/7$ of the RF frequency, which it distributes to each digital receiver board. The digitization clock tracks all changes in the MI RF frequency, ensuring that the pass-band of each DDC channel remains centered on the signal frequency.

Analog Signal Pre-Processing System

Analog signal processing occurs in both the Combiner Boxes and the Transition Modules. The Combiner Boxes accept beam signals from the four stripline electrodes, pass them through a low-pass filter ($f_c \approx 100$ MHz) and combine them to form either a vertical measuring or a horizontal measuring signal pair. The analog Transition

Modules interface the signals from the Combiner Boxes to the Echotek digital receiver, adapting signal levels and further filtering frequency content.

The broadband signals from the BPM pickups require analog frequency-domain filtering and amplitude management before presentation to the digital receivers. Therefore each Transition Module channel is actually comprised of two parallel signal paths, one for the 2.5 MHz signal band and another for the 53 MHz band, each with appropriate bandwidth shaping filters and programmable gain. The bandwidth of each is about 5 MHz, chosen to allow resolution of individual batches in the Main Injector. All signal paths have remote controlled features, including enable/disable, gain control (range: 0...48 dB), and low/high post-amplification switch.

Test signals of 2.5 or 53 MHz, supplied by the timing generator, can be applied to the Transition Module inputs to simulate centered or displaced beam. This test signal capability, as well as all remote operations is controlled through local programmable logic. The central control of an analog crate is managed by a control module that is connected to the VME Timing & Fan-out Generator module through a rear transition board (Fig. 1). Further details on the analog pre-processing electronics are presented elsewhere in these proceedings⁶.

Digital Signal Processing System

Eight-channel EchoTek ECDR-GC814-FV2 VME64X digital receiver boards are used for beam signal digitization and frequency conversion. Each GC814 channel is comprised of a 14 bit ADC (AD6645), a four-channel digital down DDC (GC-4016), an FPGA accumulator, and 128K x 32 dual-ported RAM. These boards shift the information content of the beam signal from a RF analog form to a stream of digital, baseband I and Q data pairs. The square root of the sum-of-the squares for each I,Q pair is computed and referred to as A or B. These form the basis for the beam position given by $(A-B)/(A+B)$, and multiplied by a polynomial in with coefficients determined by bench test measurements of the pickups.

The 53 and 2.5 MHz beam frequencies are harmonically related in the ratio of 21:1. By design choice, the ADC clock (fADC) is locked to 10/7 times the 53 MHz frequency. This is approximately 76 MHz with a 400 kHz sweep during acceleration. Given this digitization frequency, the 2.5 MHz beam signal band is oversampled. The 53 MHz signal band, however, is undersampled and translates down to $3/10$ fADC, 22.759 MHz. The analog 53 MHz transition board bandpass filters suppress un-wanted images. Digitized signals are down-converted to baseband in the GC-4016 with numerically controlled local oscillators operating at 22.759149 MHz for the 53 MHz channels and 2.528794 MHz for the 2.5 MHz channels.

Figure 3 is a block diagram of one of four channels in each GC-4016 digital downconverter. All four channels of each GC-4016 DDC are utilized. Channels 1 and 2 are configured for 53 and 2.5 MHz narrowband operation respectively and channels 3 and 4 for wide-band operation.

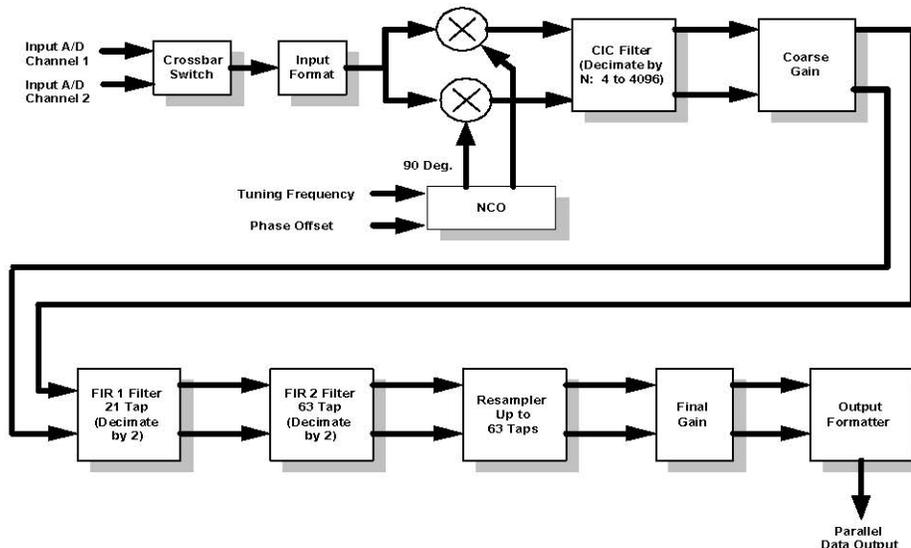


FIGURE 3. Block diagram of one of four channels in each GC-4016 digital downconverter IC.

Narrowband filtering is achieved with decimation of 1536 in the 5-stage CIC filter and decimation of two in each of the CFIR (FIR 1) and PFIR (FIR 2) filters. For $f_{ADC} = 75.86$ MHz, the CIC output update and CFIR tap interval is 20.2 microseconds. The PFIR tap interval is ~ 40.5 μ s and its output updates at ~ 81.0 μ s. Each narrowband CFIR is set up as an 11-tap Gaussian filter and each PFIR as an 11-tap boxcar filter. The eleven taps in the PFIR correspond to a signal integration window of 445 μ s (11×40.5 μ s), about 40 MI turns. The 3db bandwidth of the narrowband channels is 1 KHz.

Wideband channels, both 2.5 and 53 MHz, are programmed to have a total decimation of 32, distributed as 8 in the CIC filter and 2 each in the CFIR and PFIR filters. With a 75.86 MHz digitizer frequency, the sampling interval is ~ 106 nanoseconds in the CFIR, ~ 211 ns in the PFIR and ~ 422 ns out of the PFIR. The wide-band CFIR and PFIR are both configured as boxcar filters.

The 53 MHz wideband channel CFIR has just two non-zero coefficients and the PFIR has four, producing a signal integration time of 844 ns (4×211 ns), about $\frac{1}{2}$ of one MI beam batch. This window is a compromise to include signal from half of one batch while providing good isolation from signals due to adjacent batches. The resulting effective bandwidth of the 53 MHz wide-band channel is about 500 KHz.

The 2.5 MHz wideband channel CFIRs also have two non-zero coefficients; the PFIR has eight taps providing a signal integration window of 1687 ns (8×211 ns). This window is sufficient to include signal from all four sequential 2.5 MHz bunches, typical of the beam for which this channel is used. The effective bandwidth of the 2.5 MHz wide-band channel is 250 KHz.

PERFORMANCE

Approximately two weeks before the spring 2006 Fermilab accelerator shutdown a pre-production system was installed to read out 11 BPMs. The results presented in this section were obtained using this pre-production system.

Fig. 3 illustrates the narrow band capability for an MI cycle in which six batches are injected from the Booster, accelerated to 120 GeV, moved onto the extraction orbit and delivered to the neutrino production target for the NuMI beamline. The six injections occur at intervals of 1/15 second and the beam is extracted in a single turn.

In the following the symbols A and B denote the voltages reported by the Echotek for the two BPM channels. Part a) of Fig. 3 shows the electrode sum, $A+B$, as a function of time. For constant machine energy and bunch structure the electrode sum is proportional to the beam intensity and the six injections can be seen. The modulation in $A+B$ after the sixth injection is due to changes in the bunch length as the beam energy is increased. Part b) of Fig. 3. shows the measured position for this same dataset; the position is given by a polynomial in $(A-B)/(A+B)$, with coefficients determined by bench test measurements of the pickups.

For about the last 12 ticks (48 ms) before extraction the beam position is reasonably constant on the extraction orbit. The RMS spread of the measured position for these 12 ticks is typically 2.5 microns for the six instrumented vertical measuring BPMs. The predicted resolution for the system is less than 1 micron so it is believed that the RMS is dominated by true beam motion. At the five instrumented horizontal locations, there is obvious beam motion that is folded into the measured RMS values, which range from about 5 to 10 microns. In all cases these RMS values, even uncorrected for true beam motion, are well below the required resolution of 17 microns.

The pre-production system has also allowed us to measure the stability of the MI. Narrowband measurements were made for 281 cycles over a period of a half hour. At each BPM the mean orbit and the RMS spread about the mean orbit were computed. At most BPMs, at most times in the cycle, the RMS spread of the measured position is less than 10 microns. Even at the least stable locations the spread was less than 50 microns for most times in the cycle. This represents the most precise determination to date of the stability of the MI.

The data in Fig. 3 can also be used to illustrate the flexibility of the upgrade BPM system. The system can be programmed to make turn-by-turn measurements of the newly injected batch at each of the 7 injections, to make closed orbit measurements during the middle of the cycle, and to make turn-by-turn measurements of any one batch at its extraction. This has been demonstrated but the data is not shown here.

Fig. 4 shows a wideband 2.5 MHz measurement for a cycle that takes anti-protons from the Recycler and injects them into the Tevatron. Parts a) and b) of the figure show, respectively, the electrode sum and the position. The measurement was triggered about 28 turns before the beam was injected so the first 28 points illustrate the low noise level of the system. Part c) shows the Fourier transform of the position data, in which a strong betatron line can be seen.

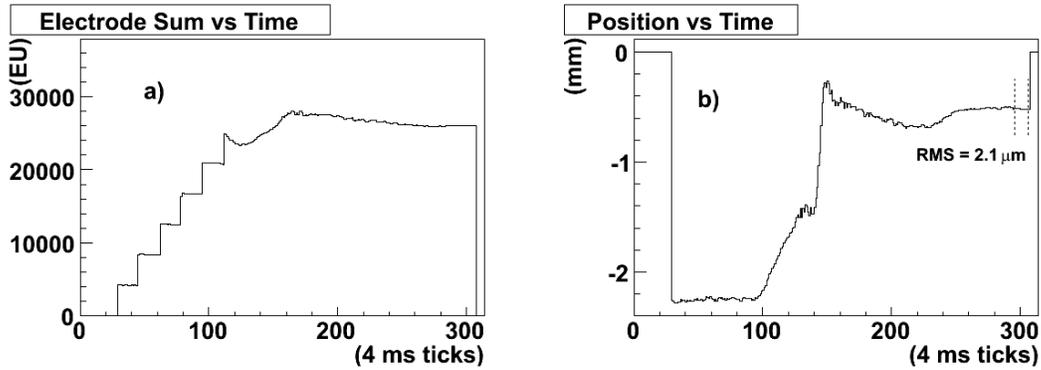


Figure 3. Part a) shows the electrode sum A+B, for a vertical measuring BPM for an MI cycle described in the text. For both parts of this figure, the horizontal axis is time, measured in 4 ms ticks; the system can deliver data with 2 ms ticks but this dataset was taken at the slower rate. The noise level in the electrode sum is not visible on this scale. Part b) shows the measured vertical position for the same dataset. The vertical lines near the right edge of the plot mark 12 measurements used to estimate the resolution of the system. The RMS spread of these 12 points is 2.1 microns.

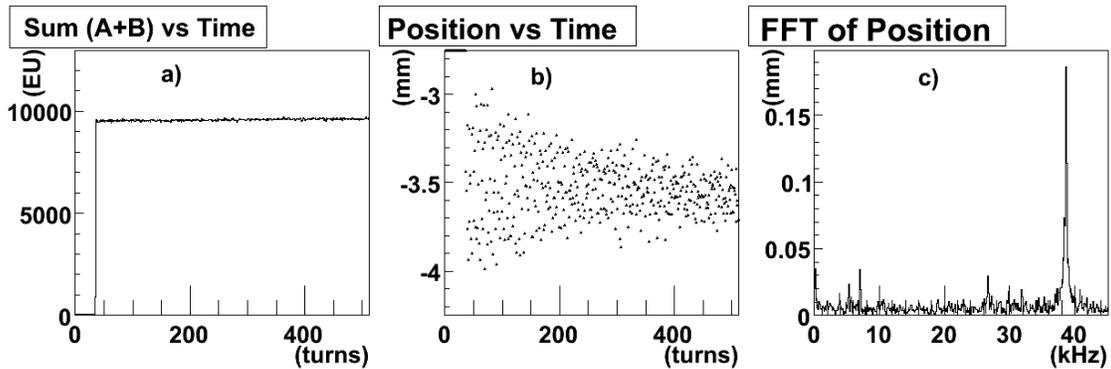


Figure 4. Part a) shows the electrode sum for a turn-by-turn measurement of anti-protons injected from the Recycler. The noise level is barely visible on this scale. Part b) shows the measured position. For parts a) and b) the horizontal axis is turn number. The FFT of the position data is shown in part c); the mean position was subtracted before computing the FFT. The betatron tune line is strong.

CONCLUSION

An upgrade to the Fermilab Main Injector Beam Position Monitor electronics and readout has been designed and prototyped. Initial measurements with the prototype system shows good performance that meets the requirements for machine operations. Final hardware and software is being prepared for final installation and commissioning in the summer of 2006.

ACKNOWLEDGMENTS

This work was supported by Fermi National Accelerator Laboratory, operated by Universities Research Association Inc. under Contract No. DE-AC02-76CH03000 with the United States Department of Energy.

REFERENCES

- ¹Webber, R. *et al.* "FermiLab Recycler Ring BPM Upgrade Based on Digital Receiver Technology" Beam Instrumentation Workshop 2004, Knoxville, Tennessee 3-6 May 2004.
- ²S. Wolbers, *et al.*, "Tevatron Beam Position Monitor Upgrade", PAC2005, Knoxville, TN, 16-20 May 2005.
- ³D. Capista, *et al.*, "Requirements for the Main Injector BPM Upgrade", <https://beamdocs.fnal.gov/cgi-bin/public/DocDB/ShowDocument?docid=1786>
- ⁴L. Piccoli, *et al.*, "The Main Injector Beam Position Monitor Front-End Software", submitted to Beams Instrumentation Workshop, Fermilab, Batavia, IL, 1-4 May 2006.
- ⁵Fitzgerald, J., Crisp, J. "A Compact BPM for the Fermilab Main Injector" IEEE PAC, 97CH36167, May 1997.
- ⁶M. Wendt, *et al.*, "Analog Signal Pre-Processing for the Fermilab Main Injector BPM Upgrade, submitted to Beams Instrumentation Workshop, Fermilab, Batavia, IL, 1-4 May 2006.