

Fermi National Accelerator Laboratory
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High-power RF sources and components for linear colliders

S. Kazakov

KEK, Tsukuba, Japan
Omega-P, Inc., New Haven, CT

Outline:

❑ Warm technology

- History of development of RF sources for LC with examples from work at BINP
- Windows for high power X-band sources
- X-band pulse compression systems

❑ Cold technology

- TOSHIBA 10 MW L- band klystron
- Coupler with capacitive cold window
- Phase-shifters

History of linear colliders probably started in 1978 from BINP. See preprint: V.E. Balakin, G.I. Budker, A.N. Skrinsky “Feasibility of creating superhigh energy colliding electron positron beam facility”.

At that time a new laboratory was organized at BINP especially to develop technologies necessary for future linear super-collider (V.E. Balakin – head of laboratory)

Many pioneering results were achieved, including

- accelerating gradient ~ 100 MV/m was experimentally demonstrated
- new type of damping of beam instability was invented (BNS damping)
- flat beams

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- And several new types of type RF sources were experimentally investigated:

Separatron, multi-cavity gyrocon, “linear gyrokon”, PPM klystron.

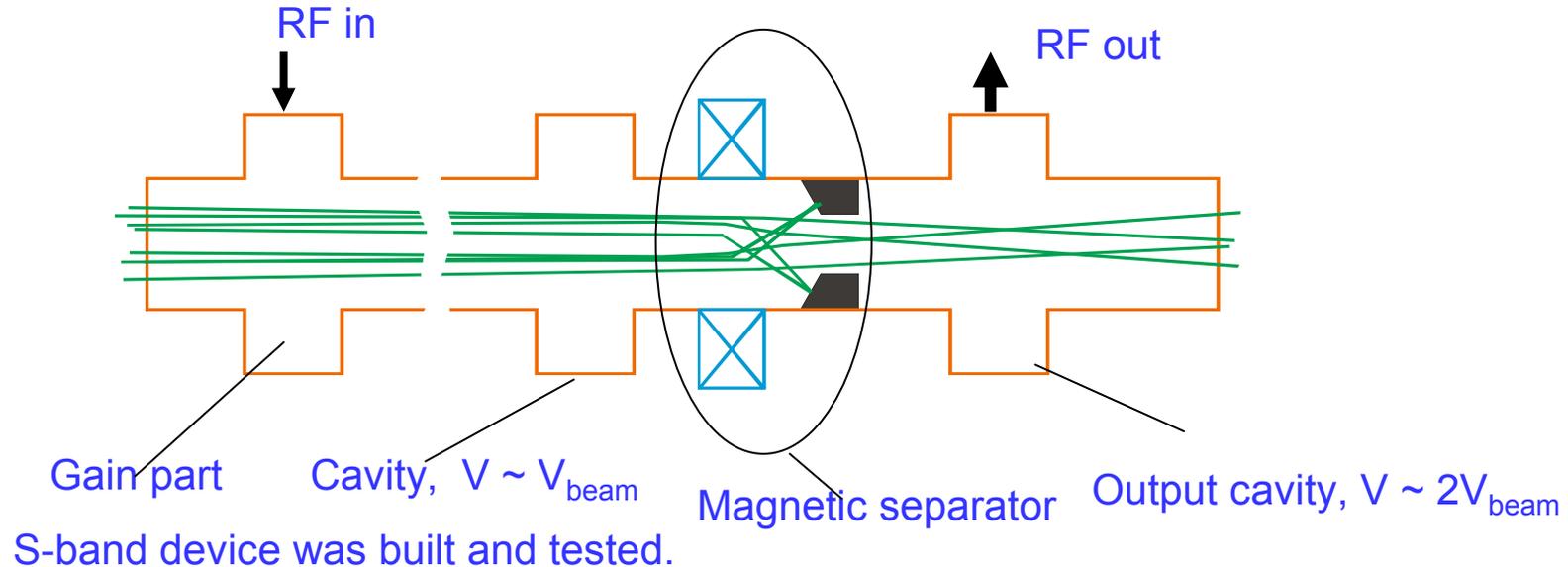
Klystrons with annular beam with conventional and magnetic bunching were studied theoretically.

It became clear from the beginning that a linear collider would require RF sources with enormous output power, which did not exist and must be created.

Some exotic devices

First project was S-band normal conducting collider. RF source for this type of collider should emit of 100's of MW of RF power to keep the number of sources within an acceptable limit. But high power means high voltage for the electron beam, maybe relativistic. But a relativistic beam it is not easy to bunch by changing its velocity - the device would become very long. A couple of clever schemes were invented.

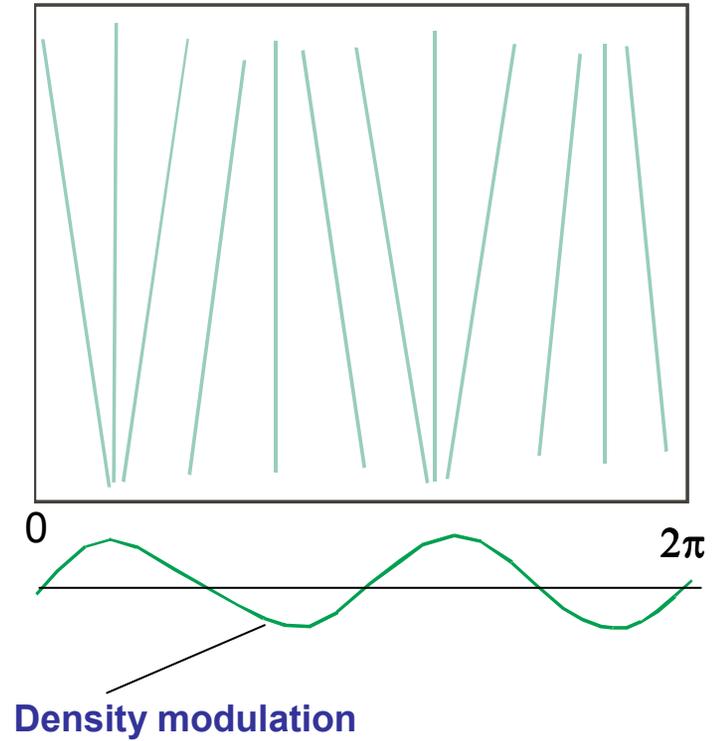
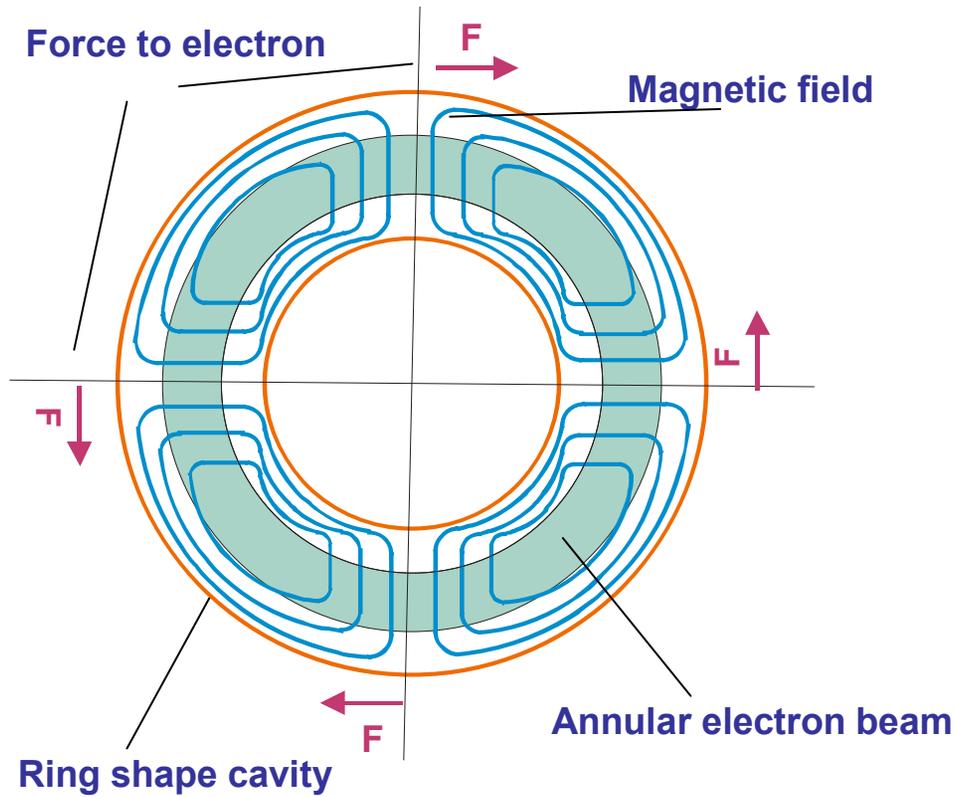
Separatron* (V.Balakin, N.Solyak, A Mikhalichenko):



Problems: corrosion of separating iris, double voltage of output cavity

* N.A. Solyak, "Studying of Separatron - a new RF power source", Diploma thesis, Novosibirsk 1976

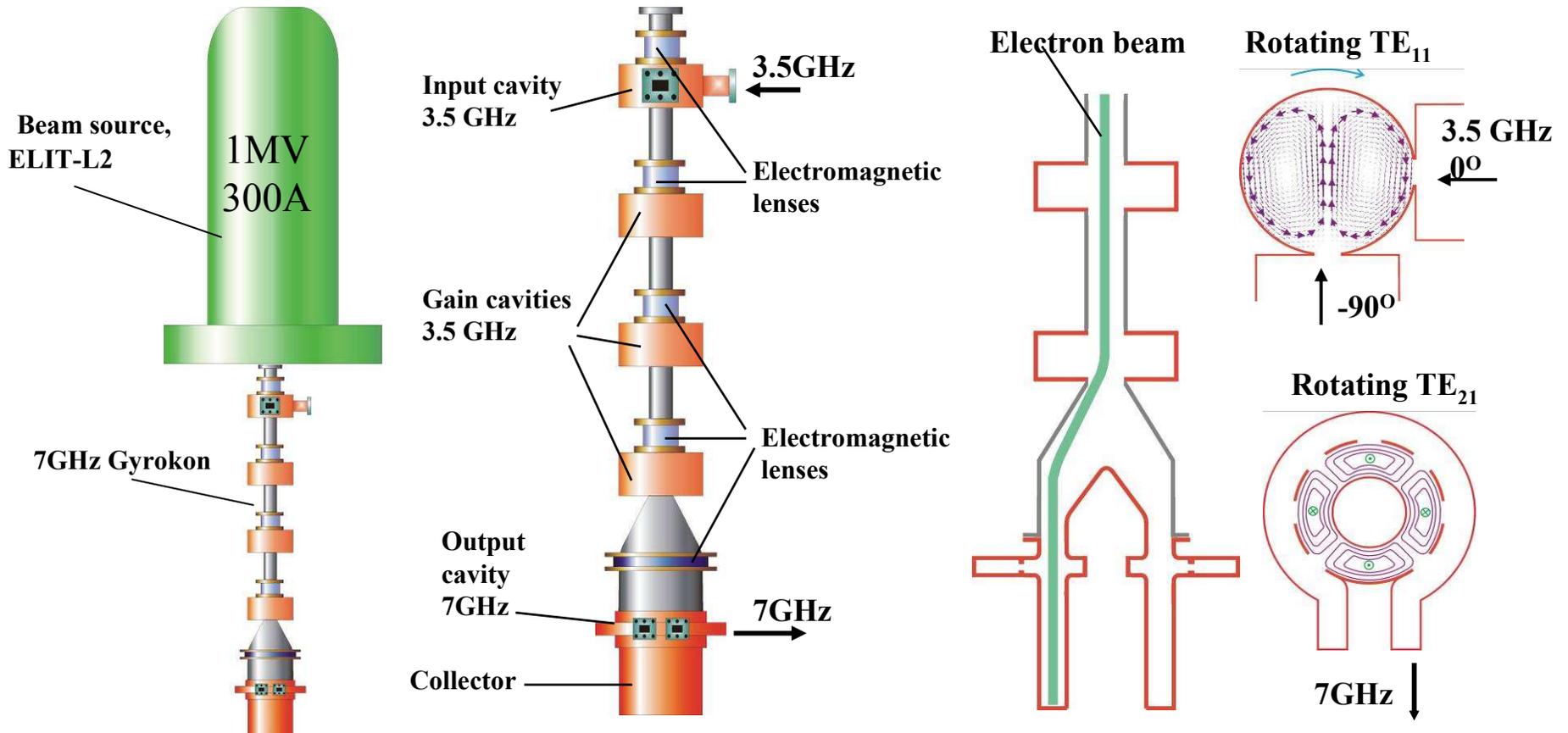
Klystron with modulation by magnetic field (V.Balakin, N. Solyak)



Wave in cavity can be traveling

Frequency of VLEPP project was changed, 3 GHz -> 7GHz

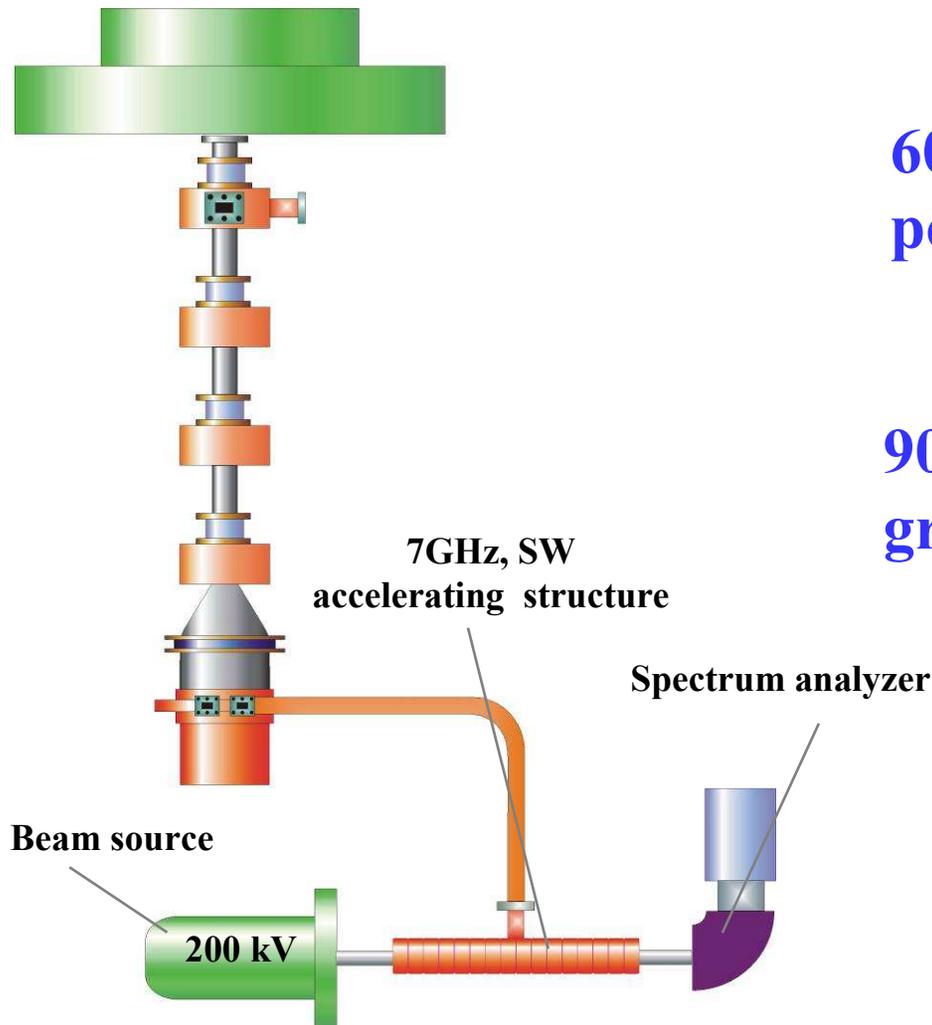
Candidate for collider RF source became a multi-cavity gyrocon*



Gyrocon worked as frequency multiplier.

*Balakin V.E., Kazakov S.Yu, Solyak N.A, "RF deflection system of Gyrocon", Workshop on Relativistic RF electronics, Gorkii, 1982.
Balakin V.E., Kazakov S.Yu., Solyak N.A., "Gyrocon", Proc. of the Intern. Workshop on Next-generation Linear Colliders SLAC, 1988.

Gyrocon was used as RF source for high accelerating gradient experiment.



60 MW, 1 μ s output power was achieved.

90 MV/m accelerating gradient was reached

But cost optimum of linear collider lies in higher frequency region. VLEPP frequency was doubled, 7GHz ->14GHz. Laboratory started design of a **14 GHz klystron**

VLEPP klystron was design on basis of several very bold ideas:

- DC voltage 1 MV (!)
- Gridded gun with multi-beam big cathode
- Permanent magnetic system
- TW output structure

Great challenge !

Table 1: VLEPP klystron, designed and achieved parameters.

Parameter	Design	Experiment
Operating frequency	14 <i>GHz</i>	14 <i>GHz</i>
Peak output power	150 <i>MW</i>	60 <i>MW</i>
Beam energy	1 <i>MeV</i>	1 <i>MeV</i>
Beam current	300 <i>A</i>	150 <i>A</i>
Grid voltage	25 <i>kV</i>	20 <i>kV</i>
Pulse duration	0.5 <i>μsec</i>	0.5 <i>μsec</i>
Saturation gain	80 <i>dB</i>	90 <i>dB</i>
Cathode	37 microcathodes	
Emitter type	oxide	
Cathode diameter	120 <i>mm</i>	
Type of magnetic system	MPFS	

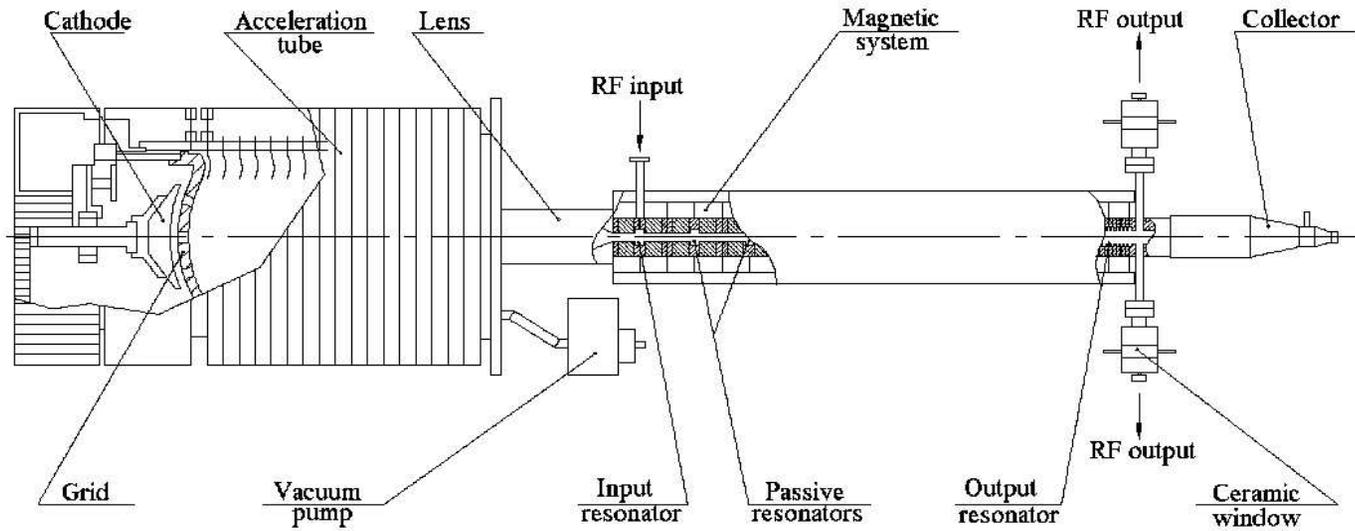
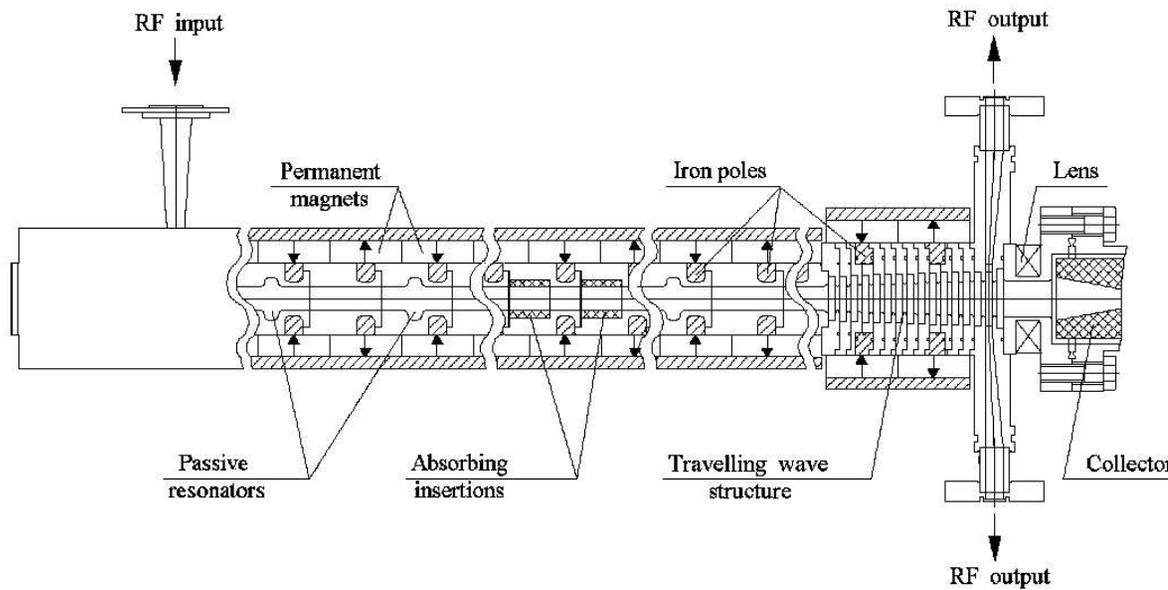
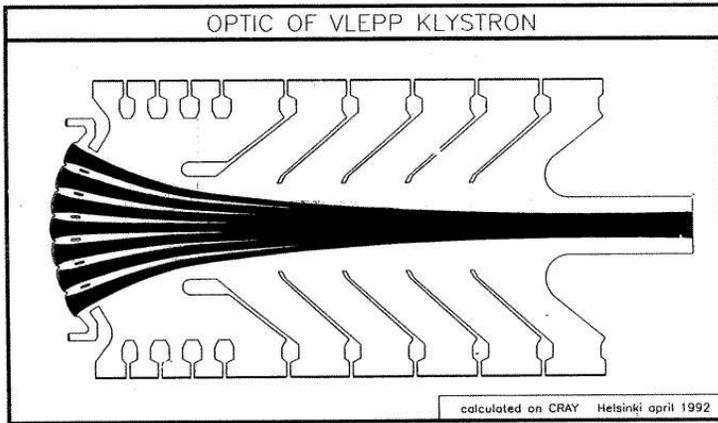


Diagram of VLEPP klystron



Cavities and magnetic systems of klystron



VLEPP gridded klystron with PPM focusing.

Achieved parameters :

Frequency = 14 GHz	RF power = 60 50 MW
Voltage = 1000 kV	Gain = 75 90 dB
Current = 150 A	Pulse length = 0.7 μ s

(Sorry for pictures quality. At that time a digital camera did not exist yet)

VLEPP klystron did not achieved design parameters, but it made great impact to modern klystron design. Following VLEPP klystron, SLAC and KEK designed their own 11.4 GHz 75 MW klystrons.

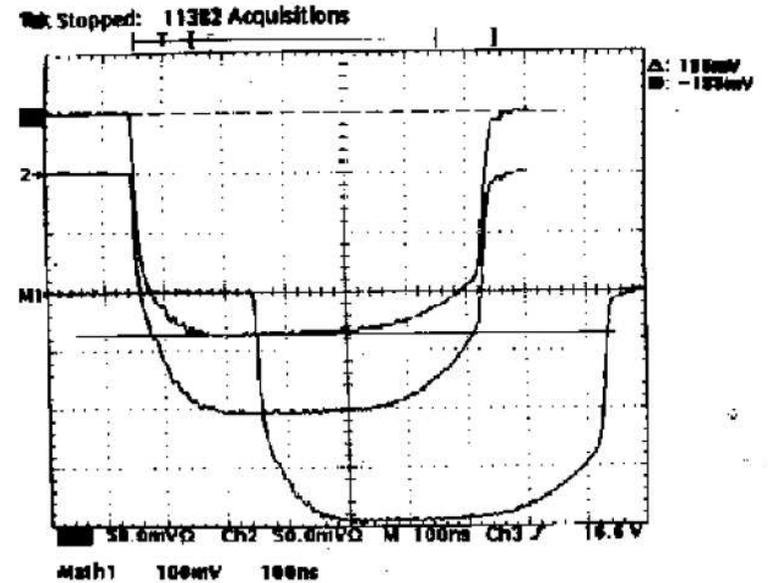
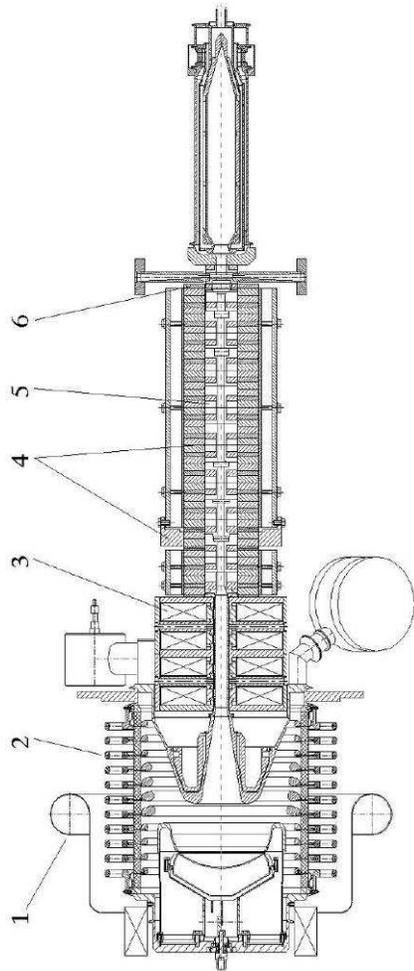


Figure 11: $P_{output} = 50 \text{ MW}$, $\tau = 600 \text{ ns}$.
1 -, 2 - signals of RF loads M1 - a total signal.

PPM klystron designed and built by BINP for KEK



SLAC 75 MW PPM klystron



TOSHIBA / KEK 75 MW PPM klystrons

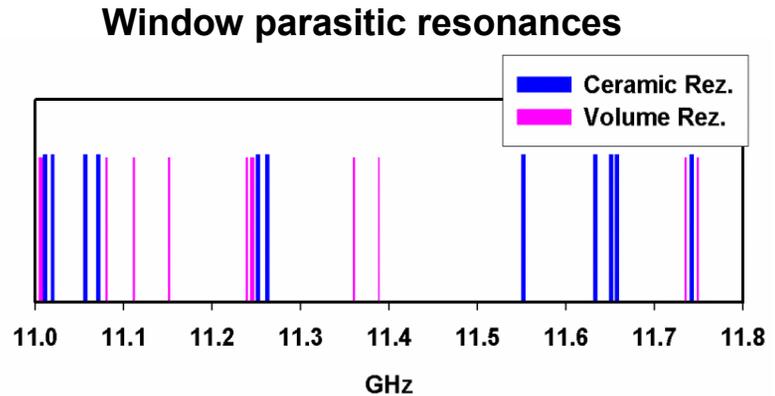
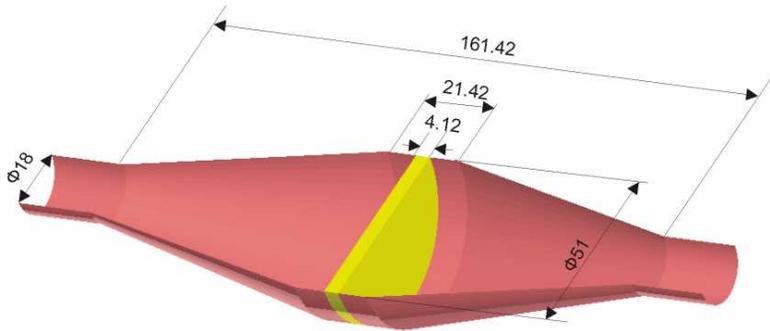
Critical elements of X-band power klystrons are the windows. The window power density was more than one order higher than in prior S-power klystrons. Old type “Pill-box” windows did not work. New ideas and approaches were needed.

Examples of attempts to make powerful X-band window in different laboratories:

X-band window	Ceramic sizes	Limiting power
$\lambda/2$, KEK	22.86 x 10.16 mm	11 MW / 70 ns
Pill-box, KEK	Ø 28.2 mm	30 MW / 200 ns
Pill-box, SLAC	Ø 27 mm	80 MW /
Long pill-box, KEK	Ø 28.2 mm	48 MW / 80 ns
Taper type, KEK	Ø 51 mm	84 MW / 700
Taper type, BINP	Ø 66 mm	50 MW / 800
Taper typer, SLAC	Ø 47 mm	20 MW /

Windows are needed which can reliably operate with power ~ 40 MW/1.6 us. If we take into account possible full reflection from a load, it means window should sustain 160 MW/1.6 us working into a matched load.

Example of BINP/KEK type tapered $\lambda/2$ window for 11.4 GHz*



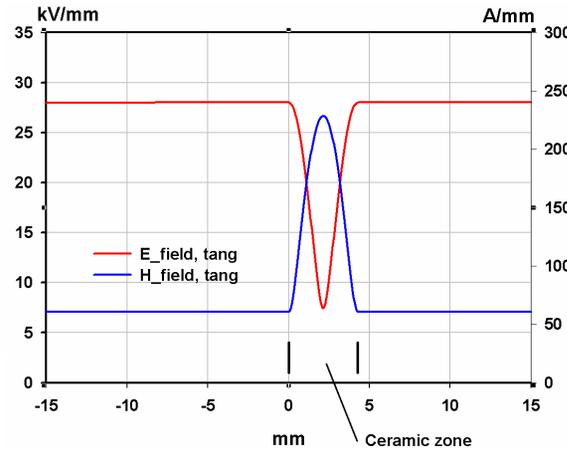
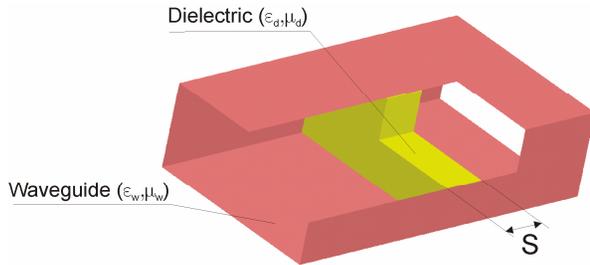
Window was broken at 84 MW / 0.7 μ s

The problem – it is difficult to increase ceramic size further because of high density of ghost modes in ceramic and trapped mode in volume. Not easy to match and field at the ceramic for a non-ideal TE₁₁ mode.

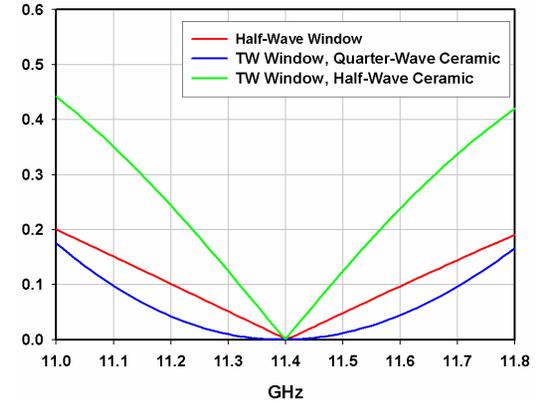
*Y.Otake, S.Tokumoto, S.Y.Kazakov, J.Odagiri, H.Mizuno, "High-Power Tests of X-Band RF Windows at KEK., KEK Preprint 96-84, July 1996.

The big step was made when the TW window was invented

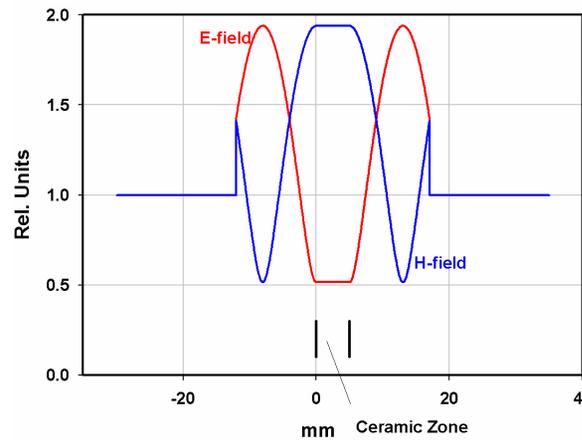
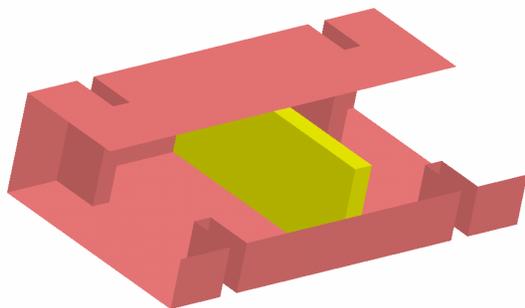
$\lambda/2$ window



TW window has wider passband

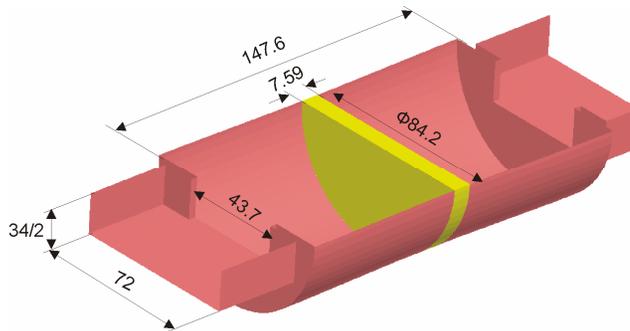


TW window

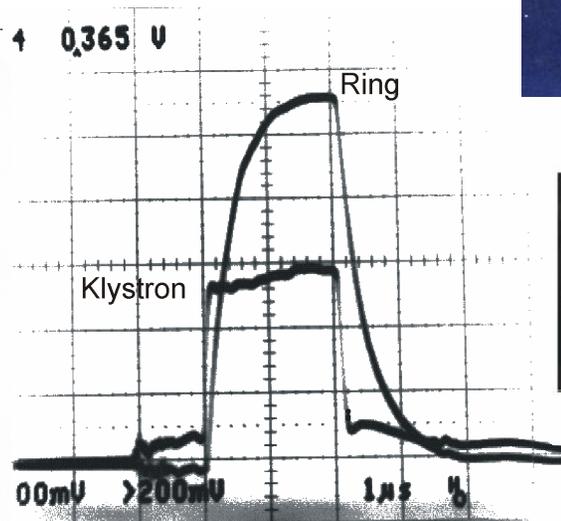
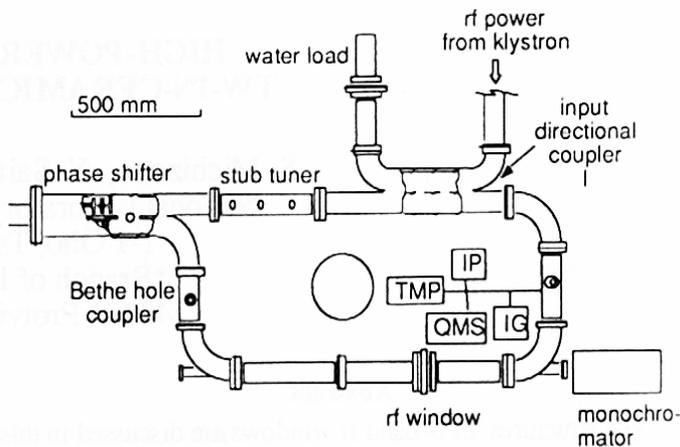


TW window can transmit $\sqrt{\epsilon}$ times more power. For Al_2O_3 it is about 3 times.

Idea was checked in S-band window*



Frequency	2856 MHz
Passband (SWR < 1.2)	100 MHz
Diameter of ceramic	84.2 mm
Thickness of ceramic	7.59 mm
Permittivity of ceramic	9.8
Max. E-tang. on the cer. surface	3.1 kV/mm (100 MW)
Max. E-norm. on the cer. surface	0.0 kV/mm (100 MW)
Max. E on the metal	18 kV/mm (100 MW)



Upstream

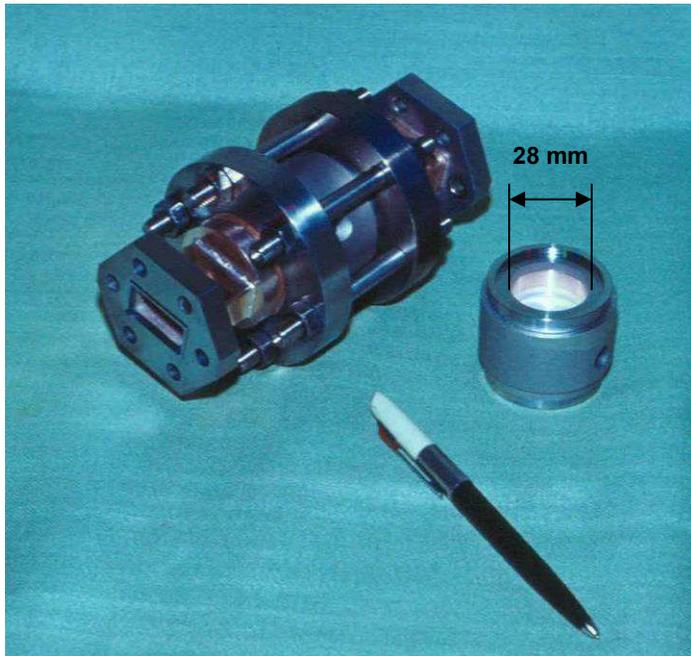


Downstream

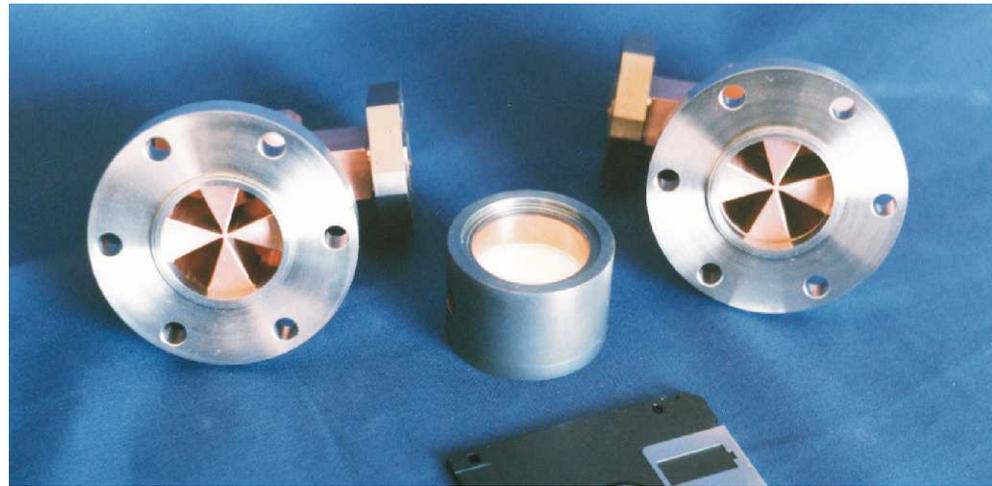
Window was successfully tested up to 470 MW / 2 us / 50pps (1 kJ in the pulse). Power was limited by resonant ring, not window. Window was not broken. We do not know the power limit for this window.

*S.Michizono, Y.Saito, H.Mizuno, S.Yu.Kazakov, "High Power Test of Pill-Box and TW-in Ceramic Type S-Band Windows", Proceedings of the 17th International Linac Conference (LINAC94), Tsukuba, Japan, August 21-26, 1994

Examples of X-band TW windows



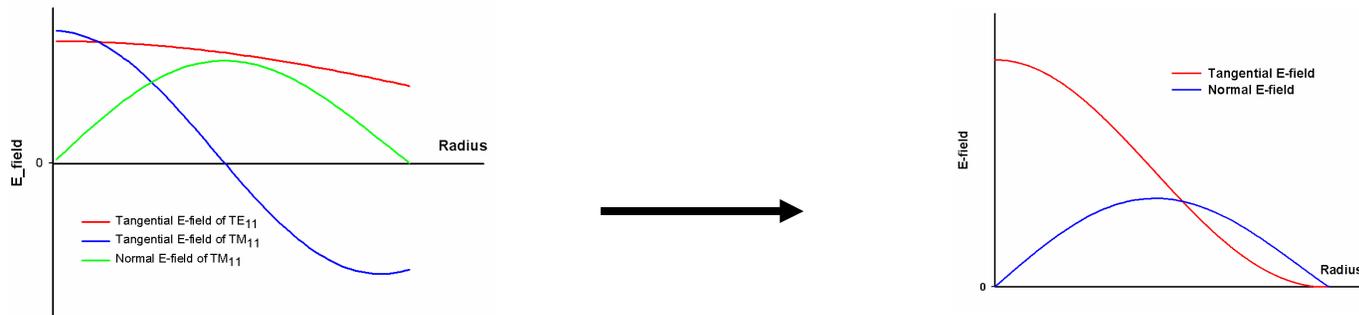
TW TE11 window. It was tested up to 80 MW / 0.2 us / 25 pps and 40 MW / 0.7 us / 25pps. Power was limited by breakdown in metal part. Ceramic survived.



TW TE01 window with mode converters, not tested

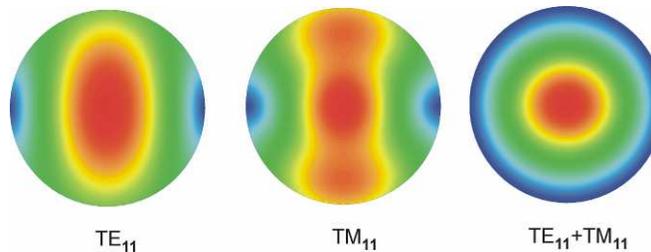
New, more interesting idea was generated – Mixed-mode TW window.

The weakest part of window is the place of brazing metal with ceramic.
The idea was to use mode mixture to decrease field at the brazing place.

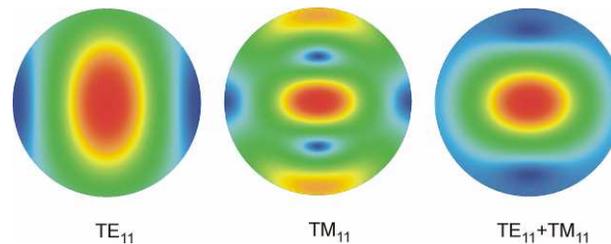


Electric fields of TE_{11} and TM_{11} modes in the circular waveguide

Electric fields of summation of TE_{11} and TM_{11} modes in the circular waveguide

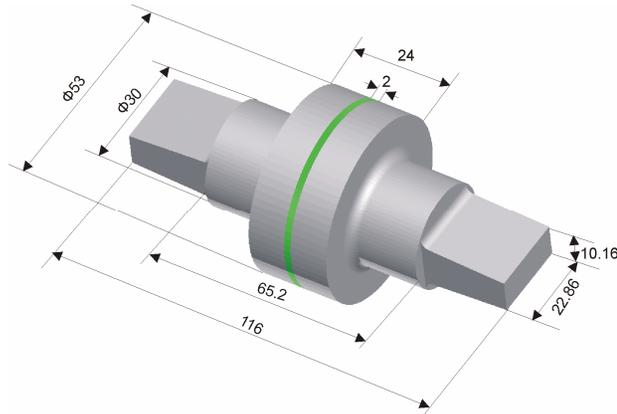


Electric fields of TE_{11} and TM_{11} modes and their summation in circular waveguide



Magnetic fields of TE_{11} and TM_{11} modes and their summation in the circular waveguide

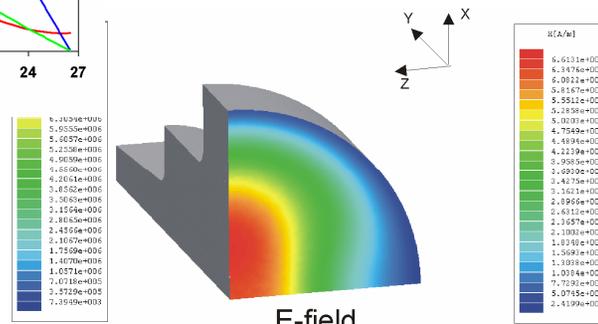
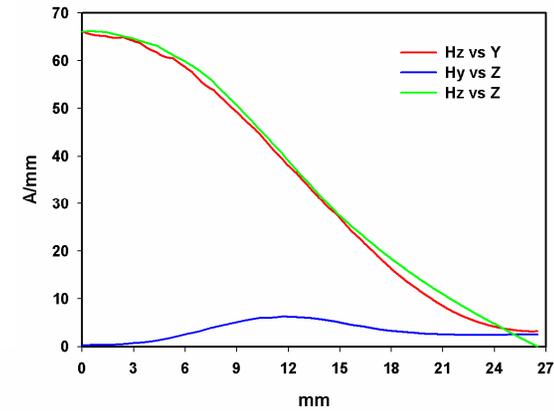
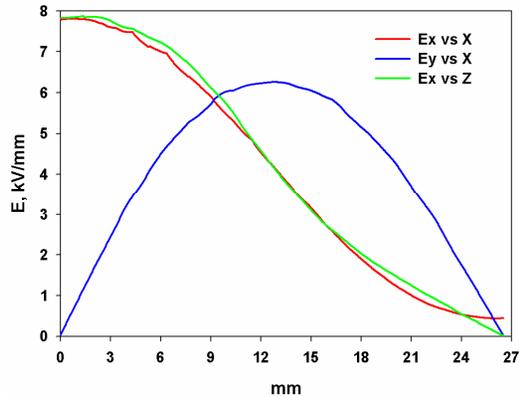
It is not a trick to mix a couple of modes, but we have to provide TW condition for both modes. To our big surprise it turned out , that the geometry of window is very simple!



Parameters of X-band TW mixed-mode window

Operating frequency	11424MHz
Ceramic diameter	53 mm
Ceramic thickness	2 mm
Ceramic permittivity	9.8
Passband (SWR < -1.2)	440 MHz
Max. E-field on the ceramic (100 MW)	7.8 kV/mm
Max. E-field on the periphery (100 MW)	0.44 kV/mm
Max. E-field on the metal (100 MW)	52 kV/mm

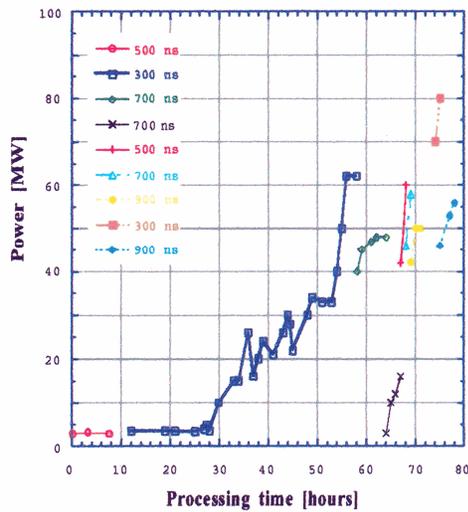
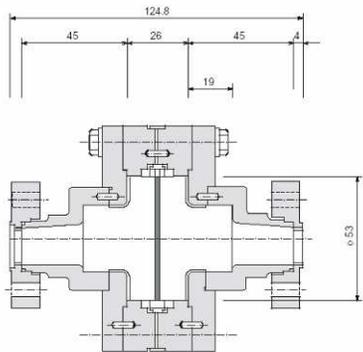
Geometry of X-band travelling wave mixed-mode window



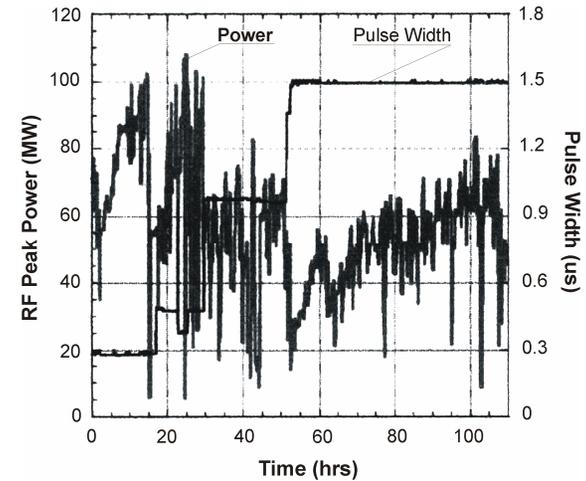
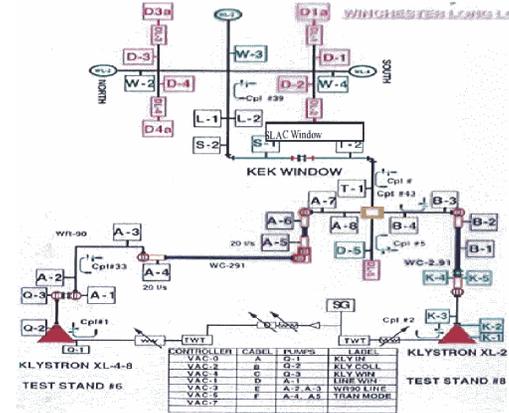
Electric and magnetic fields distributions on the ceramic of X-band TW mixed-mode window at power of 100 MW

TW Mixed-Mode window was tested in KEK and SLAC*

KEK



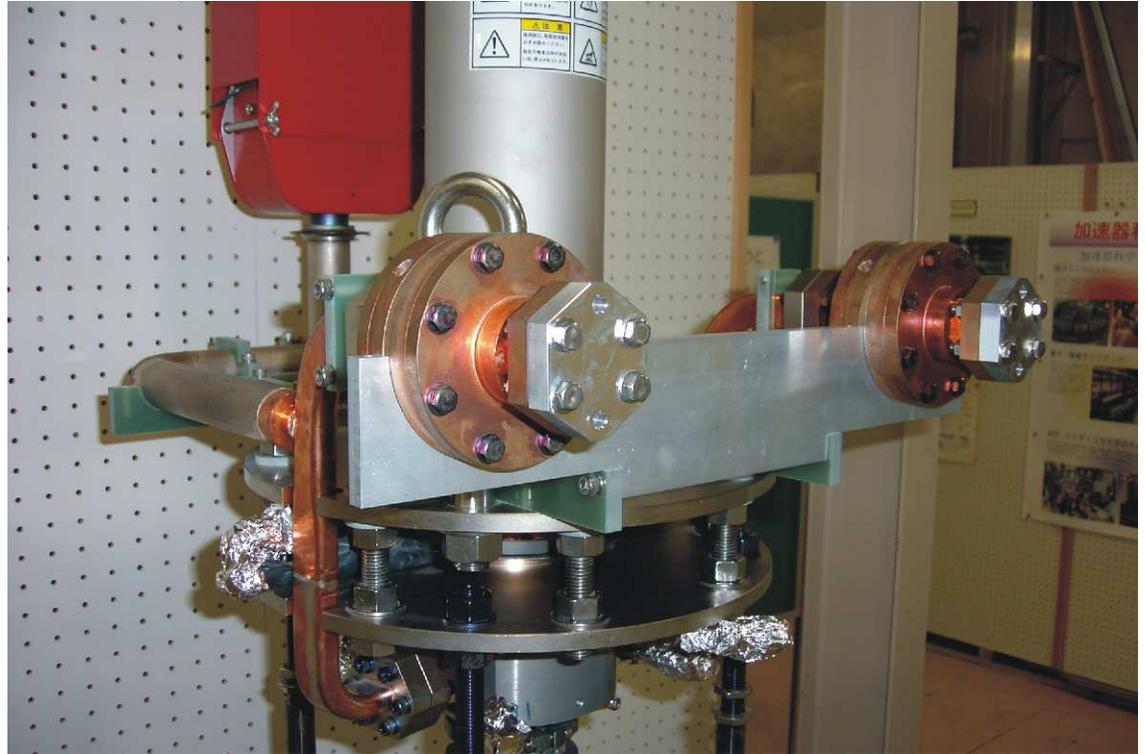
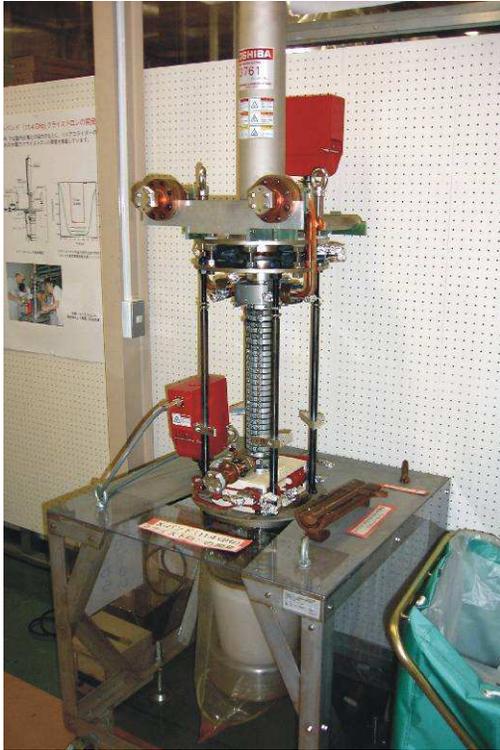
SLAC



Window was successfully tested up to 100 MW /0.5 us and 85 MW /1.5 us. Window was not broken.

*S.Yamaguchi, S.Matsumoto, S.Tokumoto, H.Mizuno, Y.H.Chin, S.Kazakov, "High-Power Test Results of Kazakov RF Window", LC99, Frascati, Italy, October, 1999

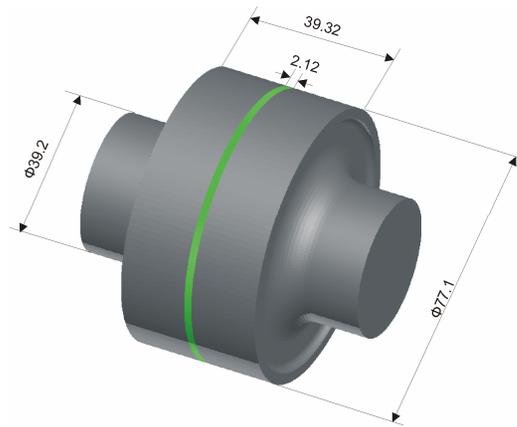
*S.Tokumoto, Y.H.Chin, H.Mizuno, K.Ohya, S.Yamaguchi, S.Kazakov, R.Loewen, R.Fowkes, A.Menegat, A.Vlieks, "High Power Tasting Results of the X-Band Mixed-mode RF Windows for Linear Colliders", LINAC 2000, Monterey, CA, USA, August, 21-25, 2000.



X-band TW mixed-mode window in the 75MW E3761 Toshiba klystron

And more powerful window!*

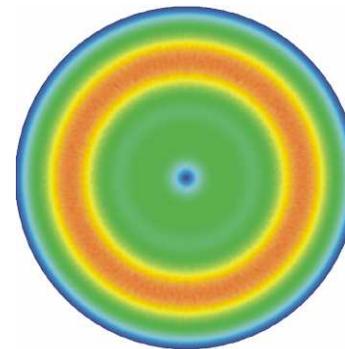
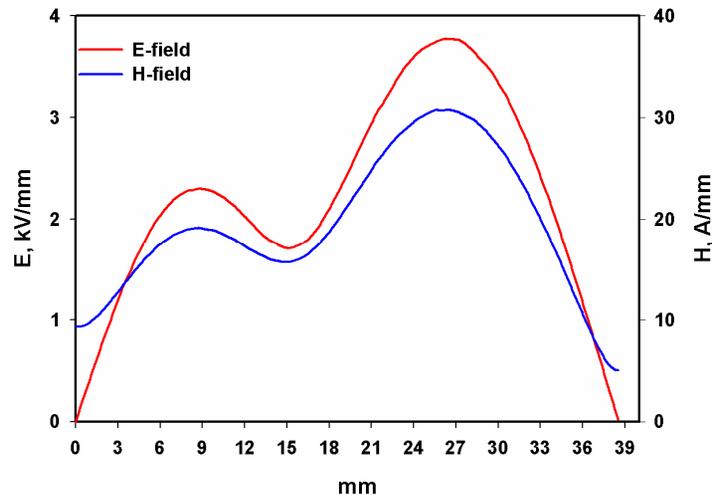
TE₀₁-TE₀₂ TW window



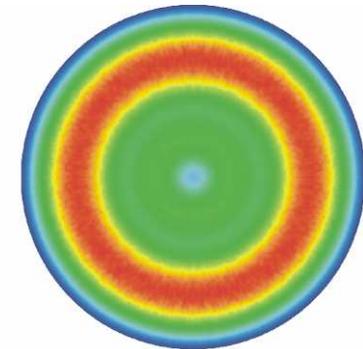
Parameters of TE₀₁-TE₀₂ TW window

Operating frequency	11424 MHz
Passband (SWR < 1.2)	600 MHz
Ceramic diameter	77.1 mm
Ceramic thickness	2.12 mm
Ceramic permittivity	9.8
Max. E-field on the ceramic (100 MW)	3.77 kV/mm
Max. E-field on the periphery	0
Max. E-field on the metal	0

Expected power limit 450 MW ! (8 kV/mm)



E-field

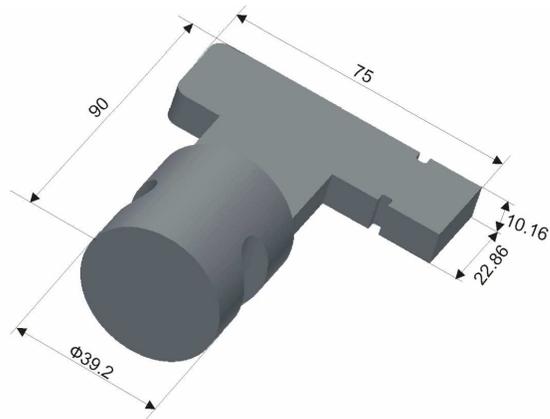


H-field

Electric and magnetic fields on the ceramic surface of X-band TE₀₁-TE₀₂ TW window at power of 100 MW

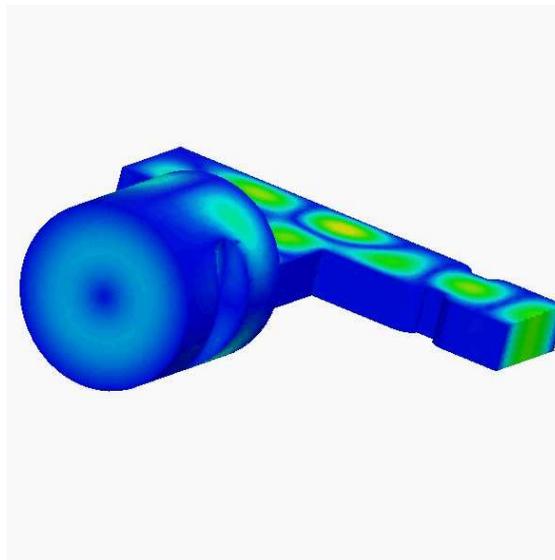
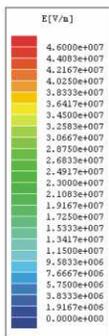
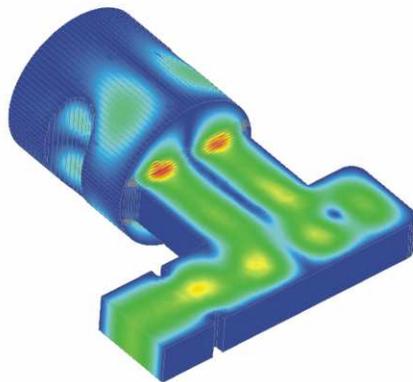
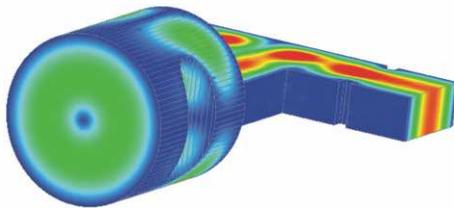
*S.Kazakov, "New Compact TE₁₀ – TE₀₁ Mode Converter and TE₀₁-TE₀₂ Window", ISG-8, SLAC, June 2002.

But we need TE₁₀-TE₀₁ mode convertor for this window.
 The mode converter with wonderful simple geometry was design.



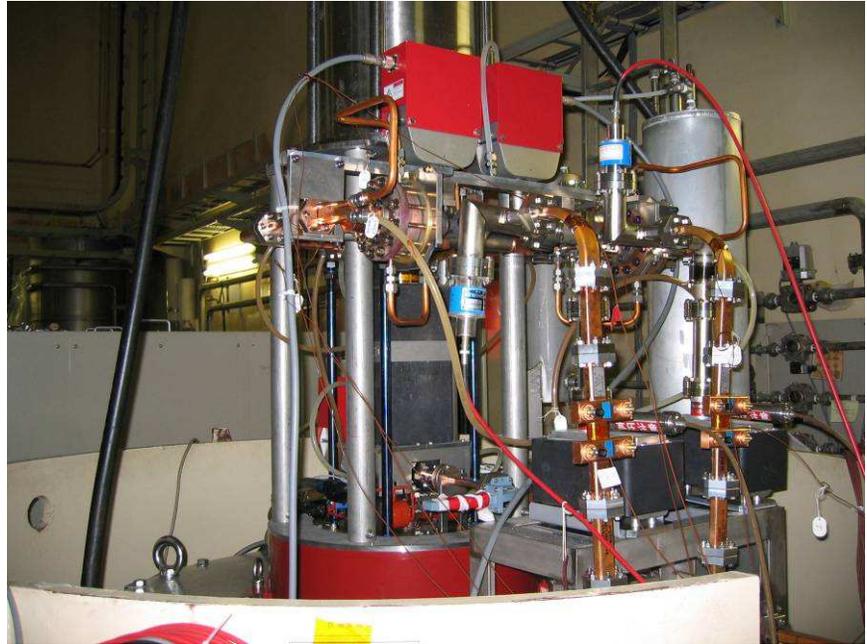
Parameters of TE₁₀-TE₀₁ mode convertor

Operating frequency	11424 MHz
Modes to convert	TE ₁₀ ← → TE ₀₁
Efficiency of conversion	99.99%
Passband (Eff. > 99%)	240 MHz
Max. E-field on the surface (at 100 MW)	46 kV/mm

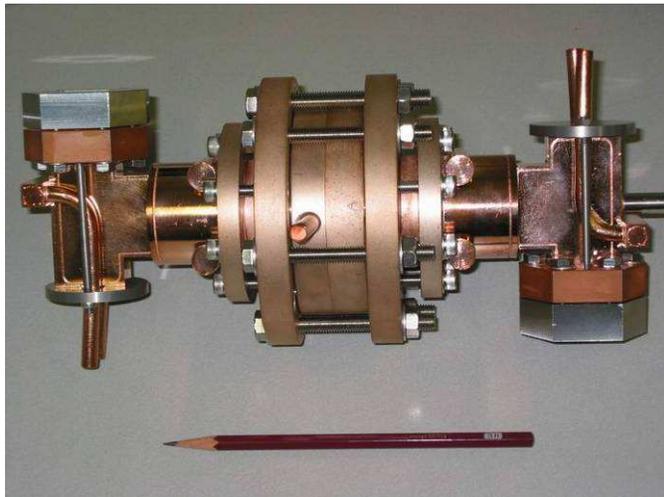




Window without converters



Toshiba/KEK X-band klystron with new windows



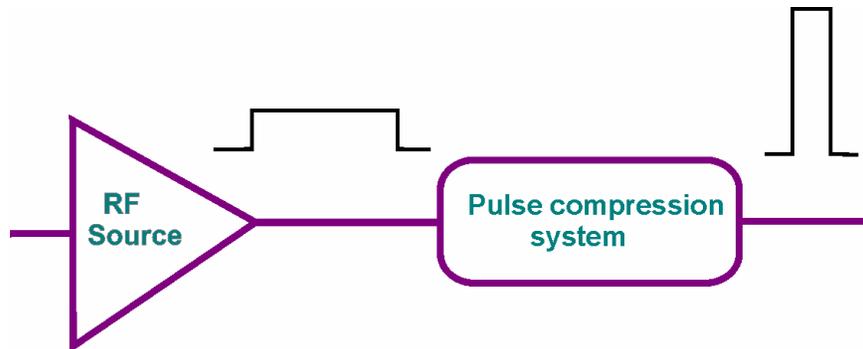
Assembled window with converters.

Behavior of new window is much “quieter” than behavior of previous type.

Pulse compression.

Experience says that it is easier (cheaper) to create RF sources with lower power and longer pulse than higher power and short pulse.

We can reduce cost by using the lower power / longer pulse RF sources together with less expensive pulse compression systems



Pulse compression systems can be divided into “active” and “passive” systems

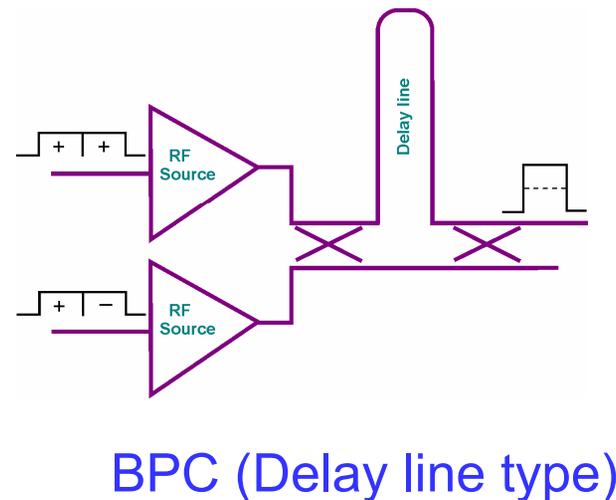
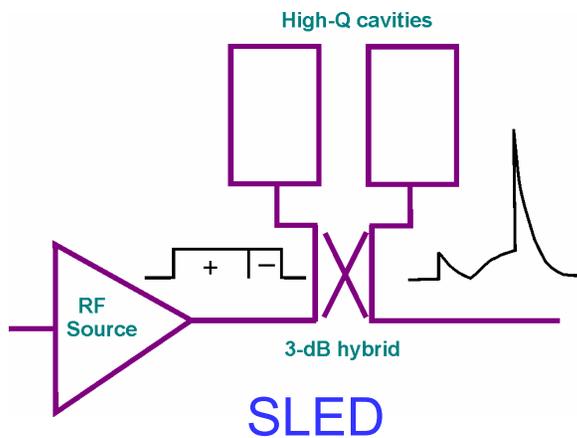
Active systems use “active” elements, which change their RF properties under external control.

Several laboratories work on active systems:

SLAC (silicon, ferromagnetic), Nizhniy Novgorod (plasma tubes, silicon), Omega-P/Yale (ferroelectrics).

But powerful enough active pulse compression system still does not exist.

Elements of passive systems do not change RF properties. Systems utilize only phase manipulation of RF sources. Passive systems can be divided roughly into SLED type and delay line type.



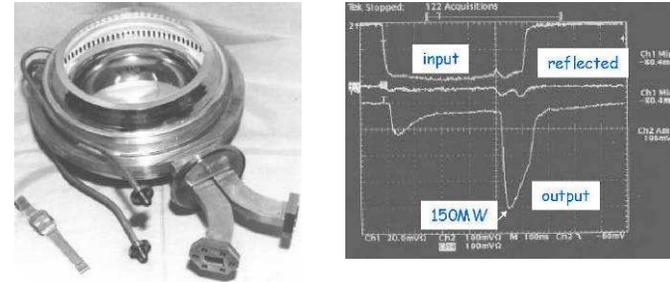
SLED Type:

Classic SLED (Z.D. Farkas,P.B. Wilson)

VPM (VLEPP Power Multiplier), based on open cavity
(V.E. Balakin, I.V. Syratchev)

Advantage: compactness.

Disadvantages: efficiency < 80%, 'funny' output shape



14GHz VPM and pulses during test

Delay line type

Binary Pulse Compression System (BPC) (Z.D.Farkas)

Delay Line Distribution System (DLDS) (H.Mizuno)

Advantages: high efficiency, flat pulse shape.

Disadvantages: big sizes



Cold model of DLDS in KEK tunnel

“Hybrid” Type (SLED + Delay line)

SLED-II (P.B. Wilson, Z.D. Farkas, R.D. Ruth)

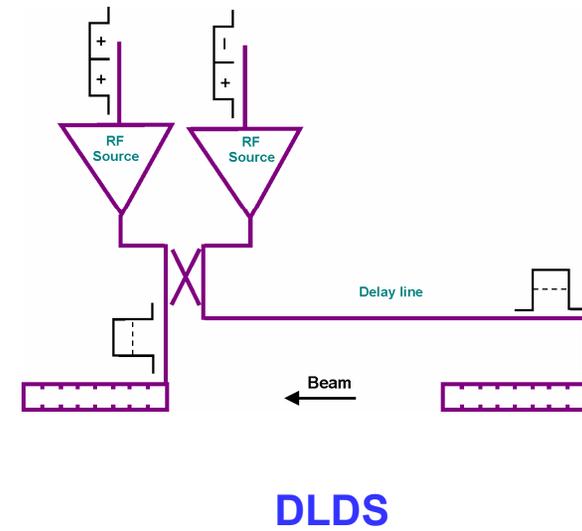
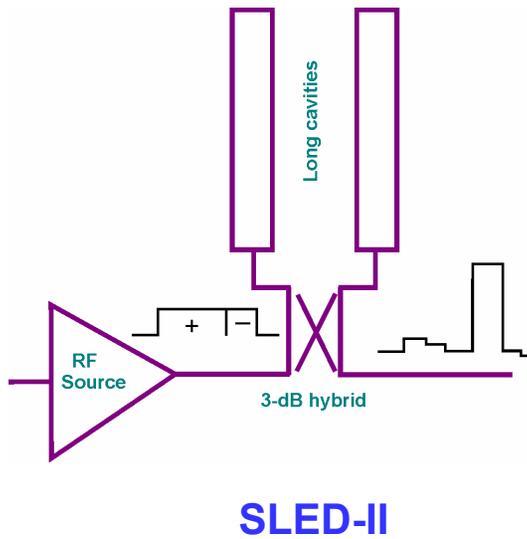
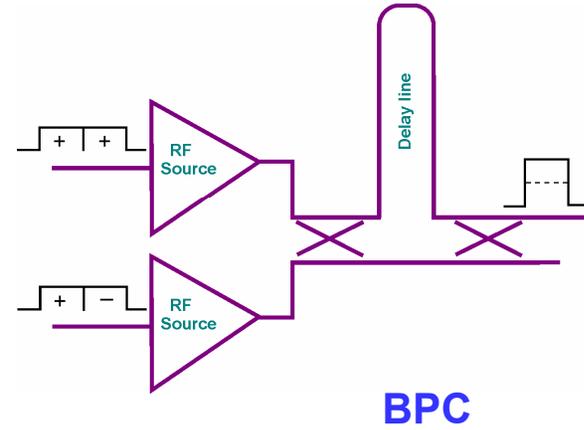
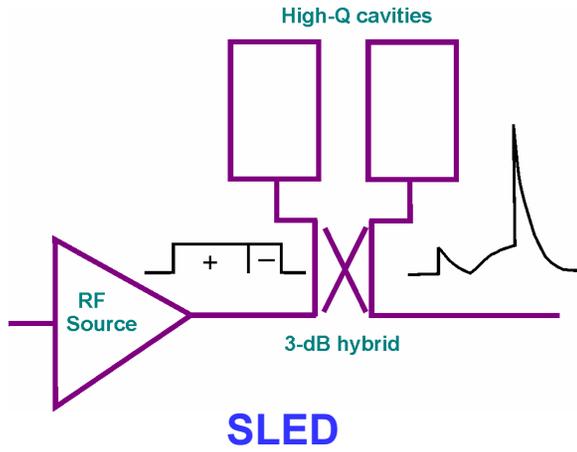
Advantages: flat pulse shape, smaller size than BPC and DLDS

Disadvantages: not 100% efficiency, still big sizes



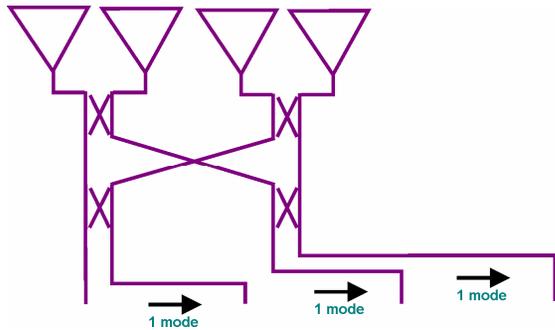
SLED-II pipes in SLAC

Principle of operation

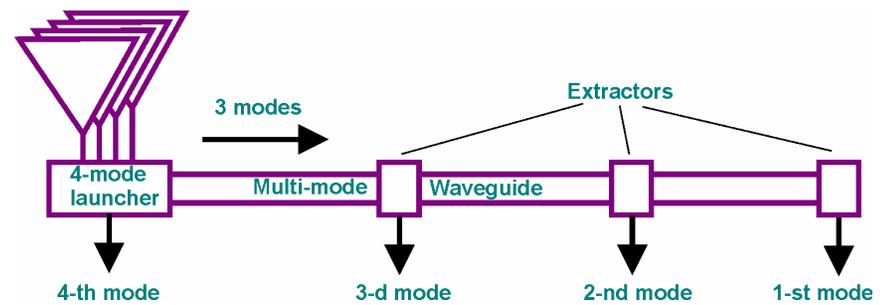


We were excited when H. Mizuno invented DLDS –flat top pulse, 100% efficiency and length of pipe was twice smaller than for BPC.

DLDS can have several output ports and can be multimode:



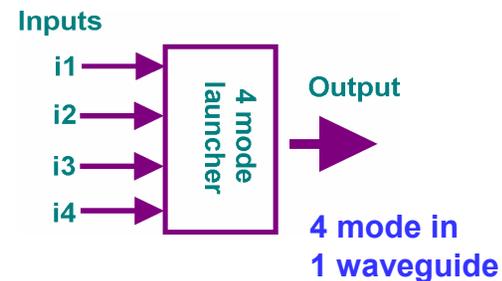
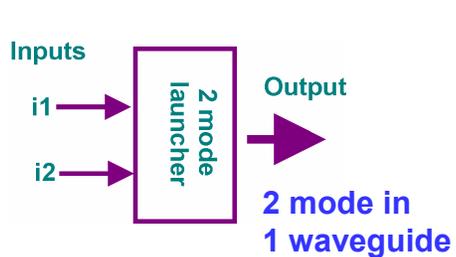
Single mode, 4 port DLDS



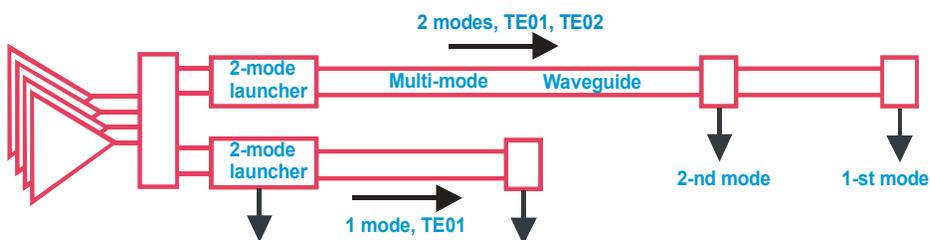
4 mode, 4 port DLDS

DLDS was chosen as Pulse Compression System for JLC and NLC.
SLAC choose 4 mode 4 port DLDS.
KEK was more cautious and choose 2 mode 4 port scheme.

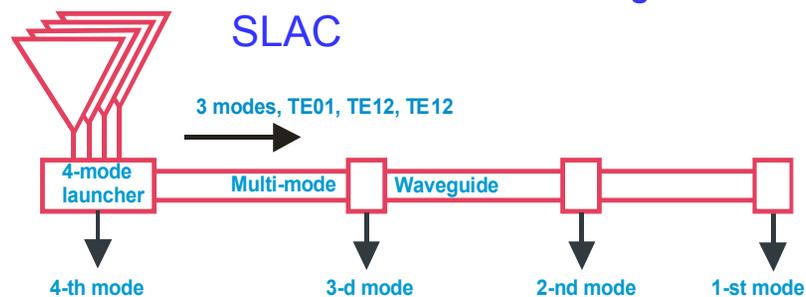
Key element of multimode DLDS is mode launcher



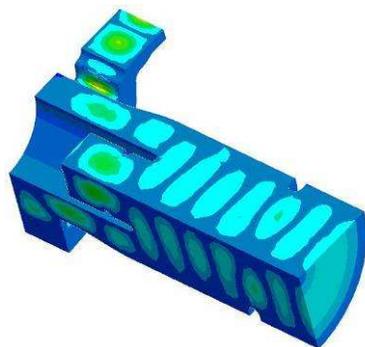
KEK



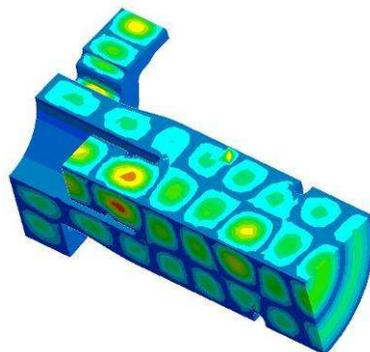
SLAC



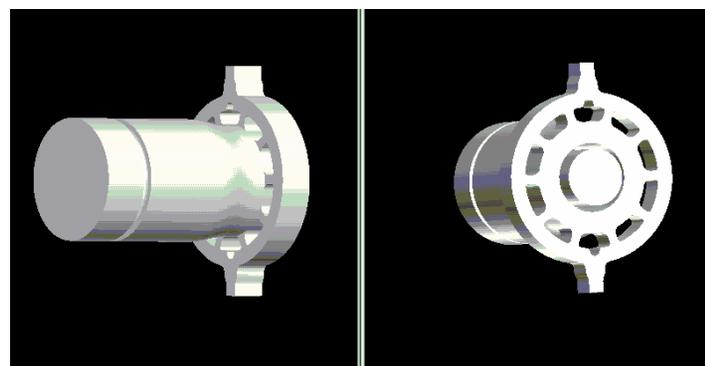
TE₀₁-TE₀₂ launcher-extractor*



Launching TE₀₁



Launching TE₀₂

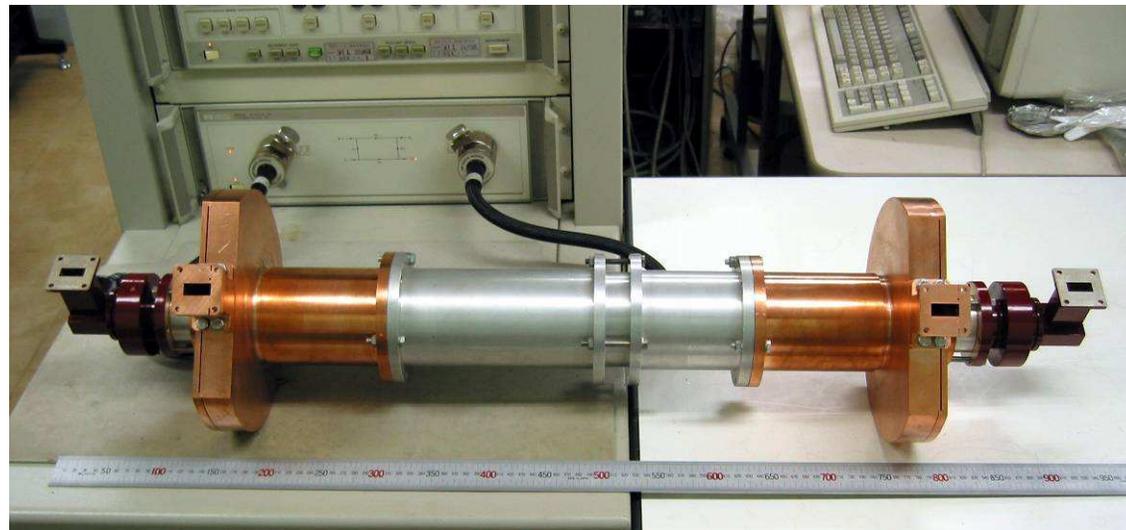


*S.Kazakov, "TE₀₁-TE₀₂ DLDS Elements", RF Pulse Compression Distribution Workshop, SLAC, October 2001

TE₀₁-TE₀₂ launcher-extractor



Part before brazing

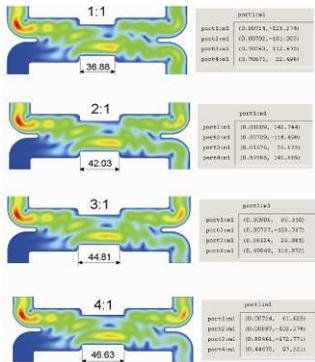


Under test

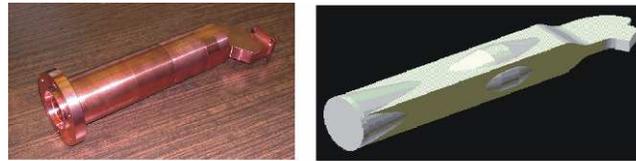


Assembled with tapers

Menagerie of elements designed for Pulse Compression



Power divider for any ratio



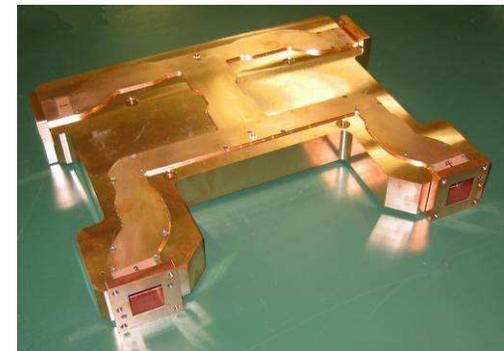
TE₁₀-TE₀₁ Converter-Launcher



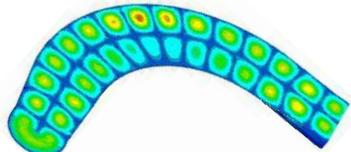
TE₁₁-TE₁₂ mode converter



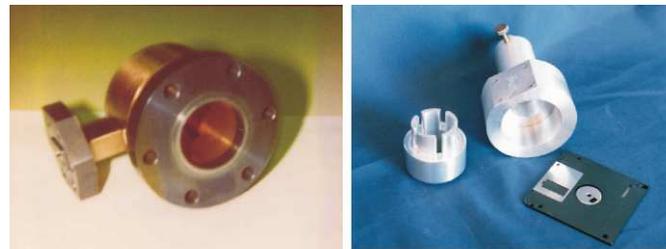
TE₀₁ power splitter



Oversized SLED-II head



TE₀₁ bend



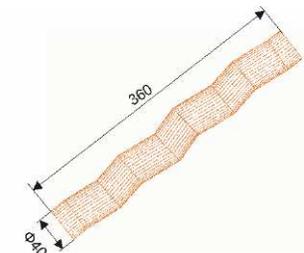
TE₁₀-TE₀₁ Wrap-around Mode Converter



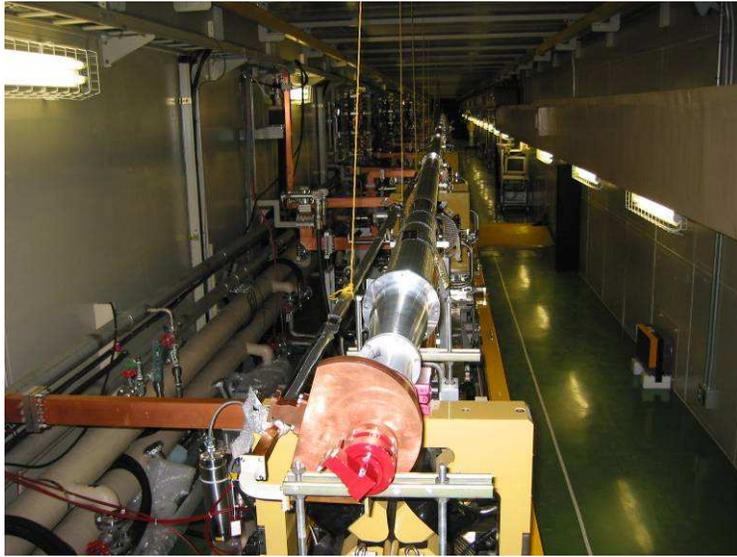
TE₁₀-TE₀₁ Choke Mode Converter



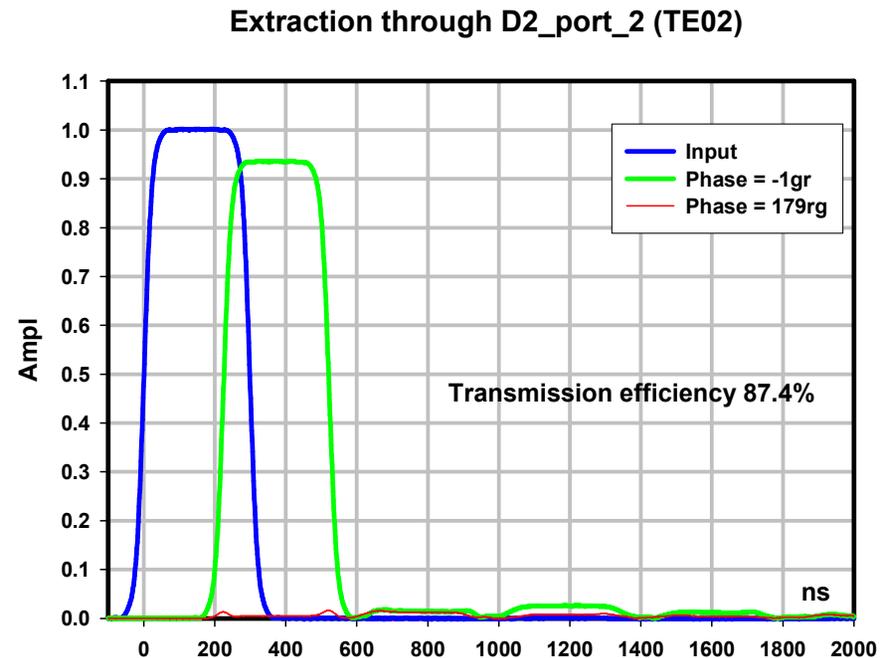
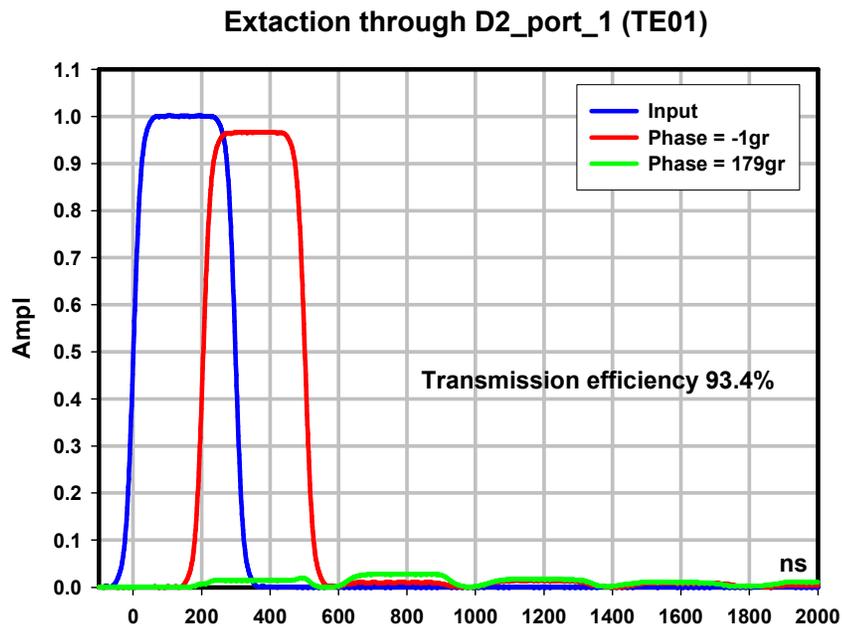
TE₁₁-TE₀₁ Zigzag Mode Converter



Cold measurements of one 55-m arm of TE01-TE02 DLDS was carried out.



Results of cold measurements of 55 m TE₀₁-TE₀₂ DLDS arm.

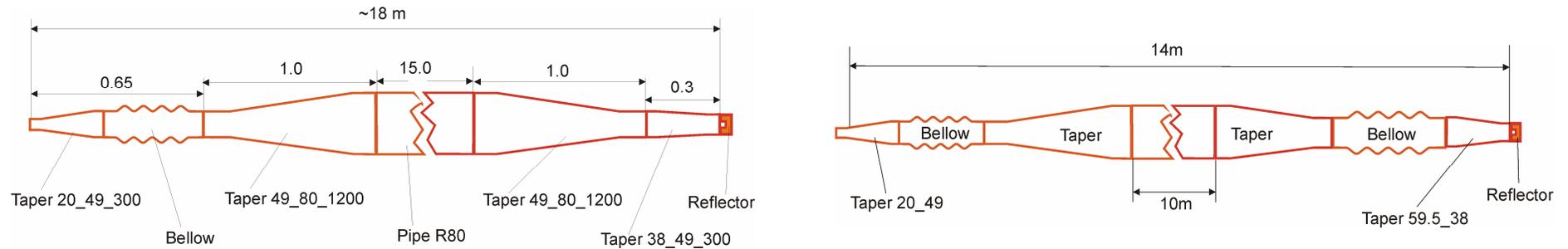


Conclusion:

The results of cold measurements are in good agreement with theoretical estimation. Based on the results we can expect efficiency about **90%** for TE01 part ($0.922 * \text{Eff.}$ Another 50m part) and **88%** for TE02 part. **Average efficiency about 89%.**

Elements of TE01-TE02 DLDS were successfully designed and tested.

Some interesting ideas now to make delay shorter for SLED – II*



The efficiency (without loss in 3dB hybrid etc.) ~ 82%. Power gain ~ 3.26.

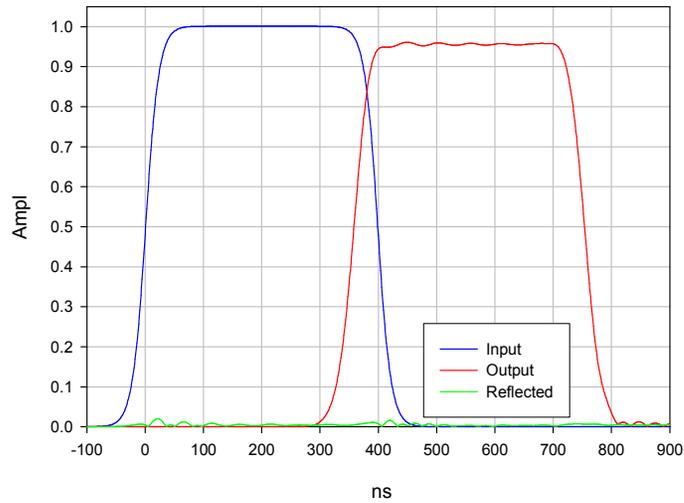
It was surprising, but for simple cylindrical overmoded cavity there is combination of sizes when pulse is transmitted without distortion but with a big delay. Delay time is close to $L \cdot N_{\text{mode}} / C$. Input mode is TE_{01} , $N_{\text{modes}} - TE_{0n}$ modes



Transmission through 32 cell

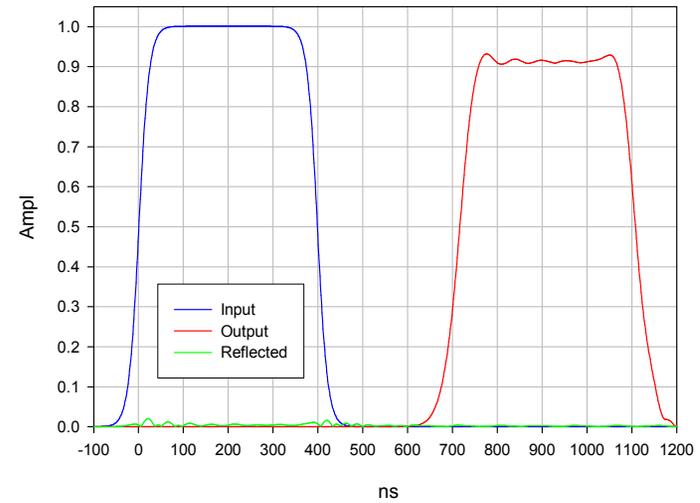
Eff = 0.9118, Loss = 0.02477%/ns

Del = 21.33ns/m, Shape = 1.23e-2

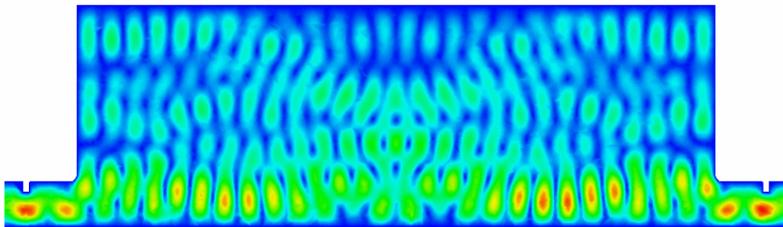


Transmission through 64 cell

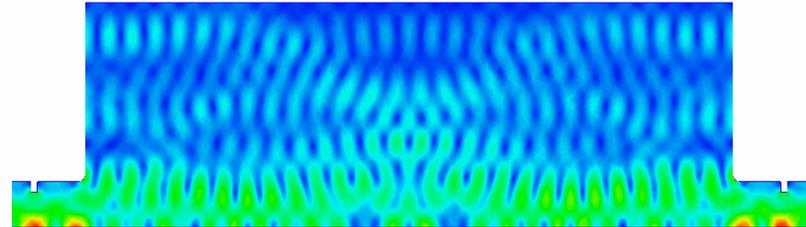
Eff = 0.8311, Shape = 3.20e-2



Field patterns in the cell

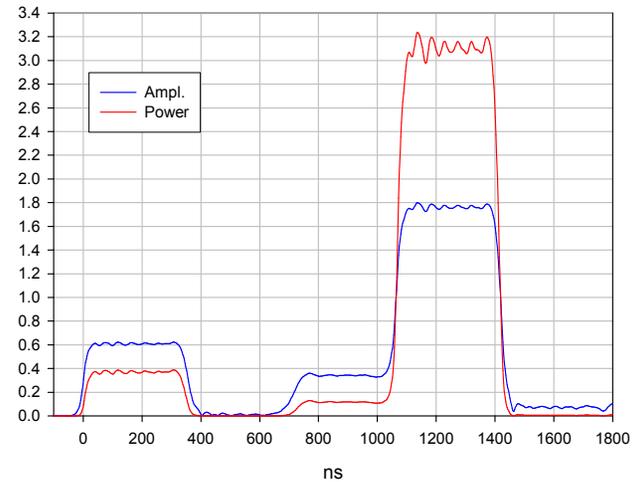
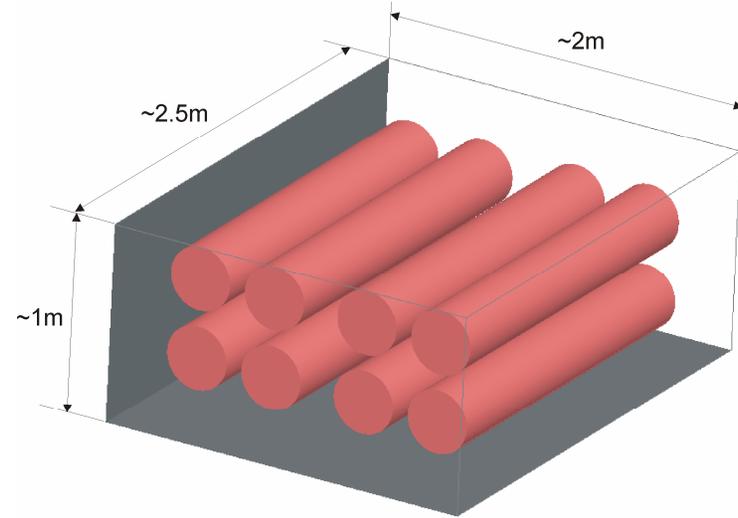
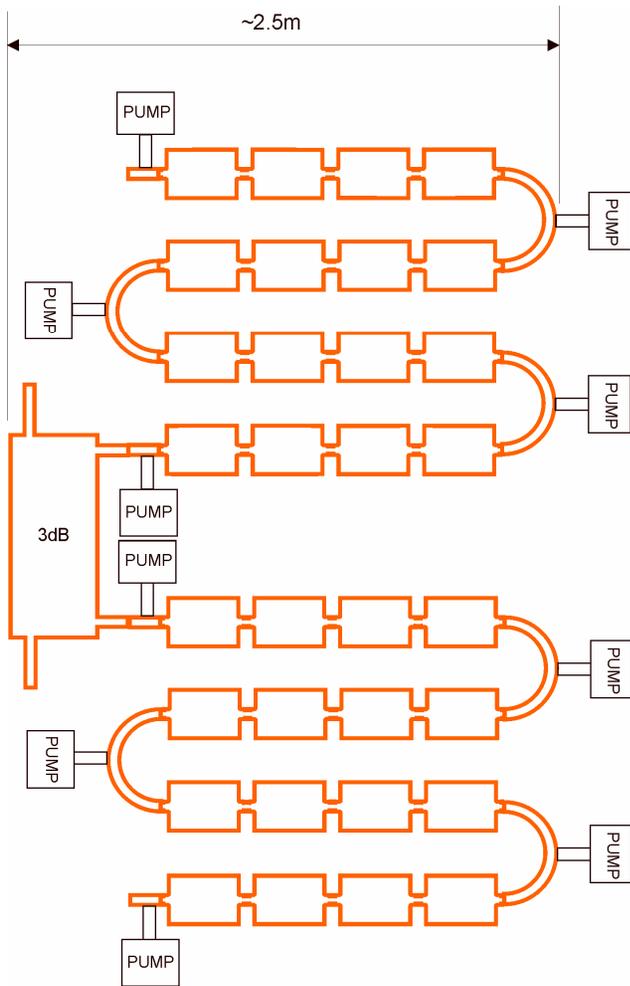


Electric field



Magnetic field

SLED-II can be packed into a box 1m x 2m x 2.5m:



Result of simulation of SLED-II based on 16-cell delay lines.

Power gain ~ 3.1, Power variation ~ 7%,



Then climate changed suddenly...

Climate changed suddenly, but not too suddenly. Some time before, the design of L-band Multi-Beam Klystron (MBK) was started to have some MW for cold time.

Design team:

Y.H. Chin (guidance)

A.Larionov (gun, optic)

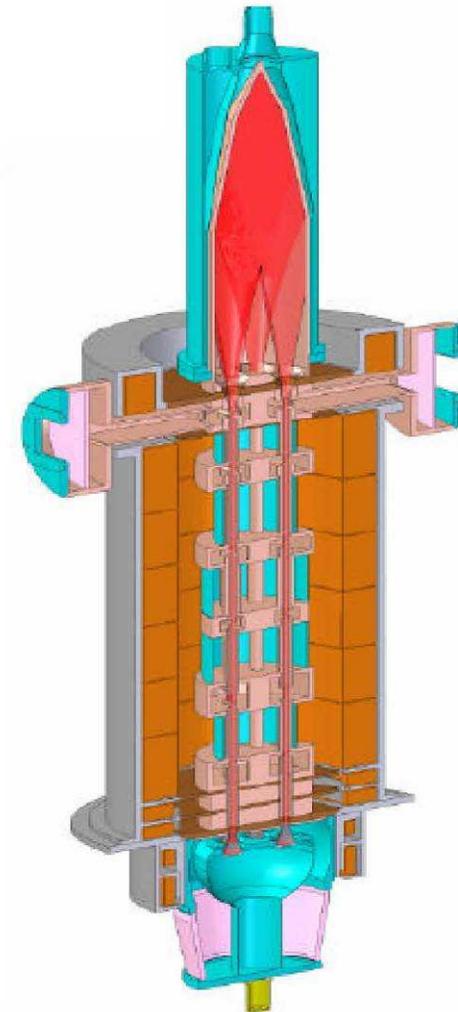
V.Teryaev (RFstructure)

S.Kazakov (3D output cavity)

TOSHIBA (mechanical design, thermal analysis, manufacturing)

Design parameters

Frequency	1300	MHz
Output Power	10	MW
Average Output Power	150	kW
Beam Voltage	115	kV
Beam Current	132	A
Efficiency	>65	%
RF Pulse Width	1.5	ms
Repetition Rate	10	pps
Saturation Gain	47	dB
Number of Beams	6	
Cathode Loading	<2.1	A/cm ²
Structure	6	cavities
RF Window	Pill Box WR-650	
Tube Length	2270	mm
Solenoid Power	<4	kW



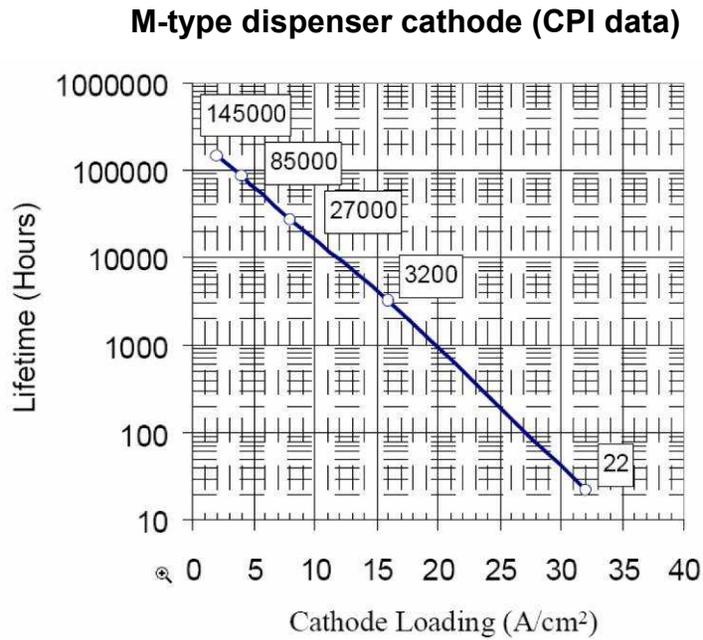
THE TOSHIBA E3736 MULTI BEAM KLYSTRON

S. Miyake, A. Yano (Toshiba Electron Tubes & Devices Co., Ltd., Japan)

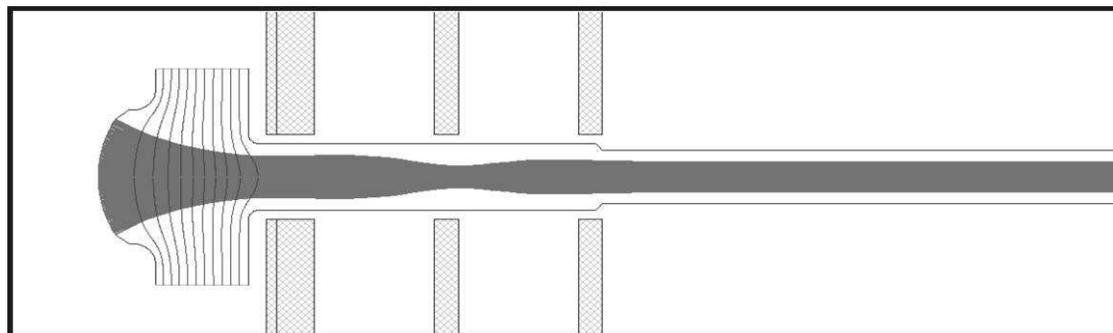
S. Kazakov, A. Larionov, V. Teryaev (BINP, Russia)

Y. H. Chin (KEK, Japan)

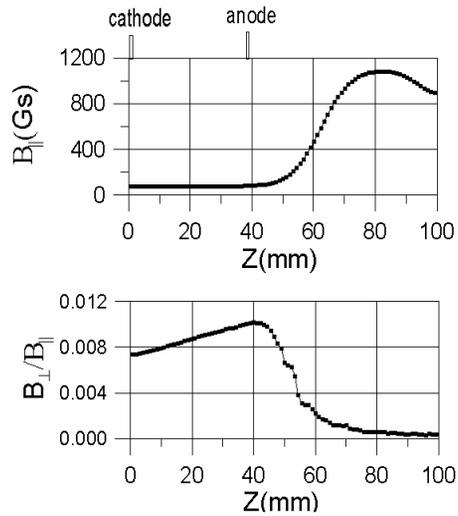
To increase the lifetime of cathodes, the cathode current density was reduced $< 2.1 \text{ A/cm}$



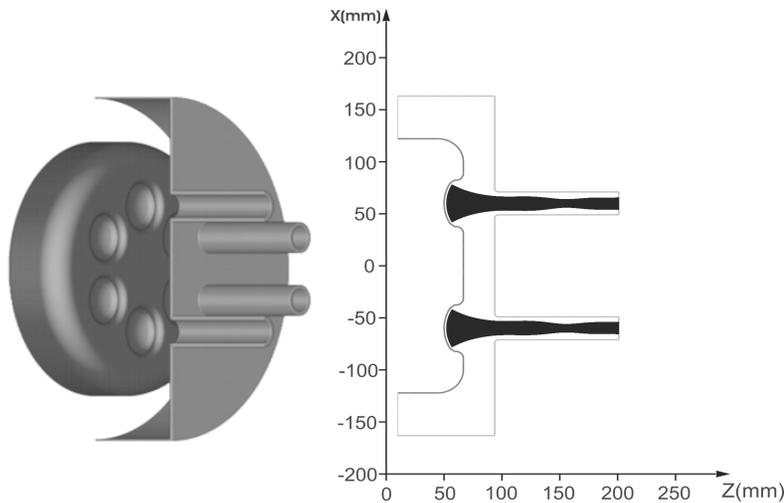
Cathode of TOSHIBA E3736 MBK



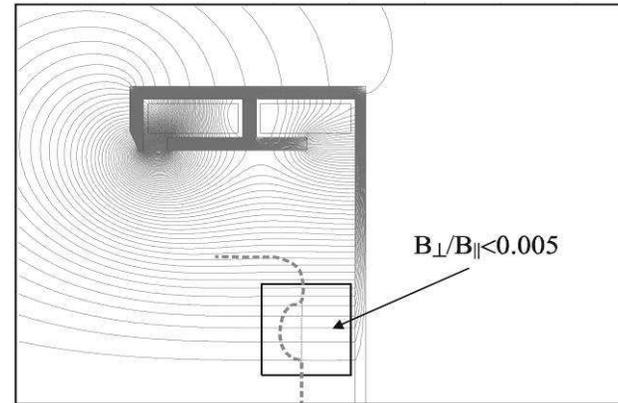
The main challenge was to form magnetic field with high local “axial” symmetry at the area of beamlets, by making beamlets 2D.



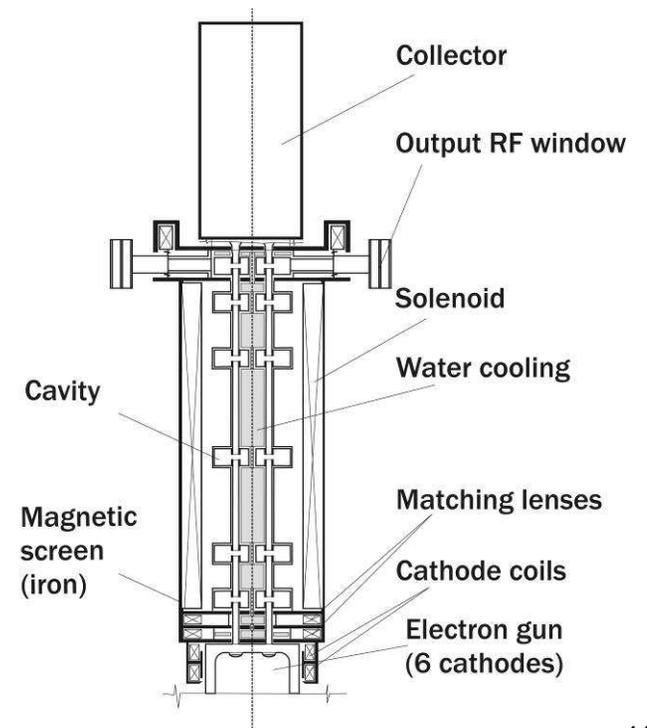
Magnetic field distribution along axis of the individual beamlet.



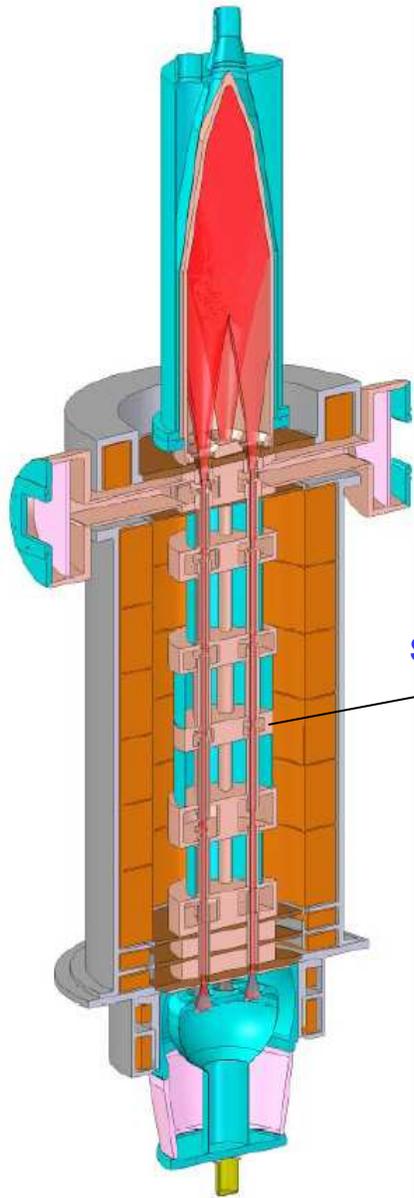
Example of gun modeling by GUN3D code. (A.Larionov)



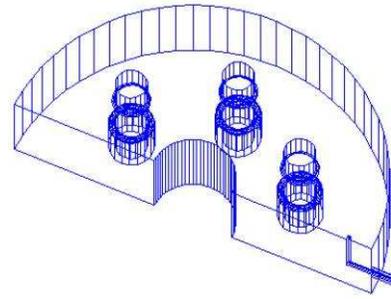
2D analog of magnetic system for the magnetic field forming at the gun (A.Larionov)



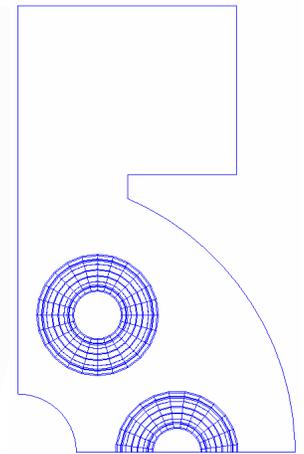
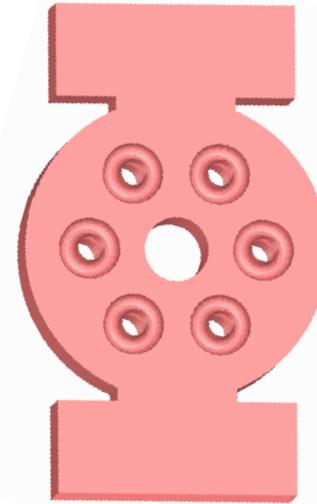
Cavity system of MBK



Second harmonic cavity

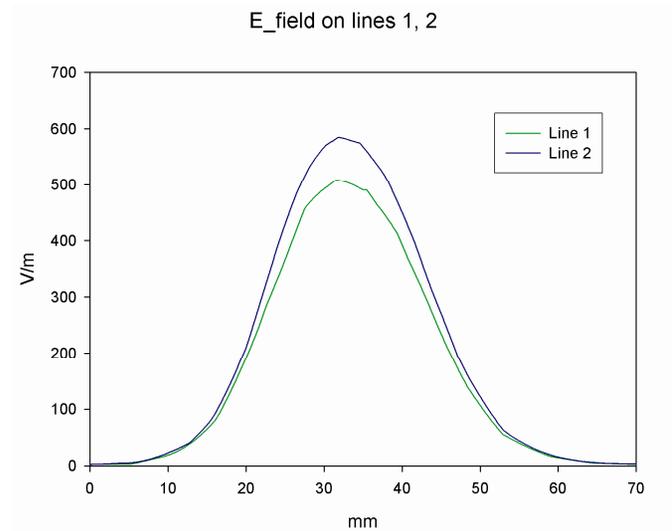


Input cavity (half)

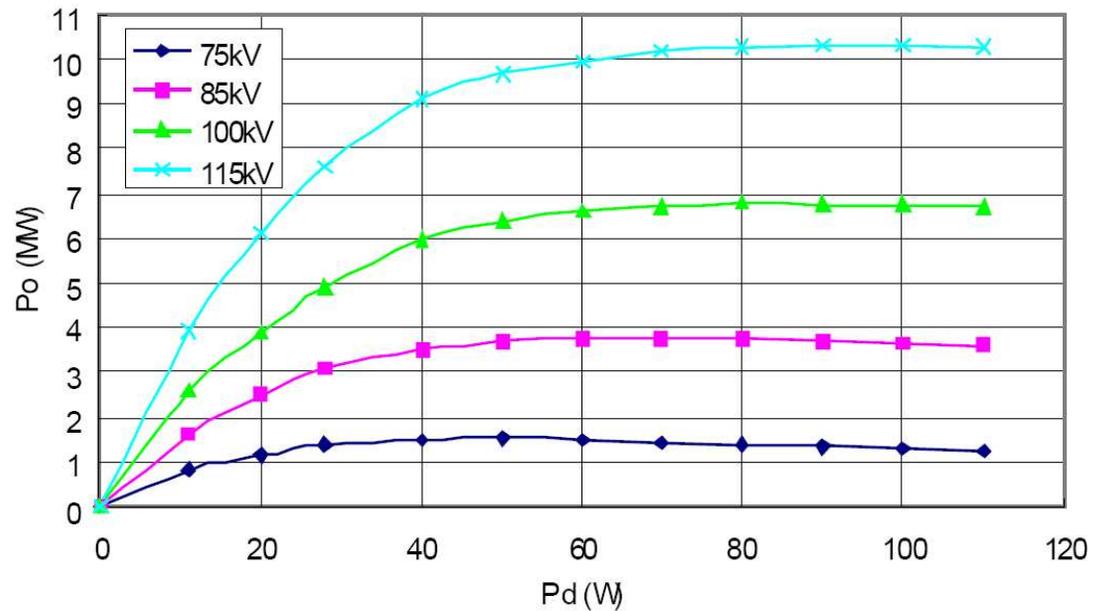


Output cavity

Second harmonic cavity made klystron noticeably shorter



Electric field profile at the axis of different beamlets



Measured performance

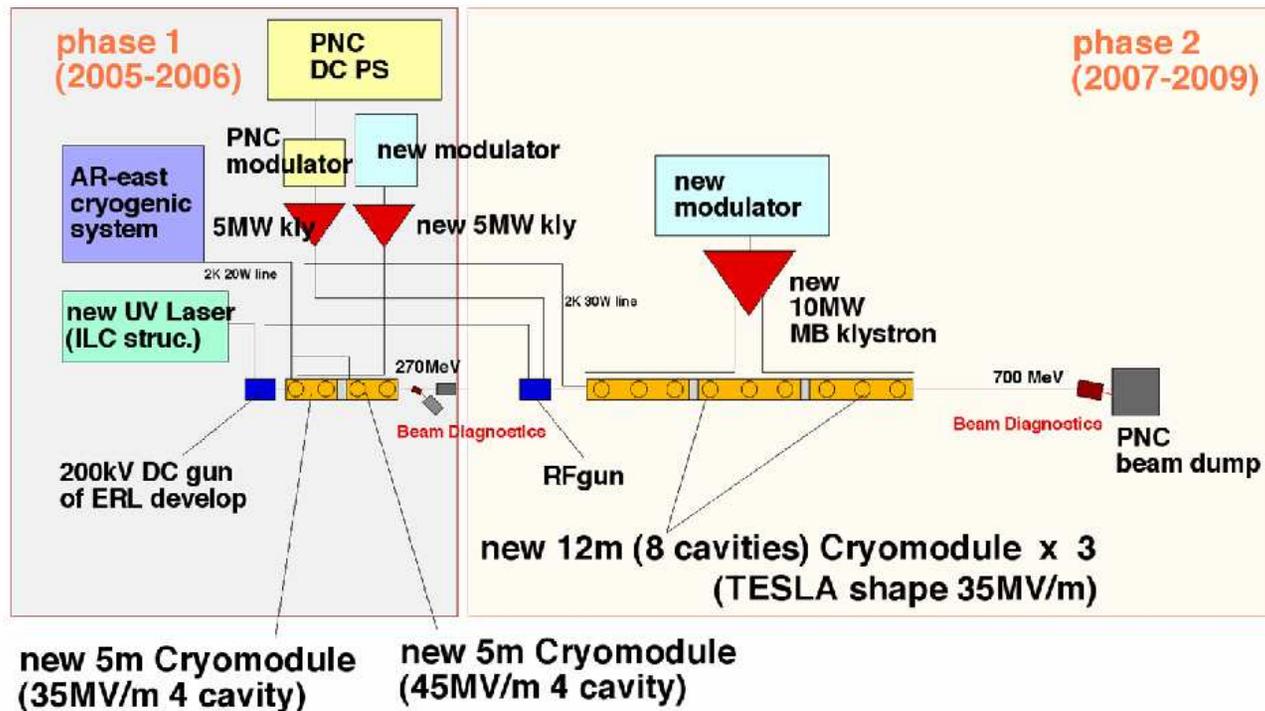
- Voltage: 115kV
- Current: 135A
- Output Power: 10.4MW
- Efficiency: 67%
- Pulse duration: 1.5ms
- Rep. Rate: 10Hz



STF in KEK

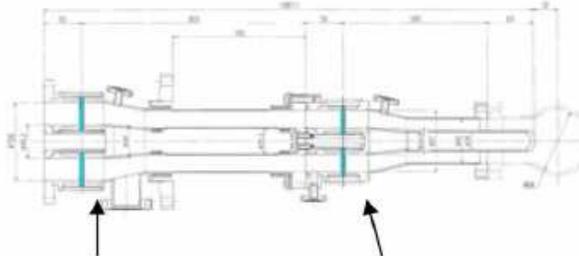


Plan of Superconducting RF Test Facility (STF)



TESLA style Baseline Cavity Package

Two Disk Window Input Coupler



Warm Window Cold Window



Warm Coupler & Cold Coupler



HOM Coupler

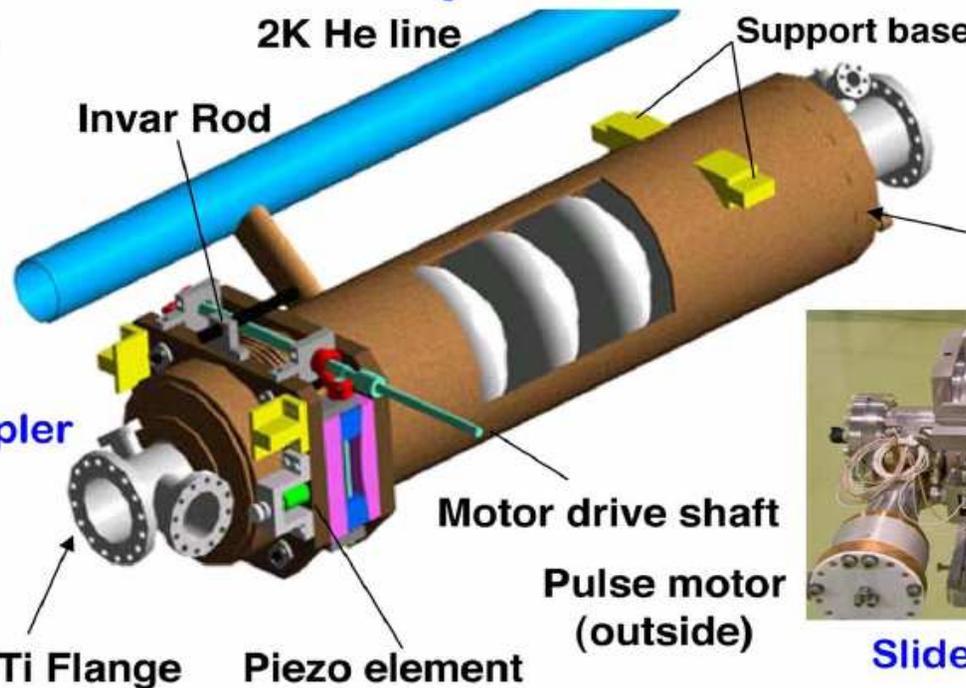


a Cavity covered with Ti Jacket

2K He line

Support base

Invar Rod



3 Cavities (Vertical test)

Titanium Jacket

Motor drive shaft

Pulse motor (outside)

Nb/Ti Flange

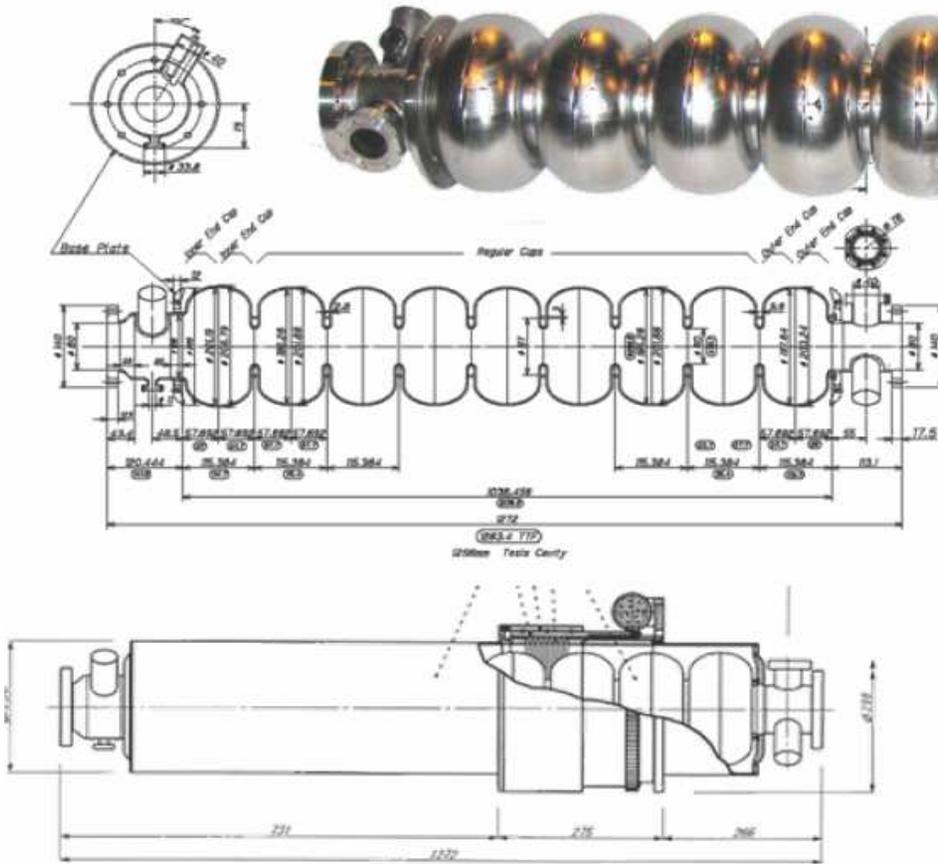
Piezo element



Slide Jack Tuner

Total 4 9-cell cavity were fabricated.

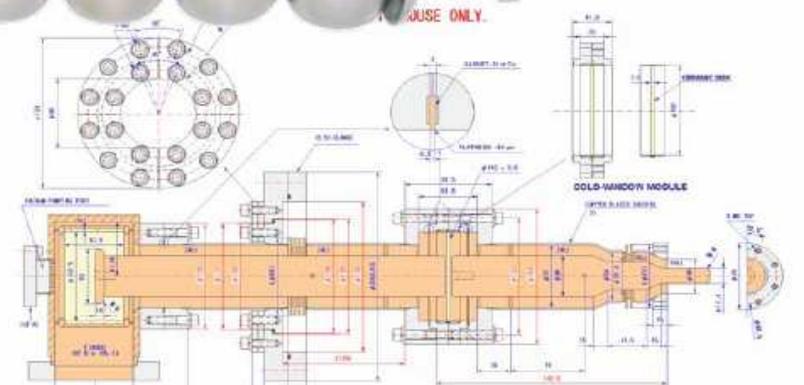
LL ICHIRO cavity package for High Gradient



Coaxial ball screw tuner

Dia: 276mm
Lead: 40mm
Ratio: 21:1

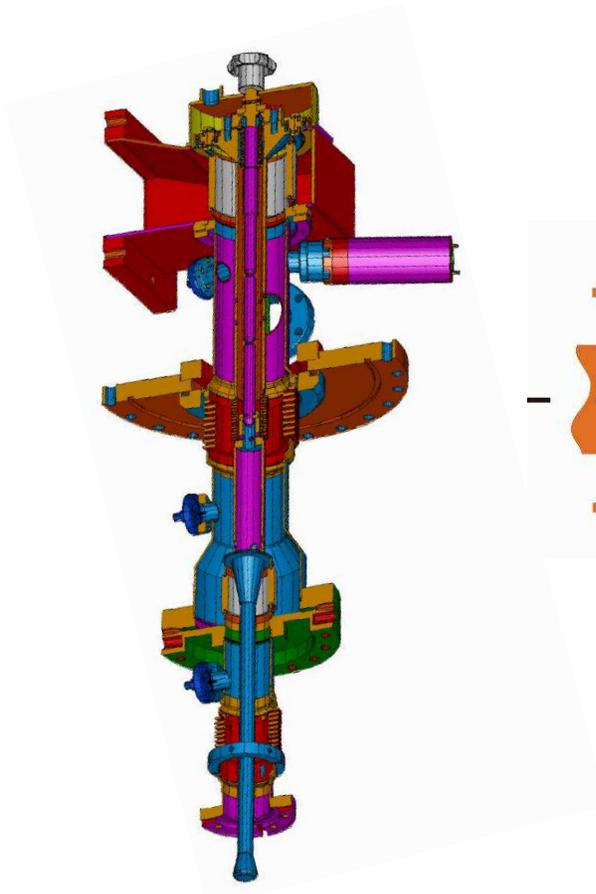
Tuner test stand



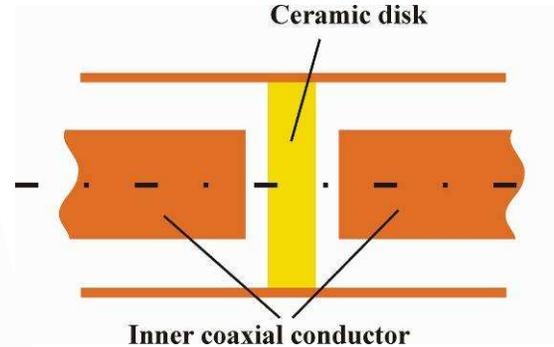
Input Coupler using capacitive coupling



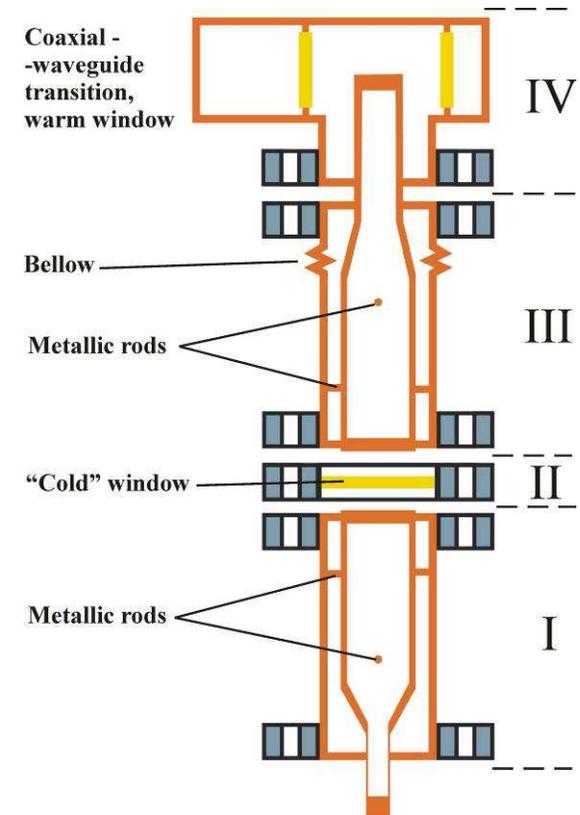
The present TTF-III coupler is quite complicated end expensive. We decided to try to design and build a cheaper coupler. As cold window we chose a simple disk window. In this case connection between cold and warm part become simple and coupler can be divided into several modules.



TTF-III Coupler

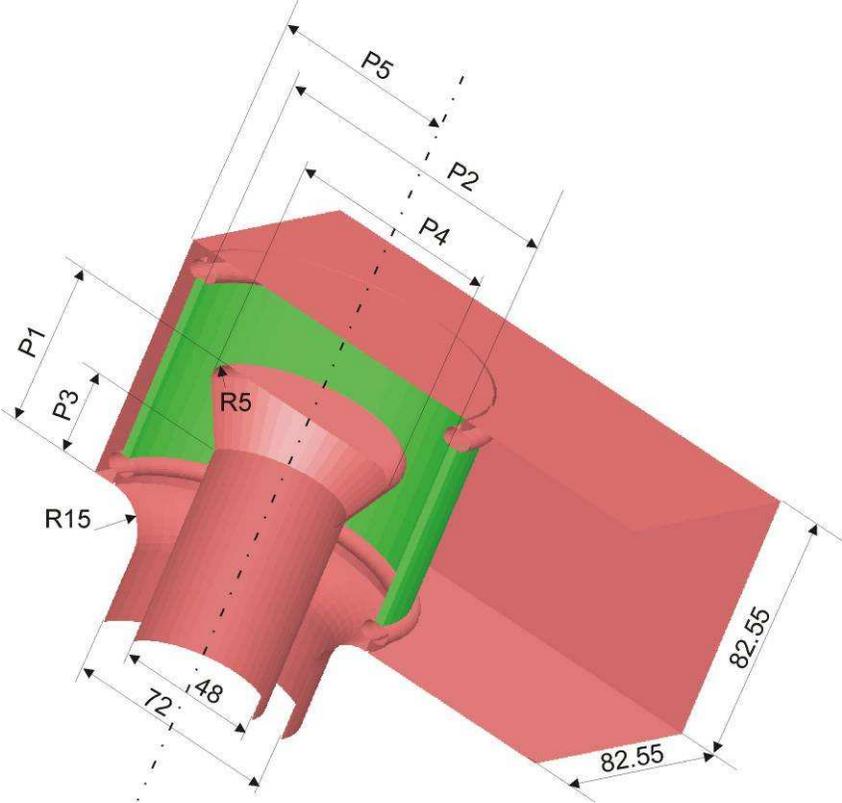
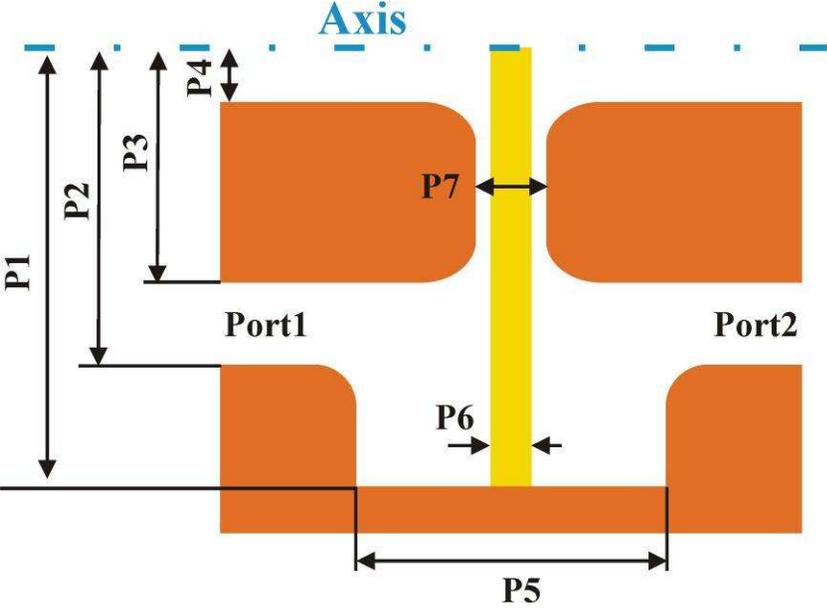


Idea of cold window with capacitive coupling



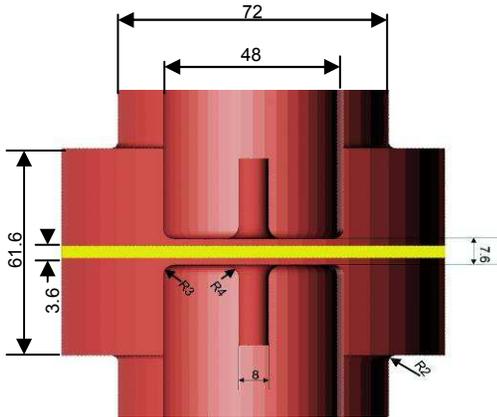
Possible module structure of new coupler

Cold and warm window were optimized in automatic regime to minimize electric field at the ceramic and in air and to maximize passband. Many thousands of version were analyzed.



Geometries of cold and warm windows and parameters of optimization (P_1 - P_N)

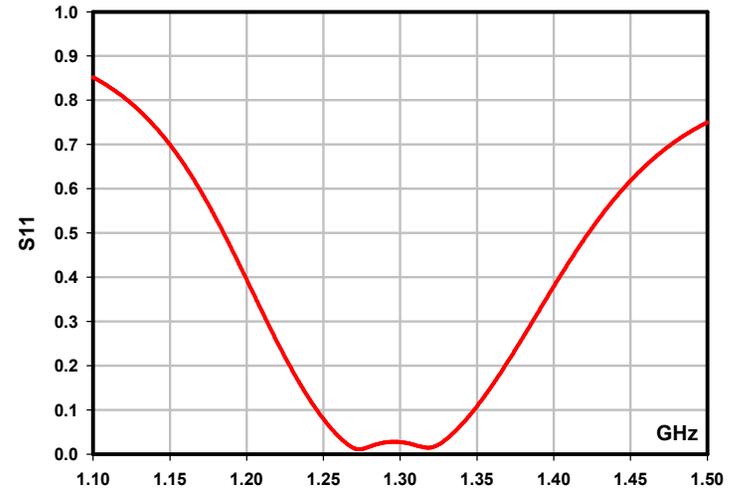
Cold window



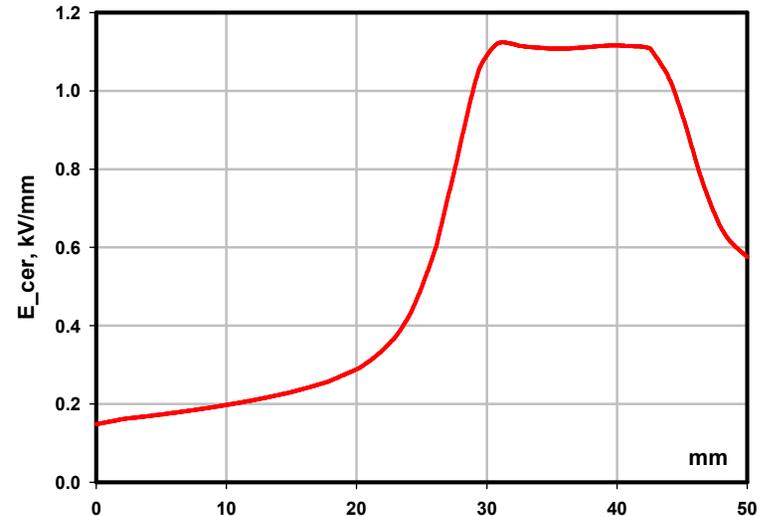
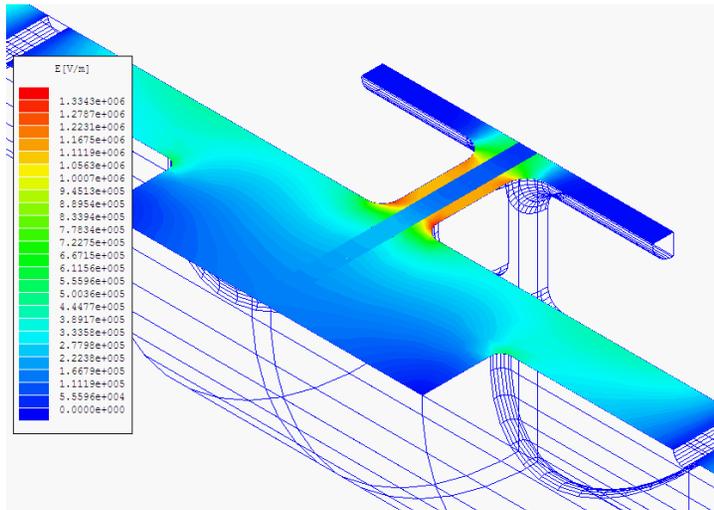
Parameters of cold window:

Ceramic diameter	102mm
Ceramic thickness	3.6mm
Outer coaxial conductor	Φ72mm
Inner coaxial conductor	Φ48mm
Passband	100MHz
Max. E-field on ceramic (500kW)	1.15 kV/mm

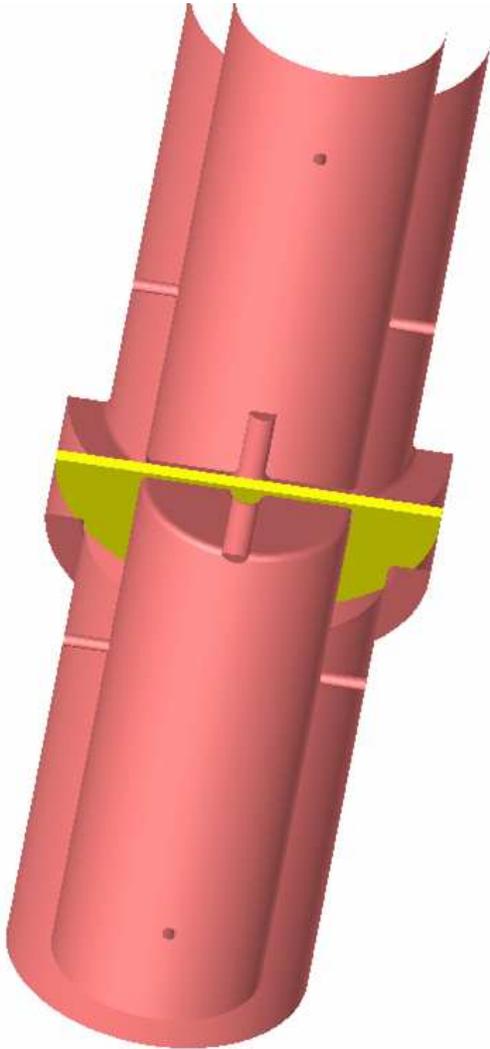
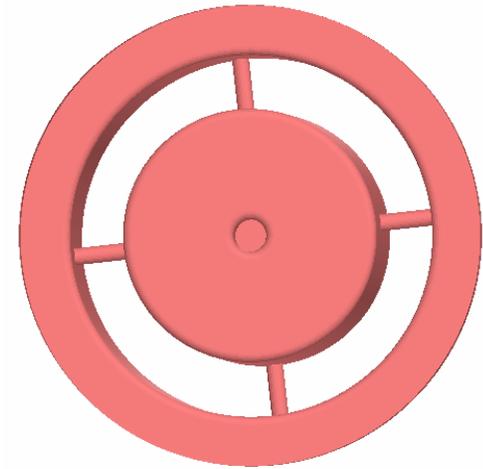
S11 of cold window



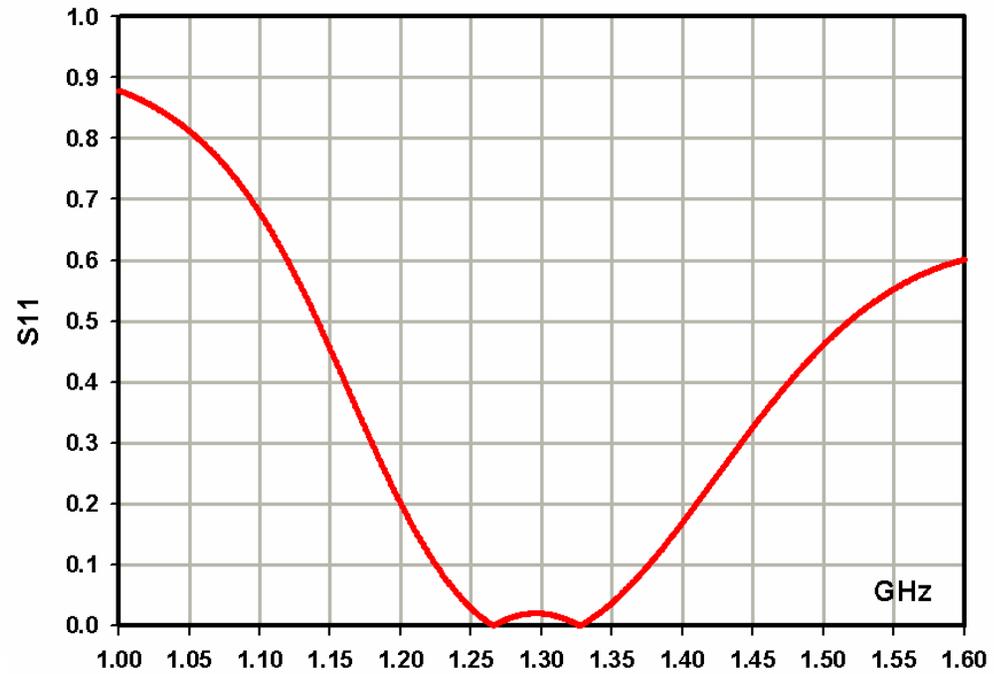
E-field on the ceramic surface for 500kW power



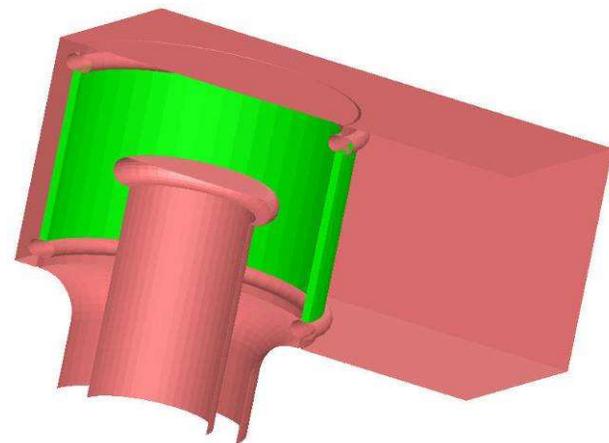
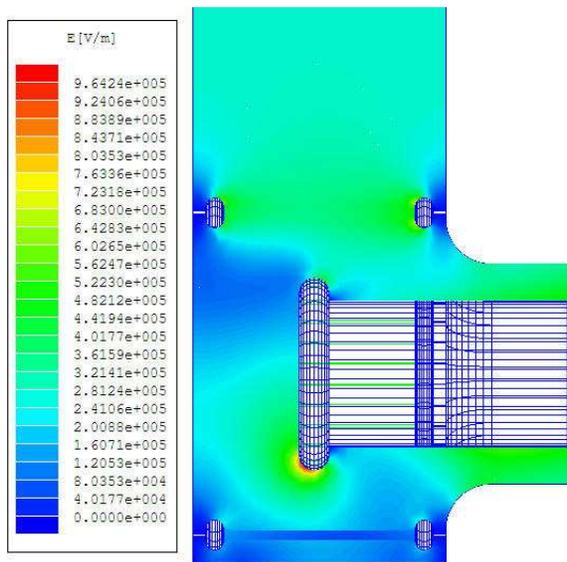
Supports of inner conductors



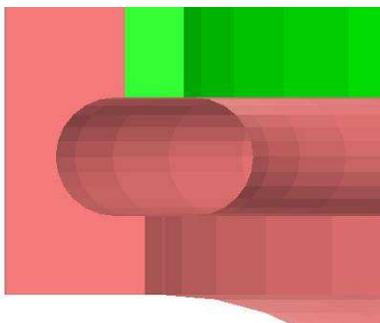
Window r3_5912_4423_1161
supports 180gr,
configuration W, -27mm, S, 12mm, S



Warm window

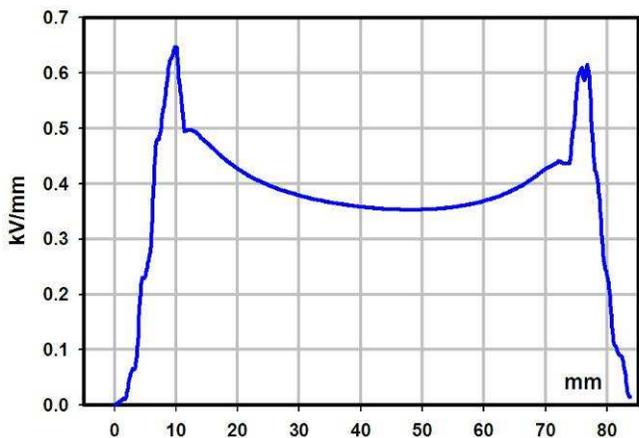


P1 = 48
 P2 = 102.5
 P3 = 38
 P4 = 63
 P5 = 61.94
 R = 3

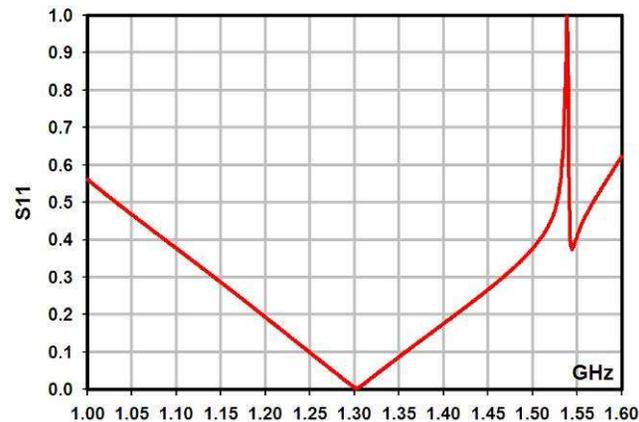


Geometry of ceramic-metal joint

Electric field, 500 kW

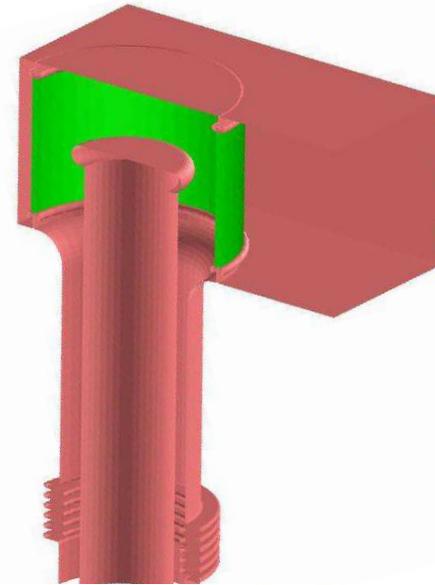
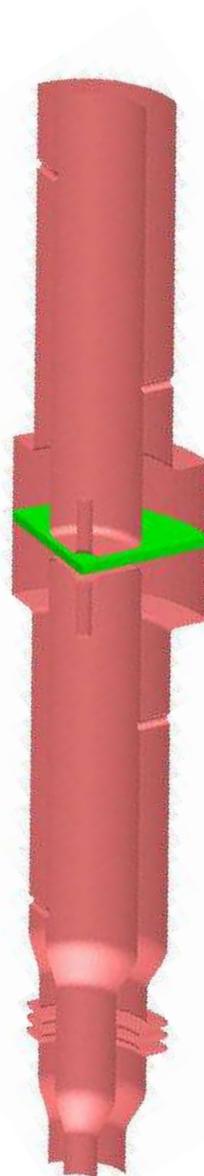
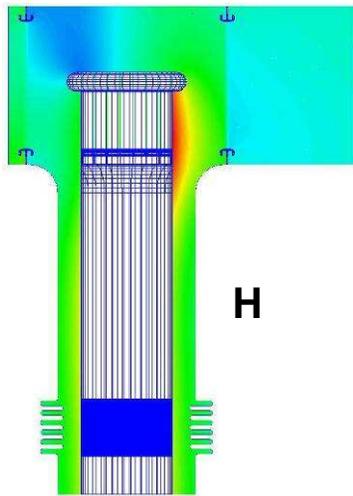
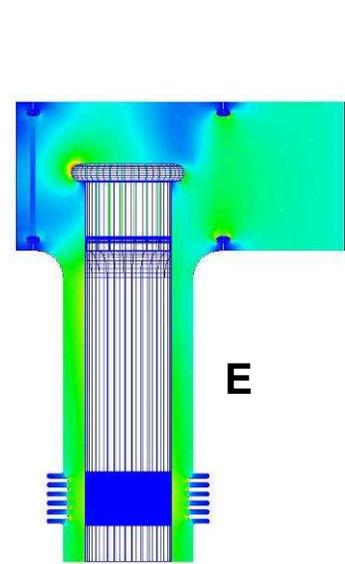


Electric field in air

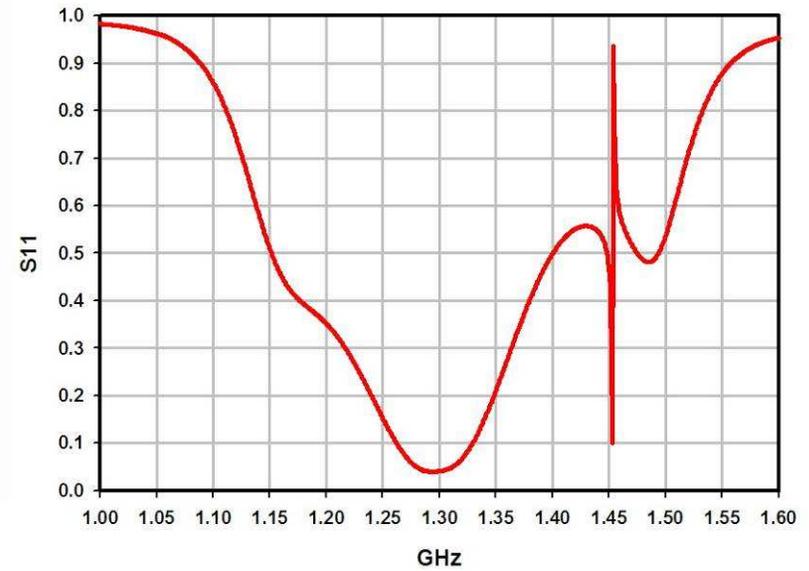


Parameters of warm window (500kW):
 Passband 109 MHz
 Max. E-field in air ~ 0.65 kV/mm
 Max. E-field on the ceram. ~ 0.5kV/mm

Full geometry, example of simulation and passband of coupler.



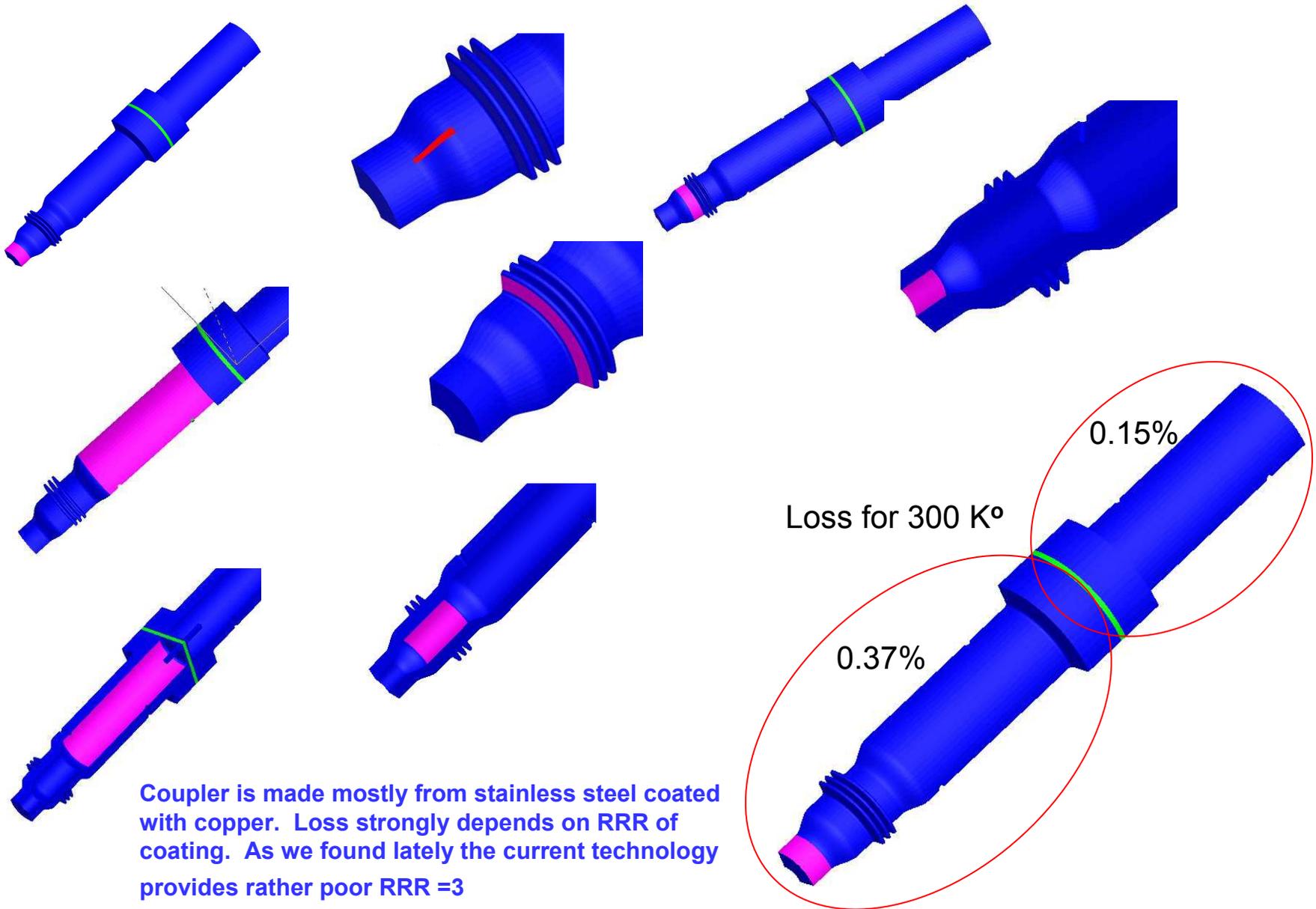
Coupler,
passband of final design



Design parameters of coupler

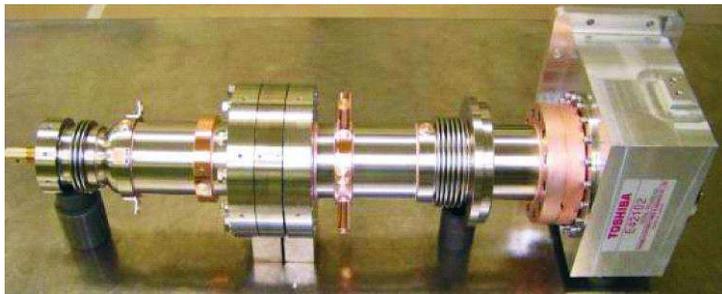
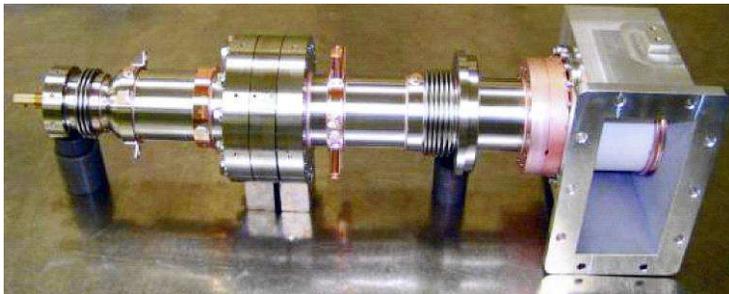
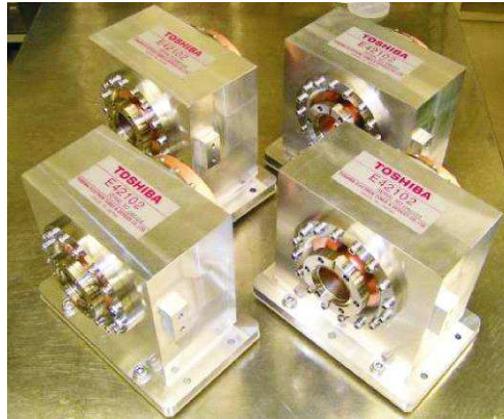
Frequency	1.3 GHz
Input	Waveguide, WR650
Output	Coaxial, D40mm x 17.4mm
Passband	70 MHz (SWR < 1.2)
Max.E-field, cold window	11.5 kV/cm (500kW)
Max.E-field, warm window	5 kV/cm (500kW)
Max.E-field, air	7 kV/cm (500kW)

Example of analyses of dynamic loss



Coupler is made mostly from stainless steel coated with copper. Loss strongly depends on RRR of coating. As we found lately the current technology provides rather poor RRR =3

TOSHIBA made four couplers



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Feature Story

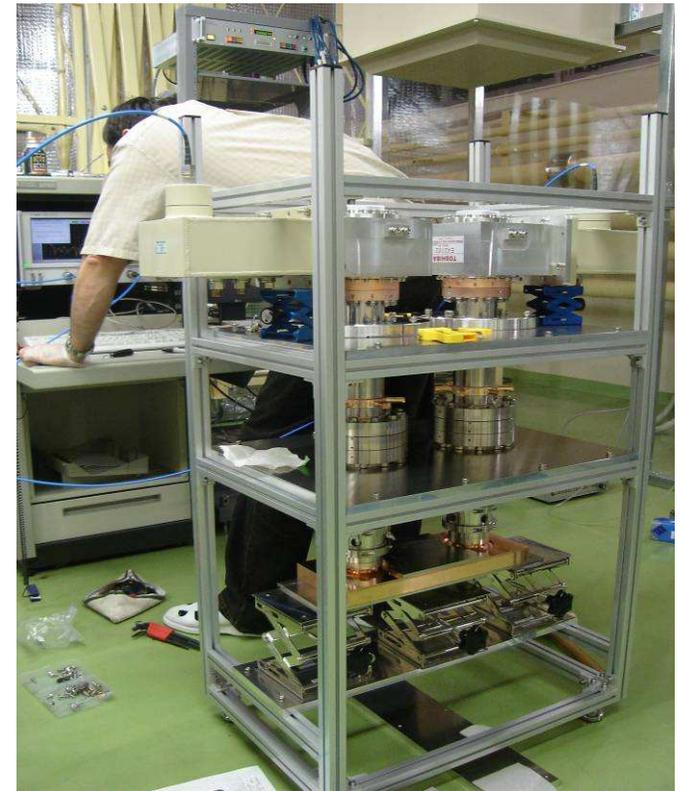
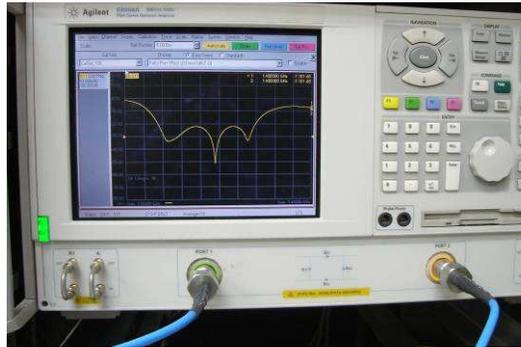
High Performance Coupler



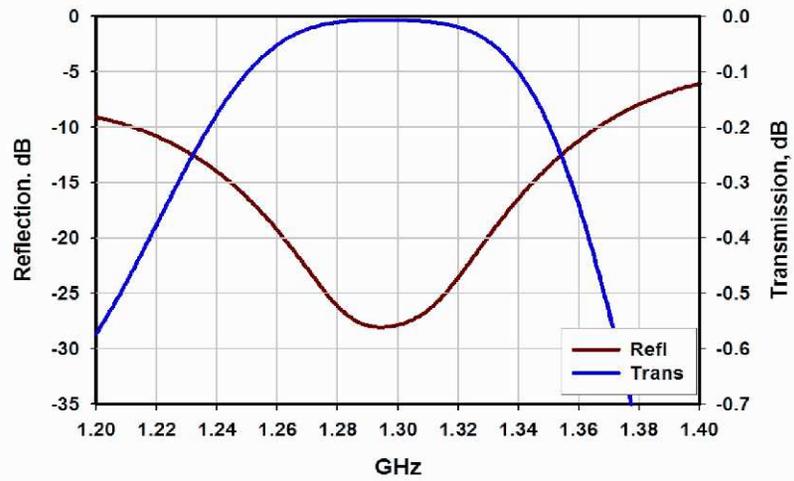
The protection cases for the couplers and three KEK scientists (left to right: Hiroshi Matsumoto, Kenji Saito, and Sergey Kazakov) in the TOSHIBA clean room in Tochigi, Japan.

Cold measurement.

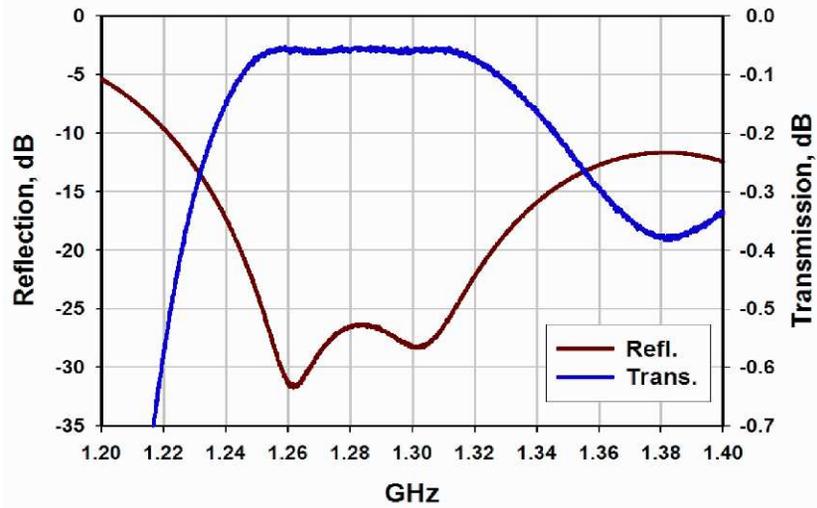
TOSHIBA did mechanical job perfectly! These rather complicated devices was built without any single preliminary RF cold measurements. But we got good SWR instantly just after assembling! It was a big relief.



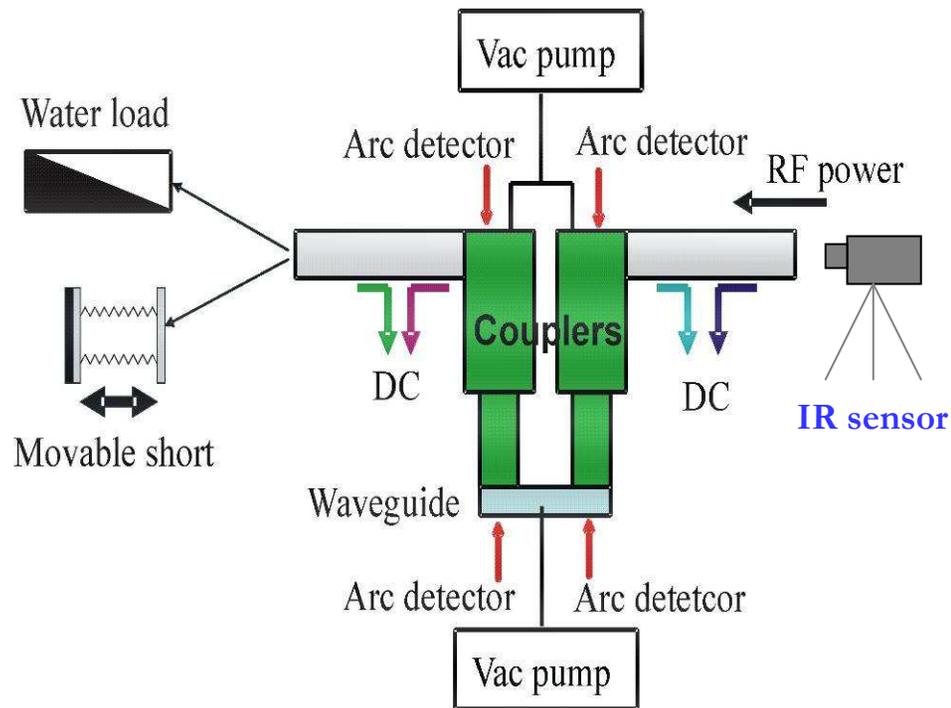
Calculated passband of single coupler



Measured passband



High power test stand

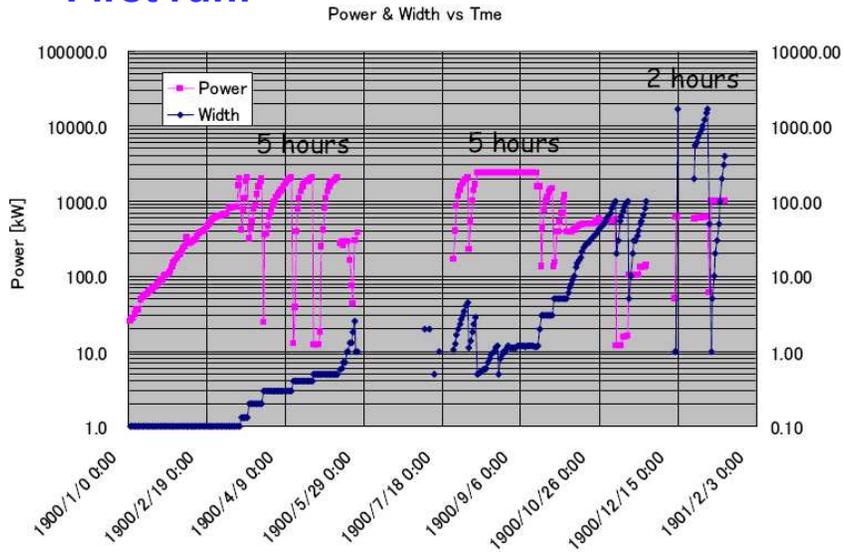


High power test

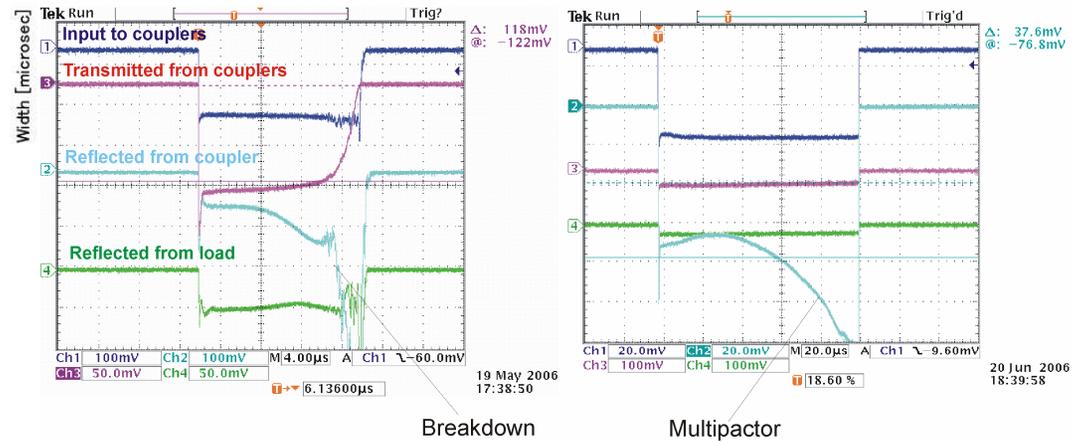


Two run were performed.

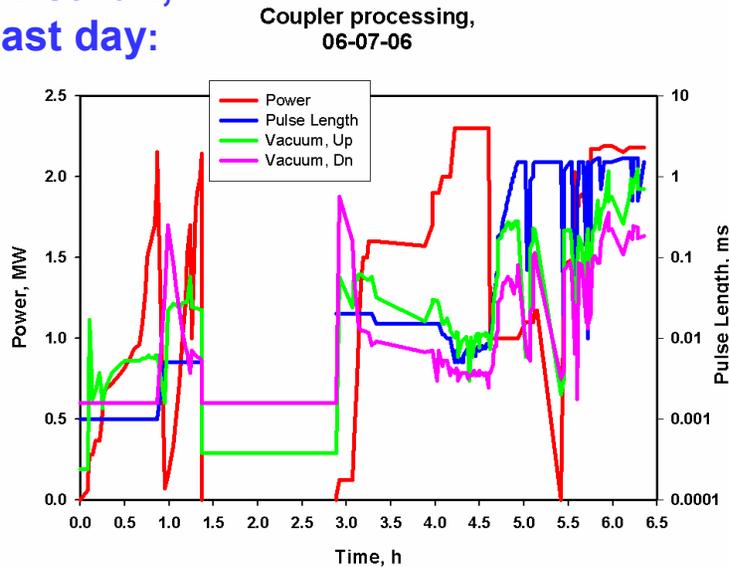
First run:



Typical pulse shapes at beginning of conditioning



First run, last day:

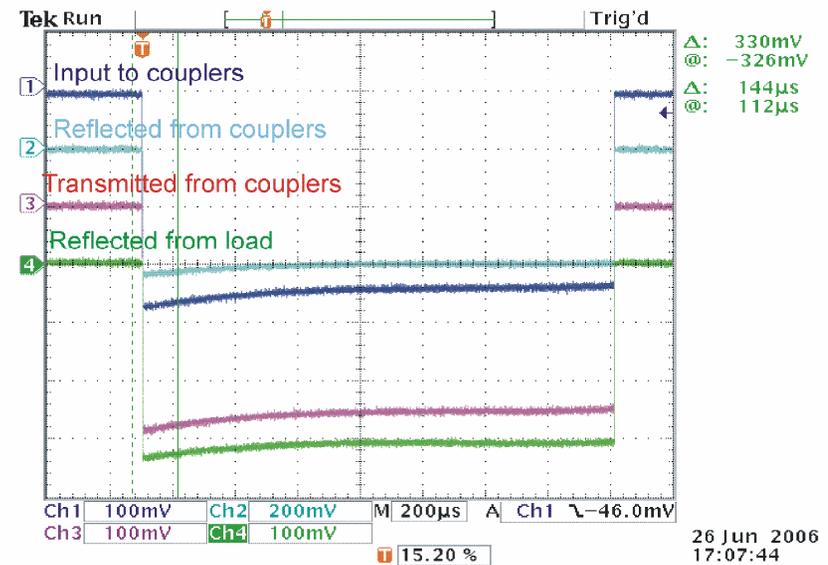
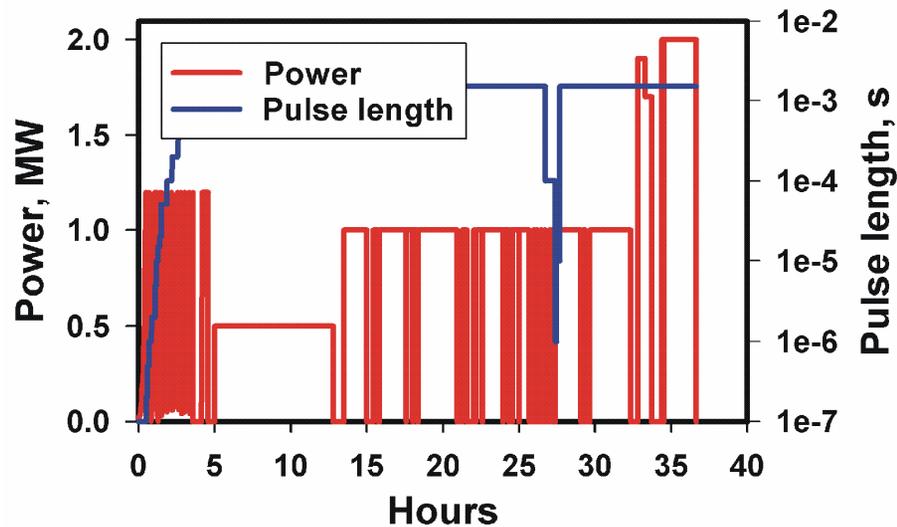


Power was raised up to 2.3MW x 1.7ms x 5pps (~20kW av.)
 At this level warm window was destroyed by thermal stress.

Second run.

Warm window was replaced and couplers were baked.

History of processing and pulse shape after conditioning:



Summary of long run of coupler:

500 kW x 1.5ms x 5Hz

~8 h

1 MW x 1.5 ms x 5Hz

~18 h

2 MW x 1.5 ms x 3Hz

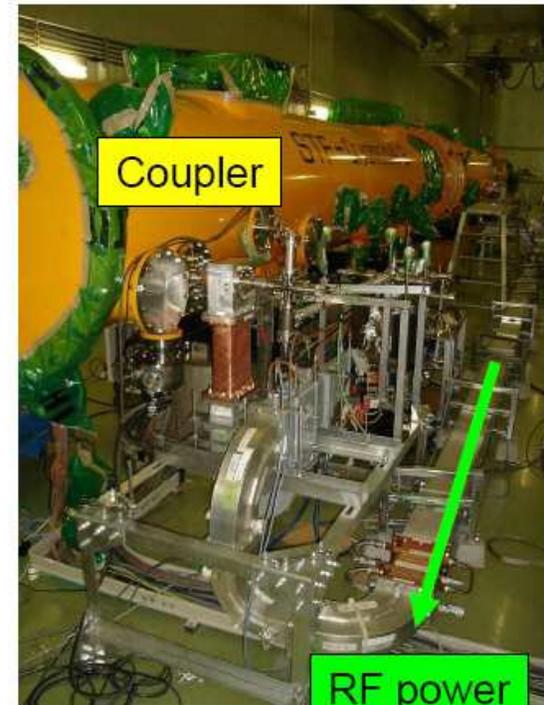
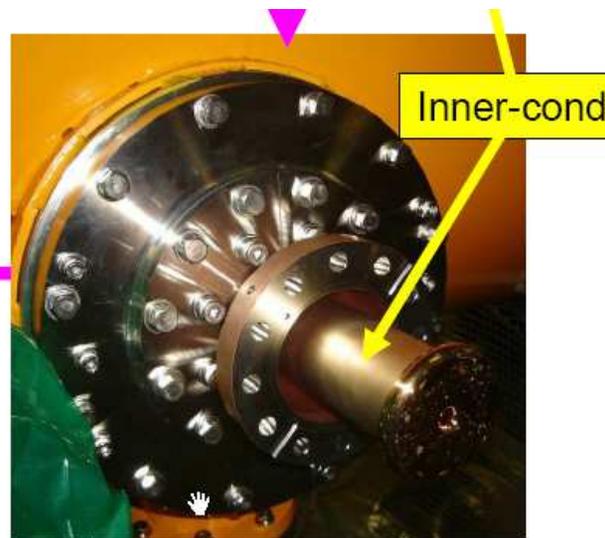
~3.5 h

High power test demonstrated that couplers can successfully operate with pulse $1\text{MW} \times 1.5 \times 5\text{pps}$ and $2\text{MW} \times 1.5\text{ms} \times 3\text{pps}$ for matched load and pulse $500\text{kW} \times 1.5\text{ms} \times 5\text{pps}$ for short.
Effect of multipactor is weak. Upper limit of multipactor is about 200 kW.

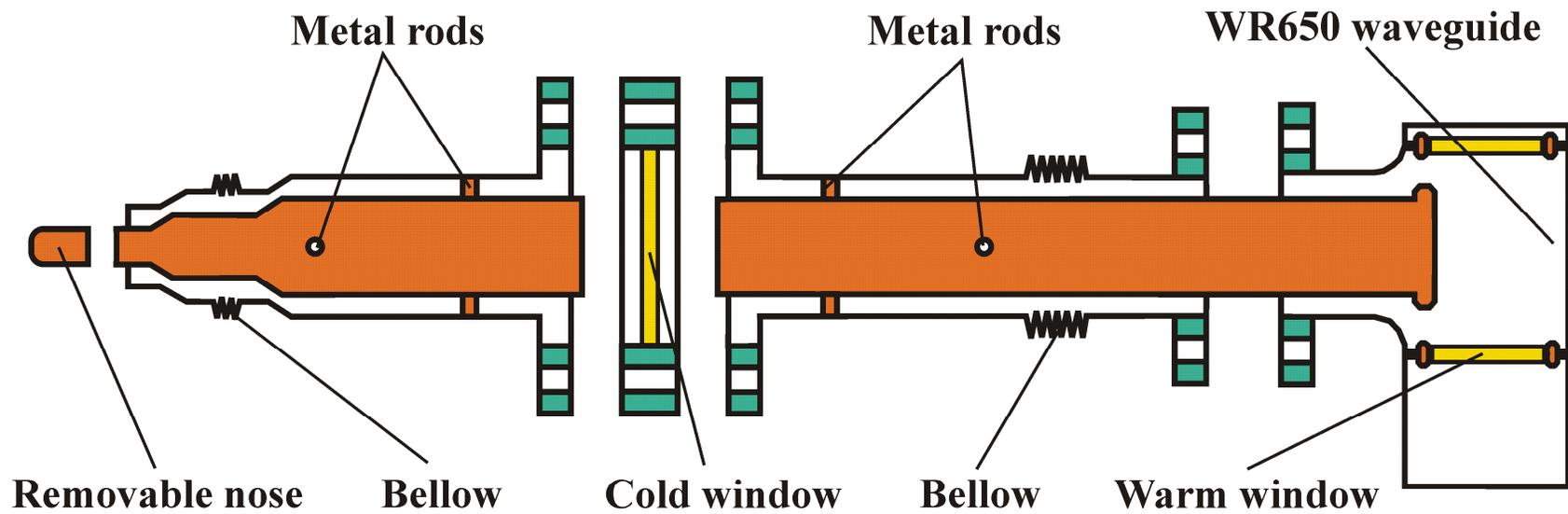
Recently coupler was installed in cryomodule and processed.



Assembly of Cold side of CC coupler



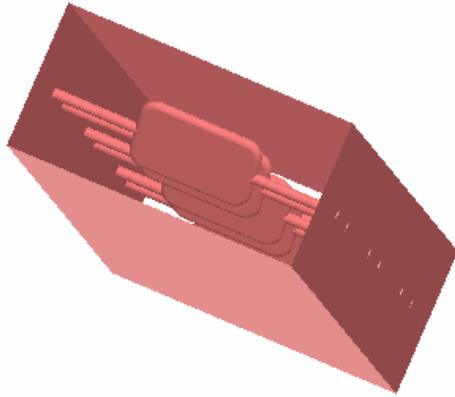
Processing of assembled CC coupler was done at room temperature.
RF power = 250 kW
Pulse width = 1.5 ms
Repetition rate = 5Hz
was achieved within the processing time of 31 hours



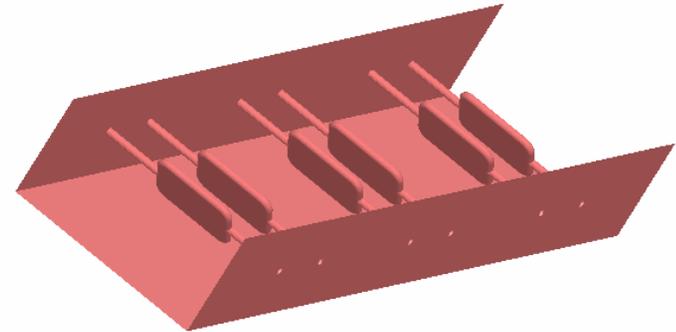
**Coupler can be modified to have changeable coupling.
 Two bellows allow to move middle part of coupler with antenna.**

Other designs for L-band

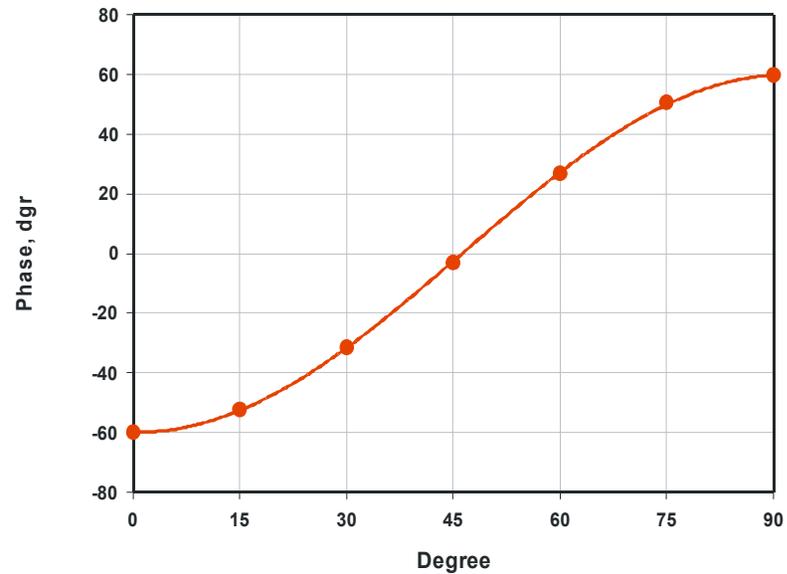
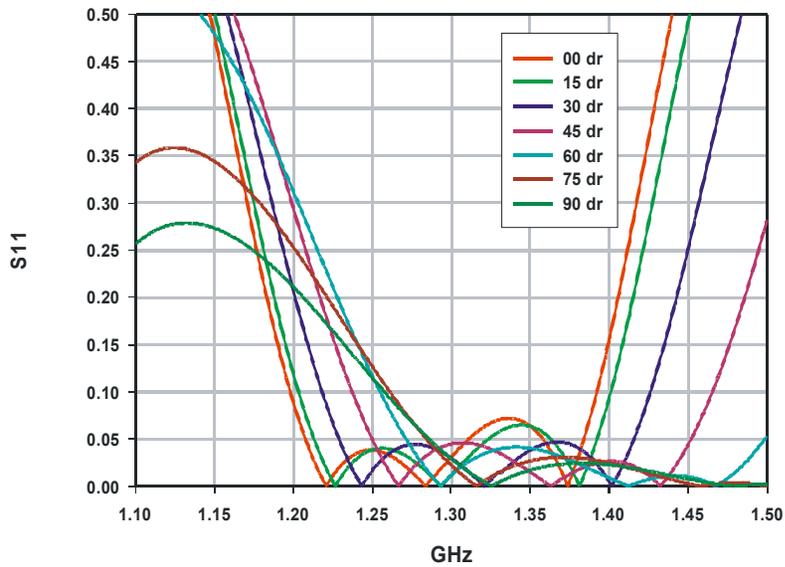
Mechanically controlled phase shifter



Passband
8mm-44mm



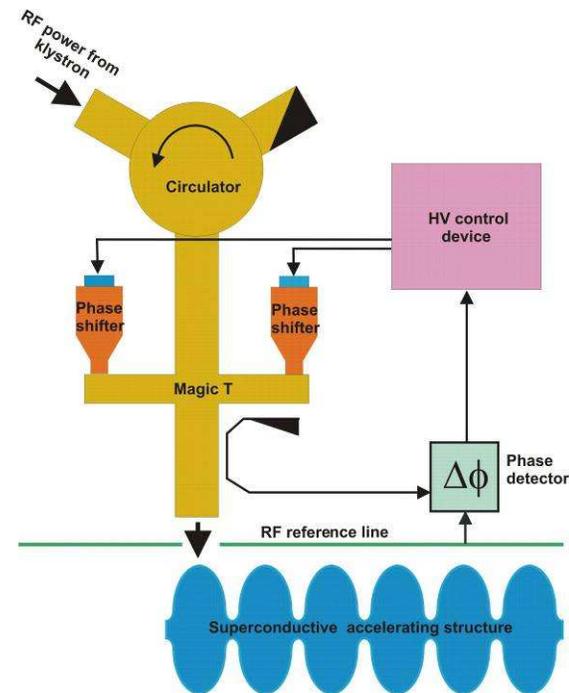
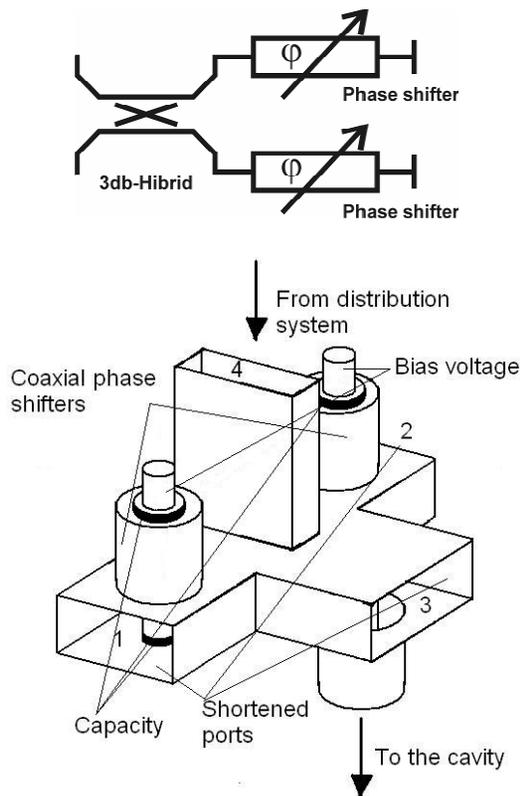
Phase vs plate angle,
plate 8mm-44mm



Phase shifter was built and work as predicted. Expected power limit in air is 2 MW

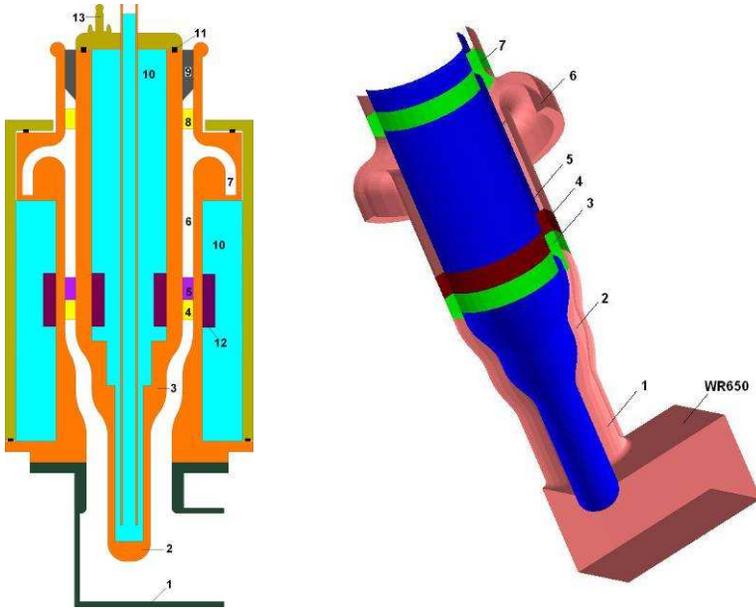
Ferroelectric phase shifter (tuner)*

Interesting project is under developing in the Omega-P/Yale Beam Lab. We trying to build ultra-fast ($\sim 1\mu\text{s}$) electrically controlled phase shifter/tuner using ferroelectric ceramic. It can be used for fast control of coupling of accelerating structures (phase and/or amplitude). It can be used as fast external tuner to control resonance frequency of cavity as well.



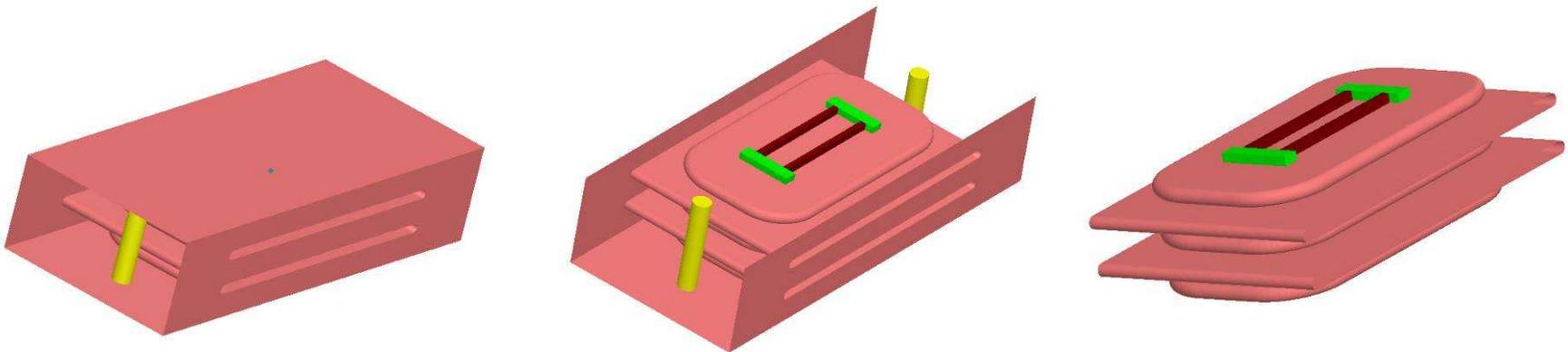
*S.Yu. Kazakov, V.P. Yakovlev, J.L. Hirshfield, "Fast Ferroelectric Tuner for L-band", 27th WG5-Asia Meeting, KEK, Japan, November 2005.

Work supported by US DoE, OHEP.



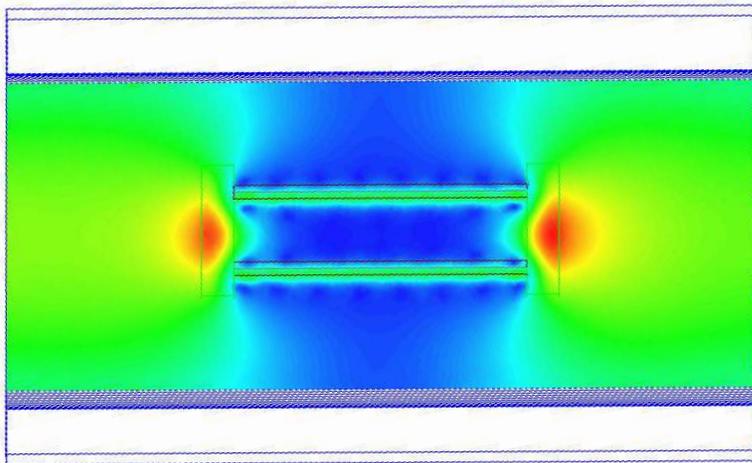
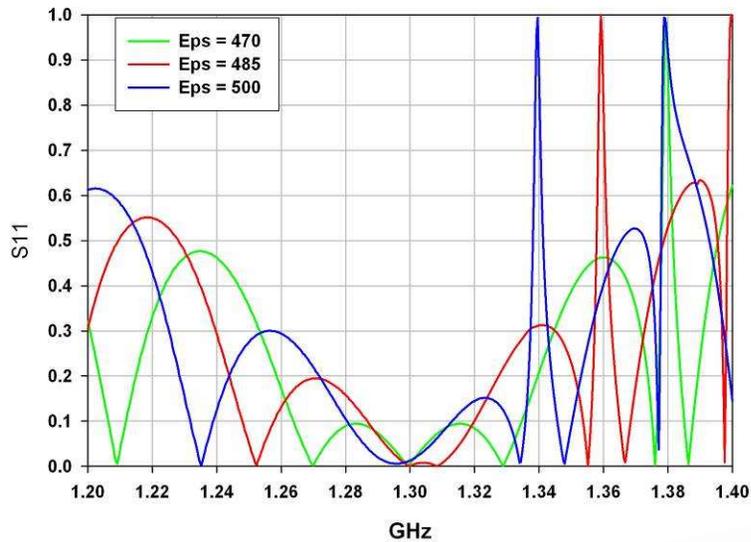
Coaxial design of phase shifter (shorted)

dielectric constant, ϵ	~500
tunability, $\partial\epsilon/\partial E_{bias}$ (E_{bias} is the bias field)	$> 2/(kV/cm)$
response time	< 10 ns
loss tangent at 1.3 GHz, $tg\delta$ (preliminary measurements)	$\sim 10^{-3}$
Breakdown limit	200 kV/cm
thermal conductivity, K	7.02 W/m-°K
specific heat, C	0.605 kJ/kg-°K
density, ρ	4.86 g/cm ³
coefficient of thermal expansion	$10.1 \times 10^{-6} / ^\circ K$
temperature tolerance, $\partial\epsilon/\partial T$	3 /°K



Waveguide design of phase shifter

Waveguide tuner 1.3 GHz
Passband

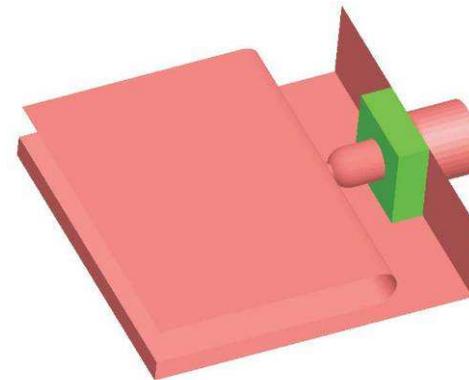


Parameters of waveguide phase shifter
(500 kW x 1.3ms x 5pps – ILC parameters)

Max. E-field in ferroelectric 3.0 kV/cm
 Avrg. E-field in ferroelectric 2.0 kV/cm
 Max. E-field in ceramic 5.9 kV/cm
 Max. E-field in air 6.1 kV/cm
 One way phase shift ~ 115 deg (15 kV/cm bias)

One way loss 6.0 %
 Ferroelectric pulse heating 0.2 K (d_eps = 0.6)
 Ferroelectric average heating 0.9 K (d_eps = 2.7)

(Loss and heating were calculating for
 5e-4 ferroelectric loss tangent and for
 1e-4 other ceramics loss tangent)



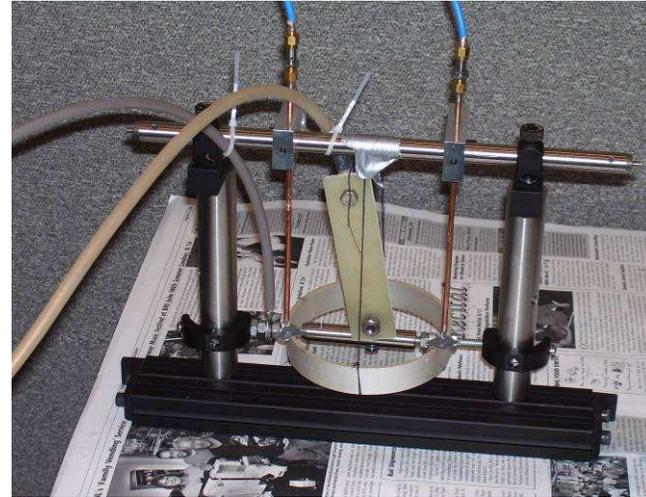
HV input

The procedure of the phase shifter tuning is developed in order to prevent EM field spread (and thus, the field enhancement) in different ferroelectric slabs.

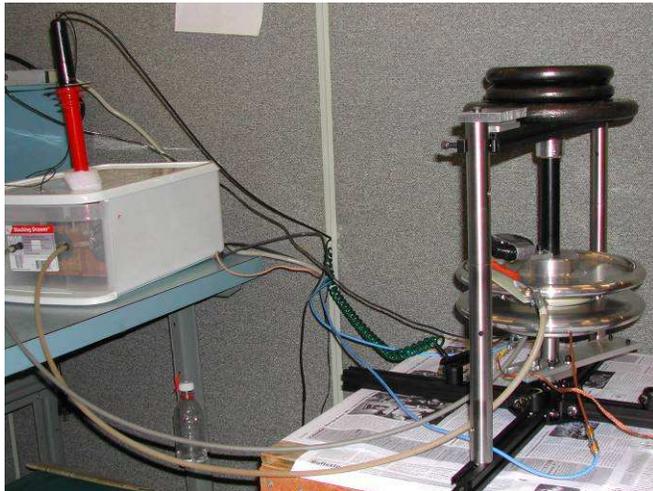
Measurements of properties of ferroelectric ceramic



Ferroelectric ring

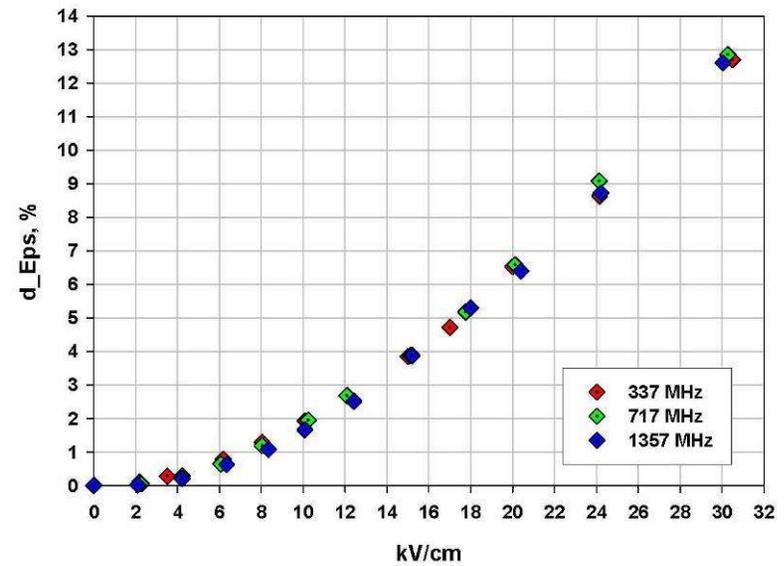


Tools for tunability measurements in the oil.



Two-disk resonator

Ring 1.075 (N 4)



Thank you for your attention!