

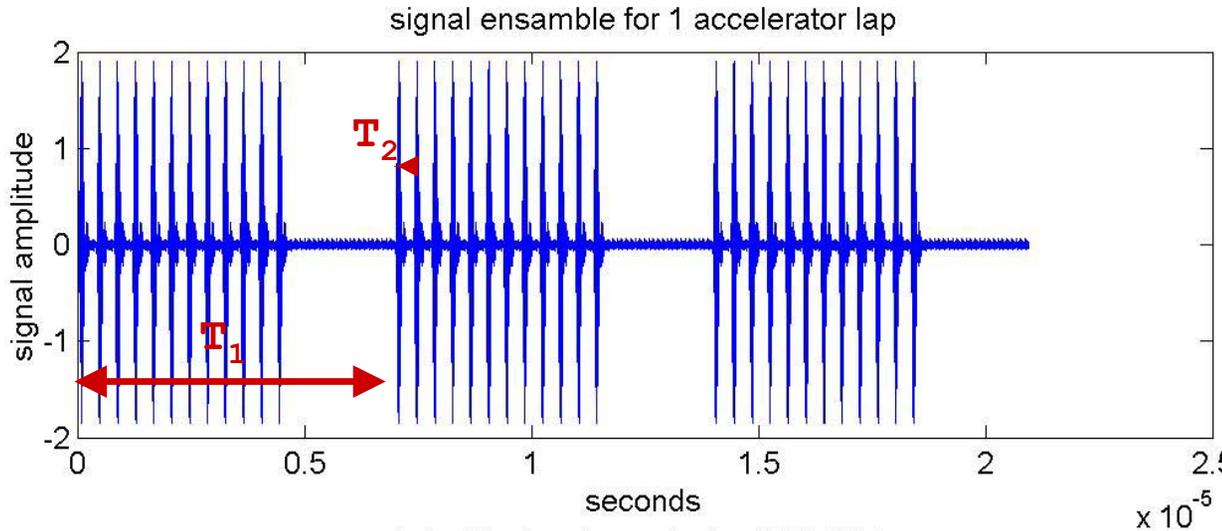
# BPM test signal Spectrum analysis

BPM project

# Gustavo's test signal

- No RF buckets = 1113.
- RF frequency = 53.104 MHz.
- 1 lap period = 20.958  $\mu$ s.
- Starting from Jim Steimel's A and B sampled at 2GHz a closed-orbit load of 36 bunches in 3 batches of 12 bunches separated by abort gaps is generated. i.e. 41917 samples.
- This signal is resampled at a freq. close to 7/5 of 53.104MHz.
- The difference between my sampling freq and 7/5 of 53.104MHz was 0.39%.
- This error shifts the spectrum centered at 53.104MHz by 207.7KHz.
- The "recicler" filter BW is ~10KHz, so it was very sensitive to this error.

# Spectrum analysis

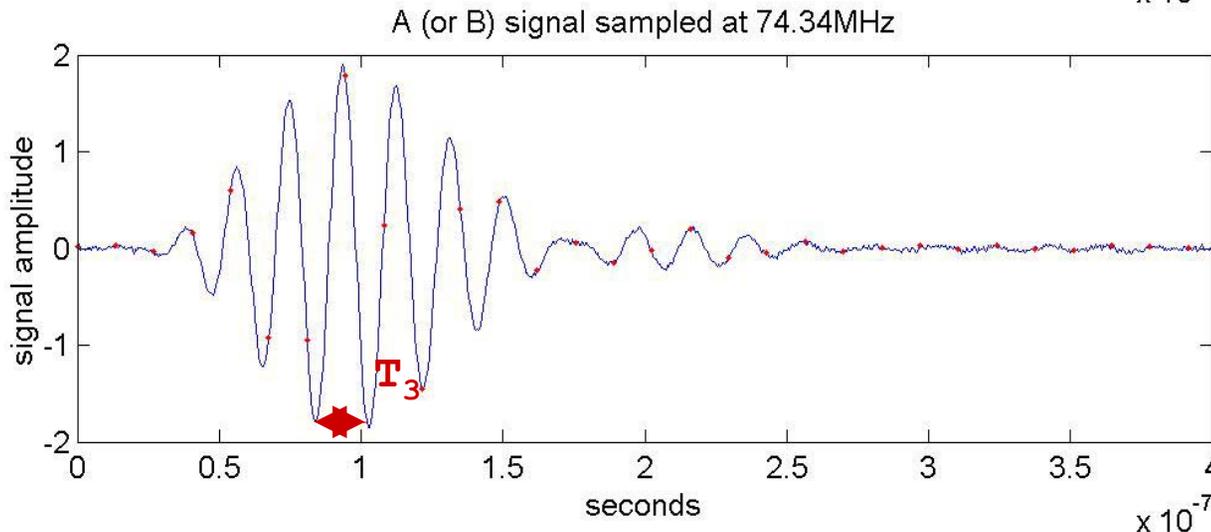


$$T_1 = 6.99 \mu\text{s} \Rightarrow f_1 = 143 \text{KHz}$$

$$T_2 = 396 \text{ns} \Rightarrow f_2 = 2.52 \text{MHz}$$

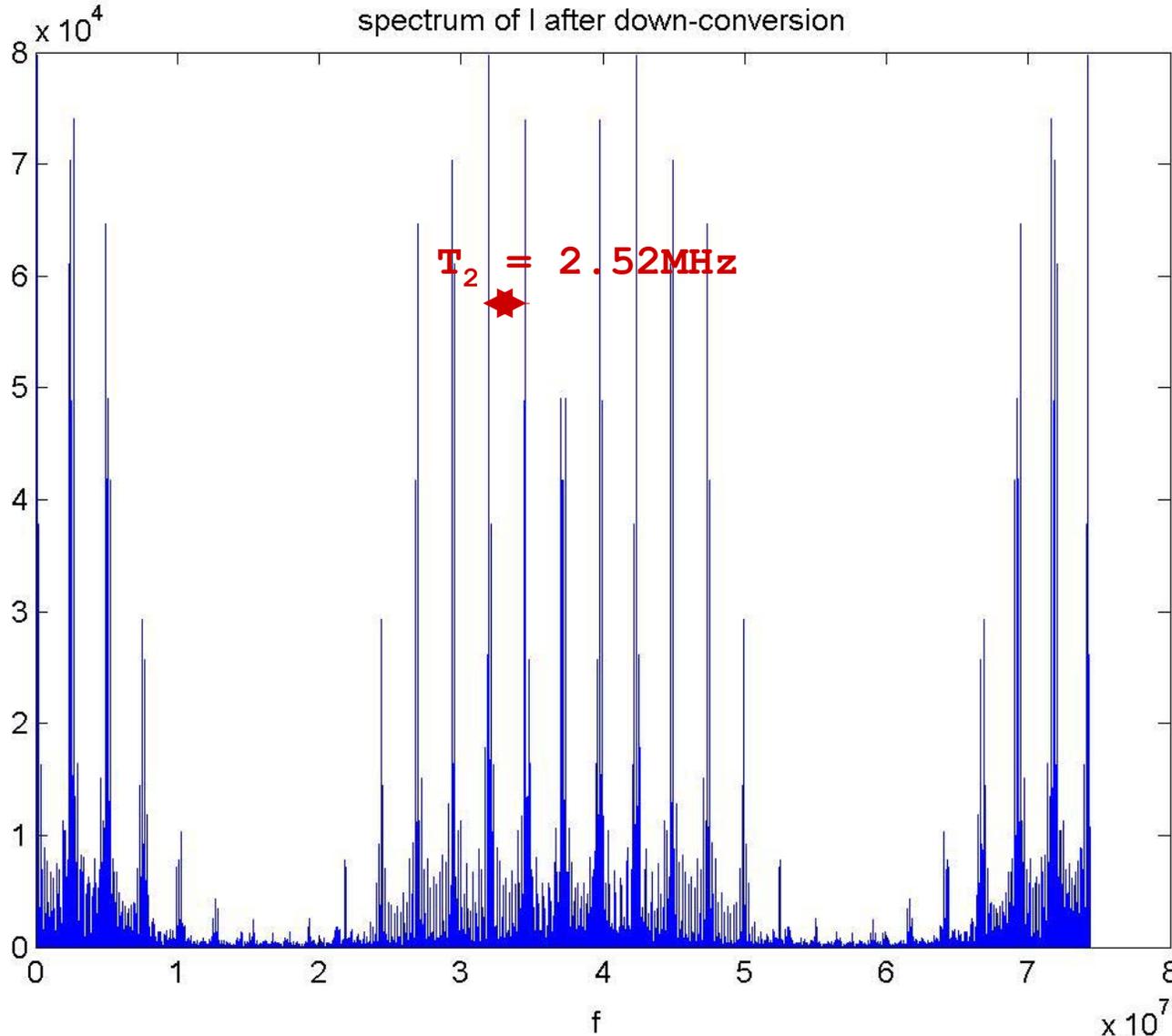
$$T_3 = 18.9 \text{ns} \Rightarrow f_3 = 53.1 \text{MHz}$$

The 53MHz signal is not periodic  $f_3$  represents its 1<sup>st</sup> harmonic.



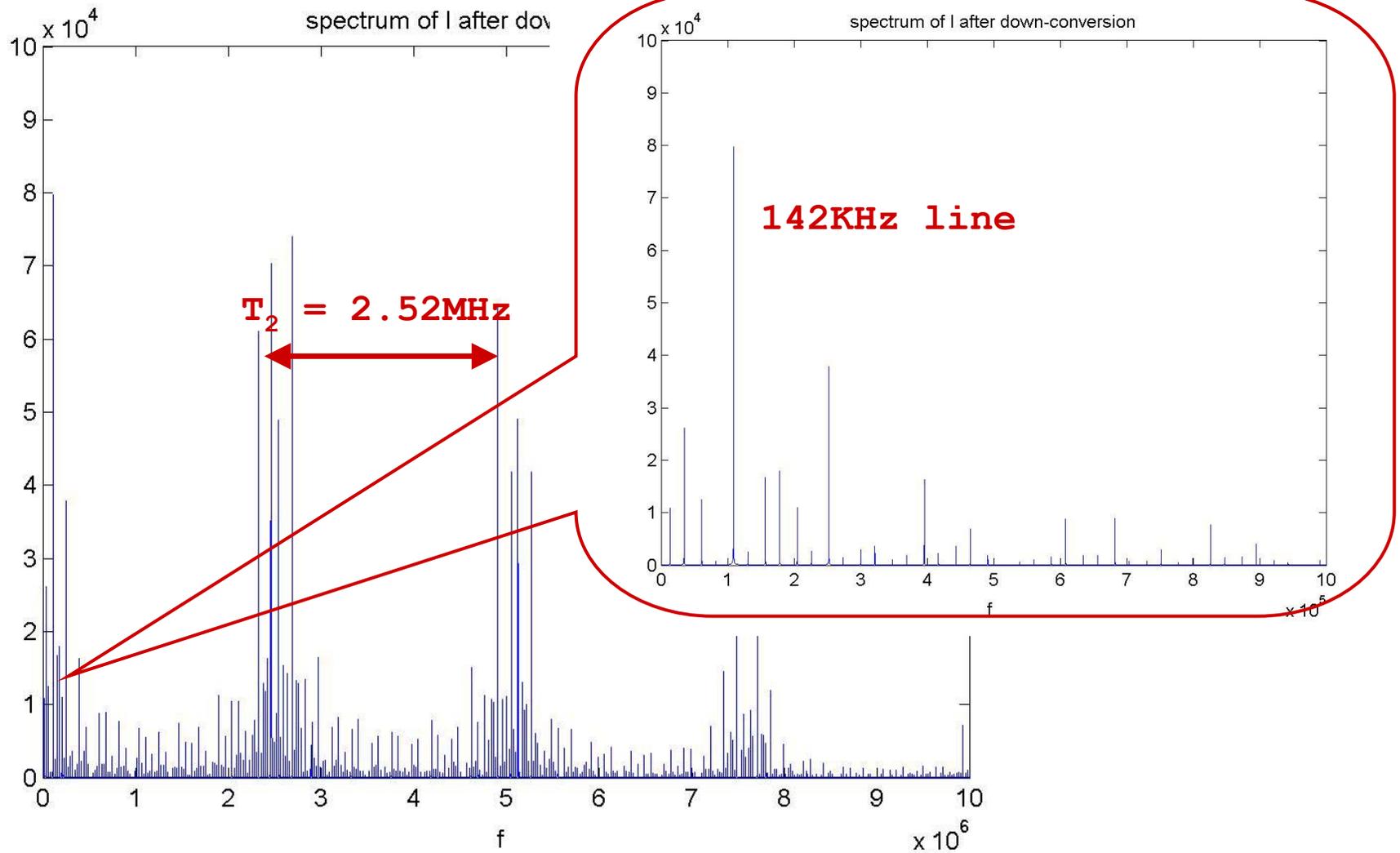
Sampling frequency  
 $f_s = 74.3 \text{MHz}$

# Spectrum analysis

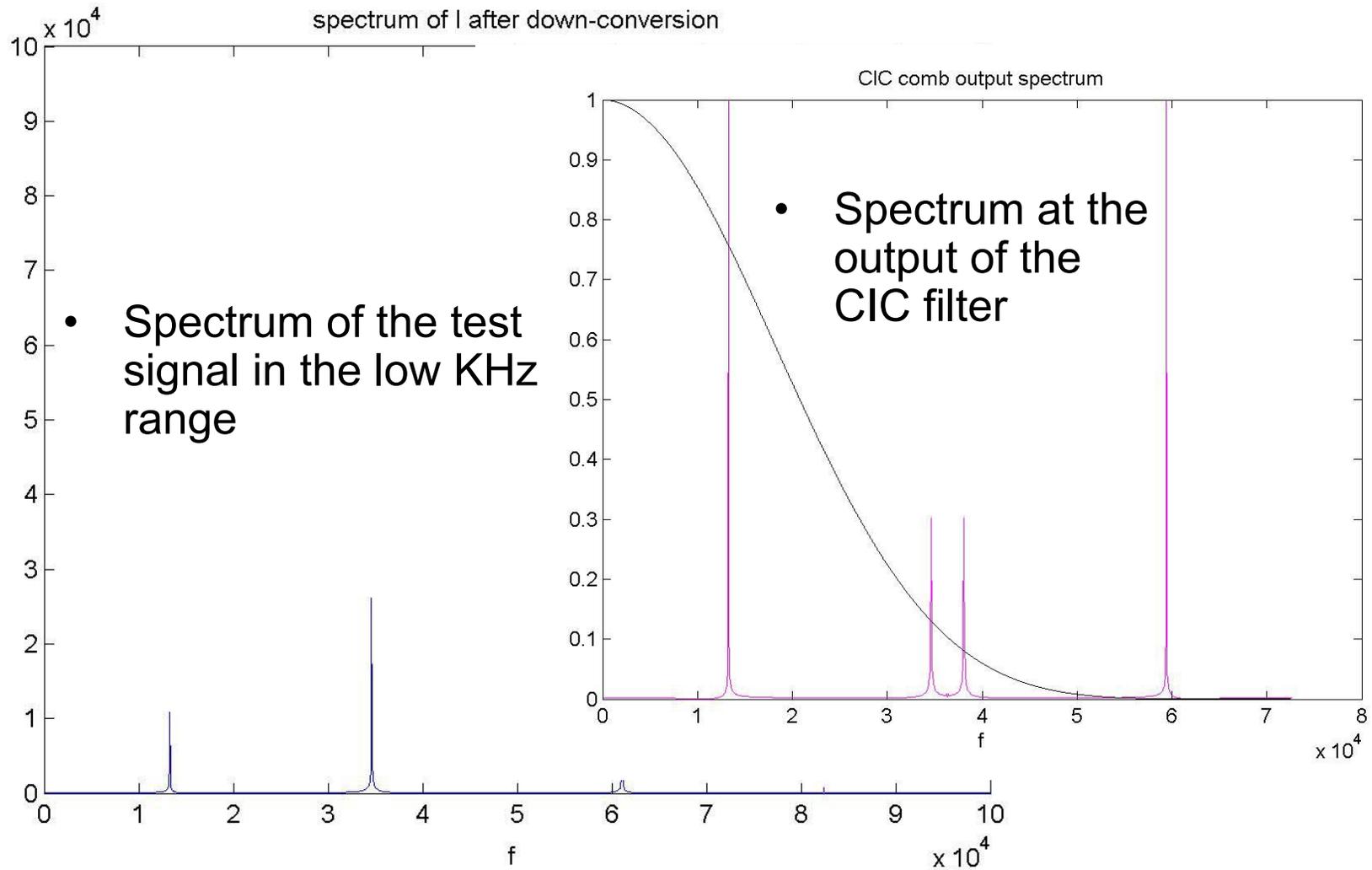


The spectrum of the input (sampled) signal is centered at 53MHz. After down-conversion a portion is sent to baseband. Most of the spectrum density concentrates around 2.52MHz lines.

# Spectrum analysis



# Spectrum analysis



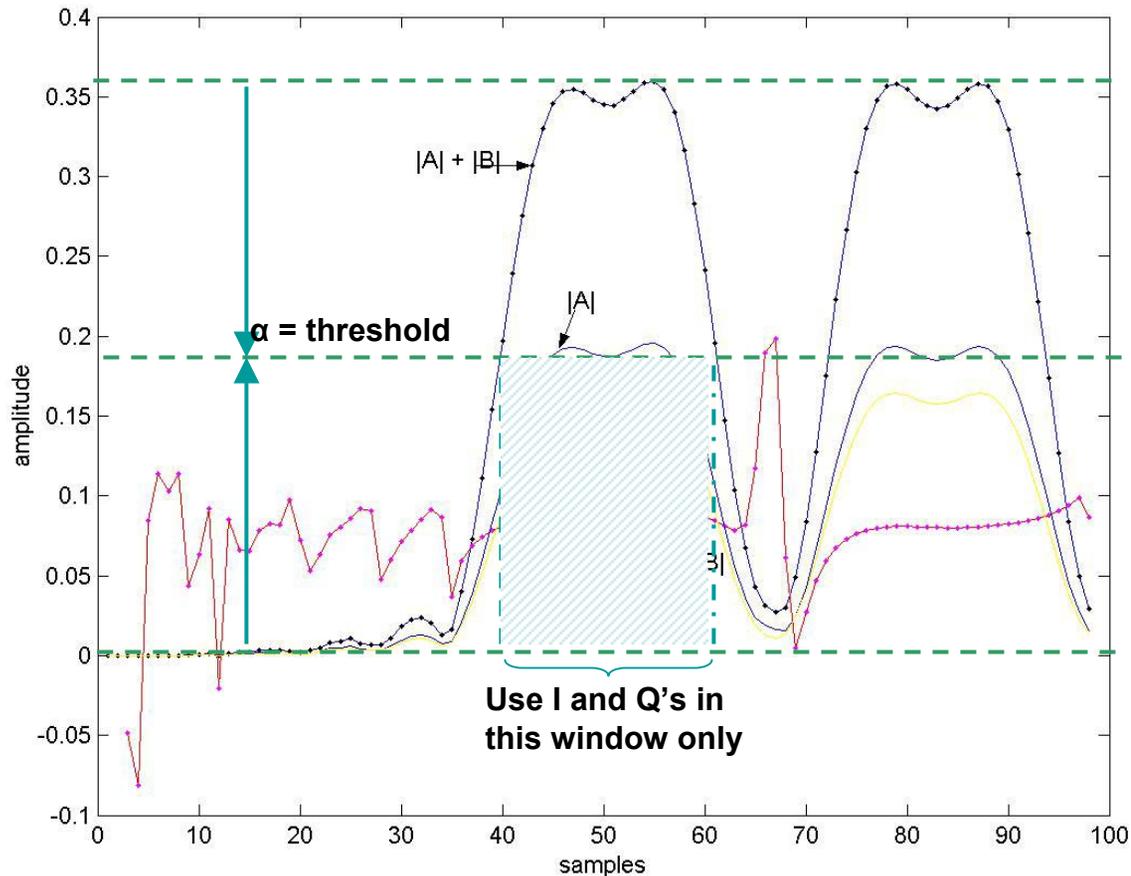
# Specifications to implement the “batch envelope filter” in the Ecotek Stratix FPGA

BPM project

# Signal to noise

- The matched filter is a linear filter widely used to recover deterministic signals embedded in white Gaussian noise (WGN) because it optimizes the S/N ratio.
  - $y(n) = x(n) * s(n)$ , where  $s(n)$  is the deterministic signal and  $x(n)$  is the noisy signal. i.e.  $x(n) = s(n) + w(n)$ . ( $w(n)$  is WGN).
  - $S/N = \mathcal{E}/\sigma^2$ , where  $\mathcal{E}$  is the energy of the signal and  $\sigma^2$  is the noise variance.
  - S/N increases with the number of “signal samples”.
  - The matched filter meets the Cramer-Rao lower bound.
- We can do better than the matched filter by choosing only “signal samples”. In a batch we have enough “signal samples” that can be detected applying a simple threshold cut.

# Filtered signals



$|A|-|B|/|A|+|B|$  is fairly constant for  $|A|+|B|>\text{Threshold}$ .

The example shows about 25 useful points per batch.

The samples are averaged to provide a single I and Q pair per batch.

Batch numbers are averaged again to improve estimate and lower the data bandwidth.

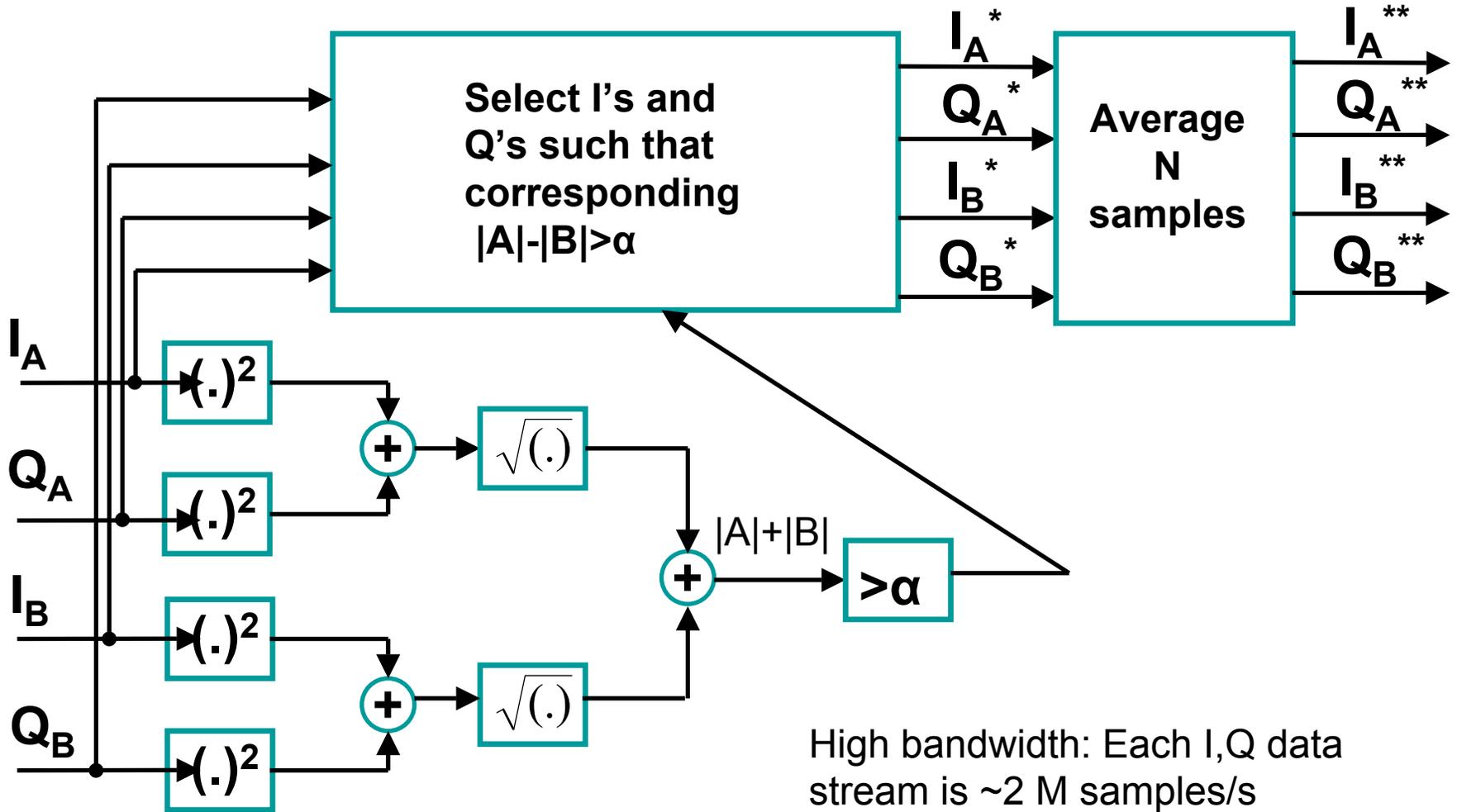
# Task Description

- Implement the “batch envelope filter” in the Stratix FPGA that handles the I and Q data stream into the Ecotek’s FIFO.
- Write the Filter algorithm in VHDL code and simulate it on the PC using I and Q signals coming from Matlab simulations.
- Port the Filter VHDL to the Ecotek FPGA.
- Test and qualify the Filter using Jim Steimel’s setup.

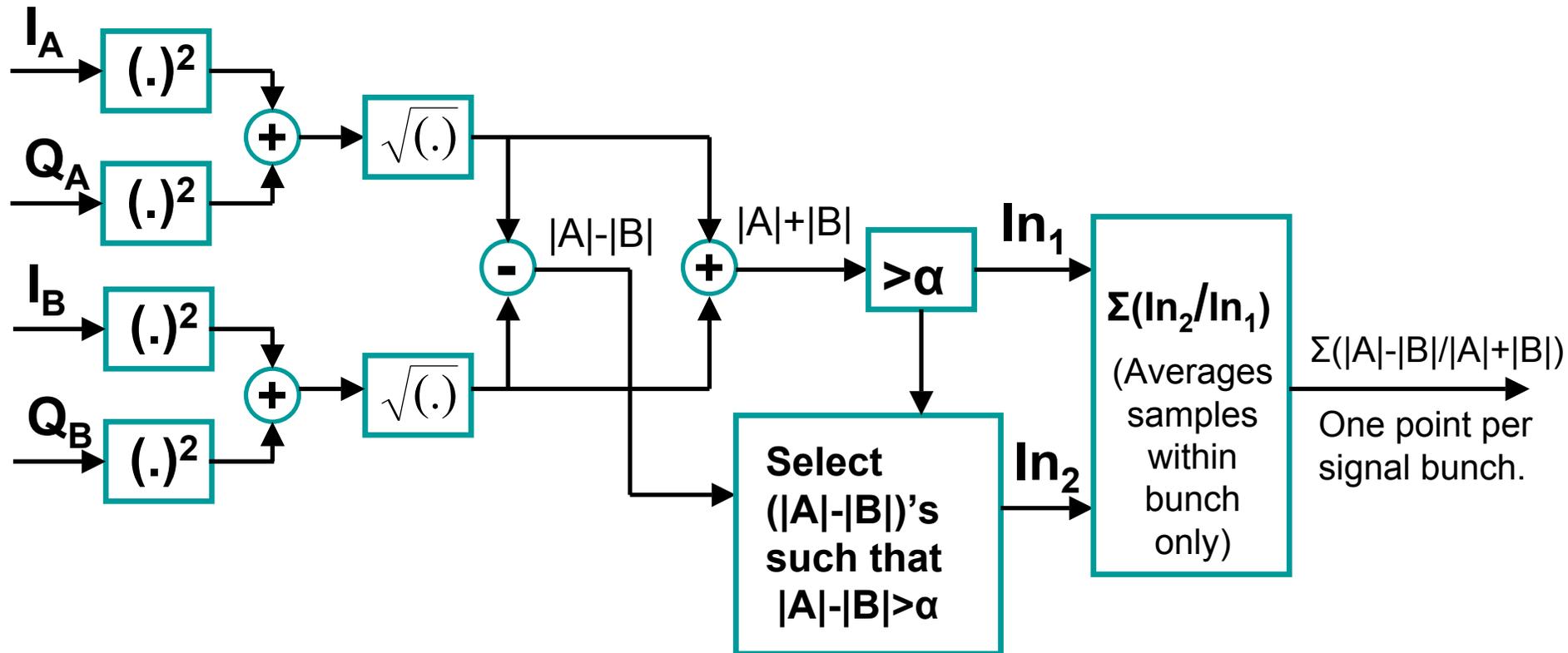
# Signal processing options

- The amount of signal processing done by the filter has alternatives:
  - i. Select I and Q based on corresponding  $|A|+|B|>\text{threshold}$  but output “raw” I and Q to VME.
  - ii. Select I and Q as above but output averaged I and Q’s to VME.
  - iii. Select I and Q as above, compute  $|A|-|B|/|A|+|B|$  for each sample above threshold.
  - iv. Select I and Q as above, compute  $|A|-|B|/|A|+|B|$  for each sample above threshold, sum up N number of points to eliminate betatron oscillations ( $N\sim 32$ ).
- How to determine the threshold?
  - The threshold can be fixed to a number above the noise level and below  $\frac{1}{2}$  the minimum  $|A|+|B|$  signal level expected.
  - It can be calculated from the signal level and set accordingly.

# Filter Algorithm (Option i. and ii.)

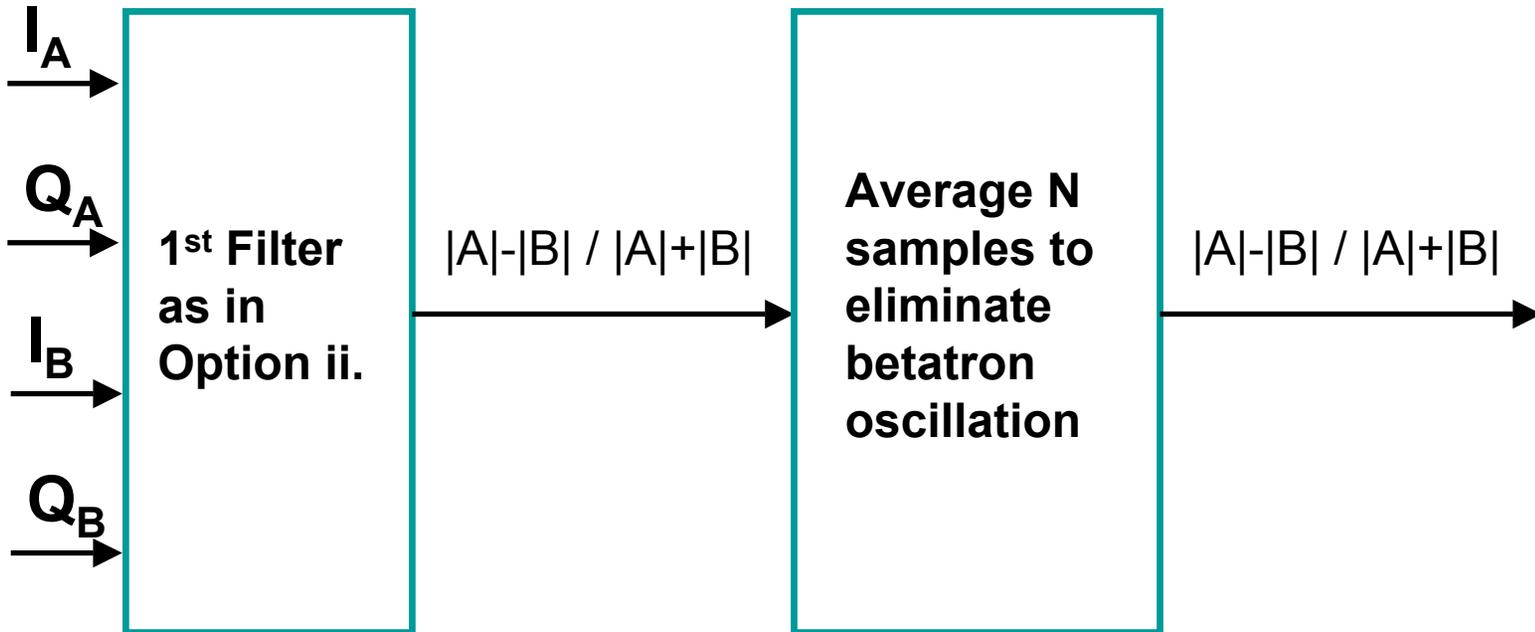


# Filter Algorithm (Option iii.)



Moderate bandwidth: 145 K samples/s

# Filter Algorithm (Option iv.)



Low bandwidth: 4.5 K samples/s

# Position estimation

Let  $\mathbf{s}_I(n) = I(n) + w(n)$  and  $\mathbf{s}_Q(n) = Q(n) + w(n)$  If the noise is WGN  $\sim N(0, \sigma^2)$ ,

the likelihood estimation function of  $\mathbf{I}$  is:  $L(\mathbf{I}, \hat{\mathbf{I}}) = \frac{1}{(2\pi\sigma^2)^{N/2}} \exp\left[-\frac{1}{2\sigma^2} \sum_{n=0}^{N-1} (\mathbf{x}(n) - \hat{\mathbf{I}}(n))^2\right]$

Similarly for  $\mathbf{Q}$ .

However, the position estimation is nonlinear with respect to  $\mathbf{I}$  and  $\mathbf{Q}$ .

$$P = \frac{|\hat{A}| - |\hat{B}|}{|\hat{A}| + |\hat{B}|} = \frac{\sqrt{\hat{I}_A^2 + \hat{Q}_A^2} - \sqrt{\hat{I}_B^2 + \hat{Q}_B^2}}{\sqrt{\hat{I}_A^2 + \hat{Q}_A^2} + \sqrt{\hat{I}_B^2 + \hat{Q}_B^2}}$$

It is better to run sums over  $\mathbf{I}$  and  $\mathbf{Q}$  before calculating  $P$ .

The filters color the noise. Not WGN any more.

# How to set the threshold

- The threshold can be fixed to a number above the noise level and below  $\frac{1}{2}$  the minimum  $|A|+|B|$  signal level expected.
- It can be calculated from the signal level and set accordingly.