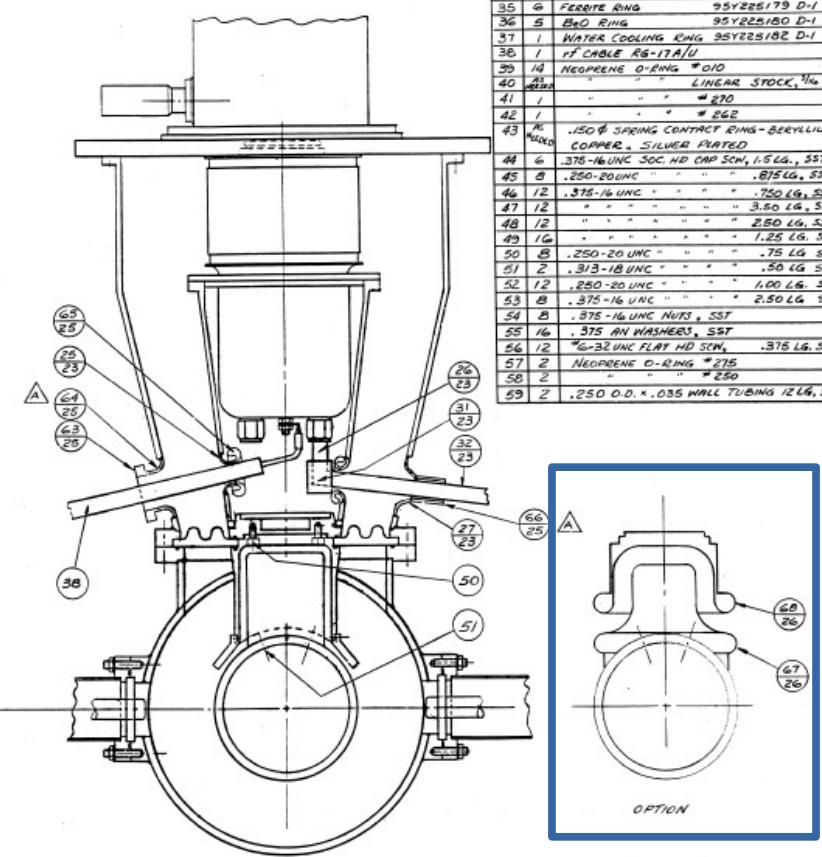
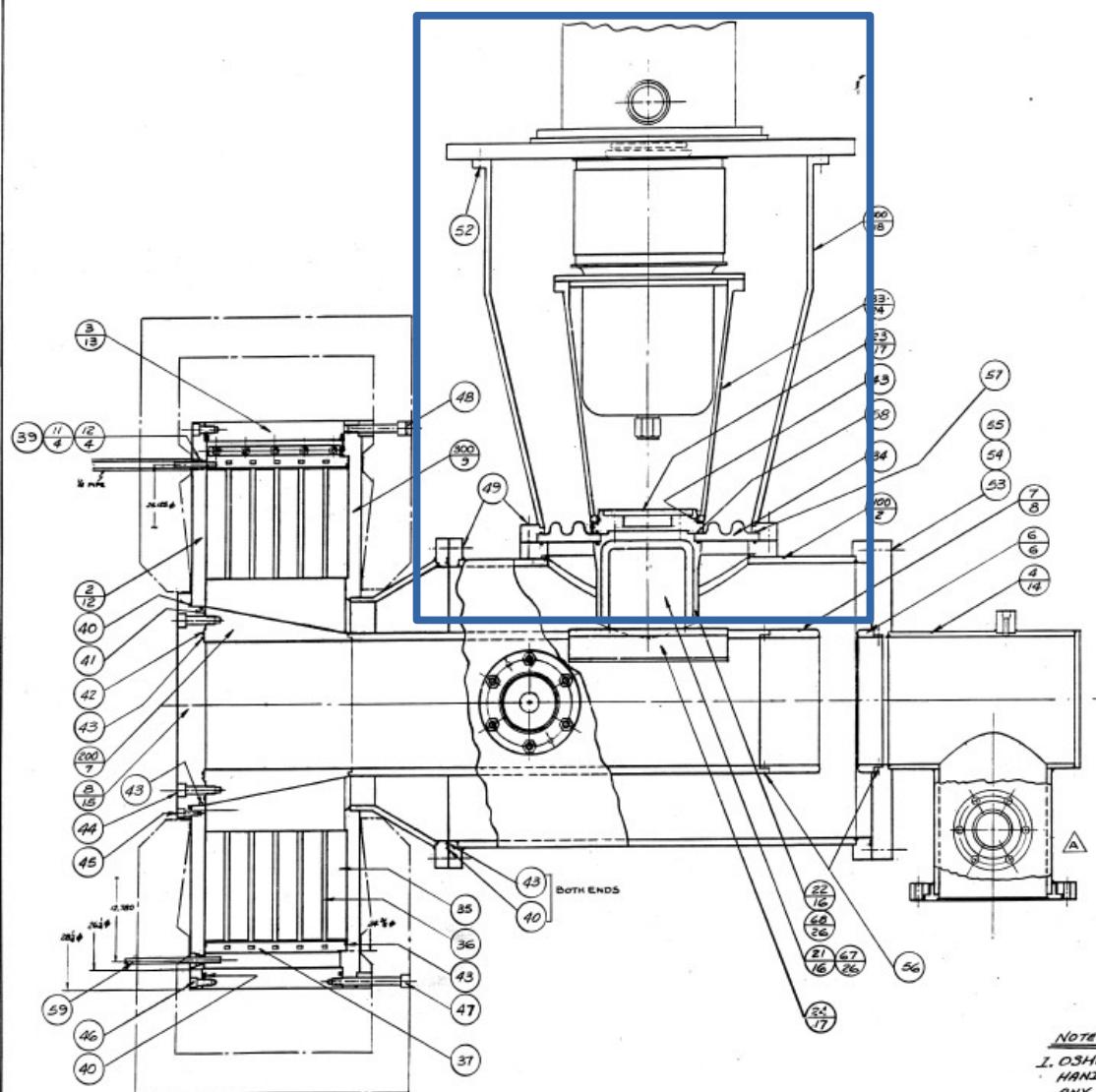


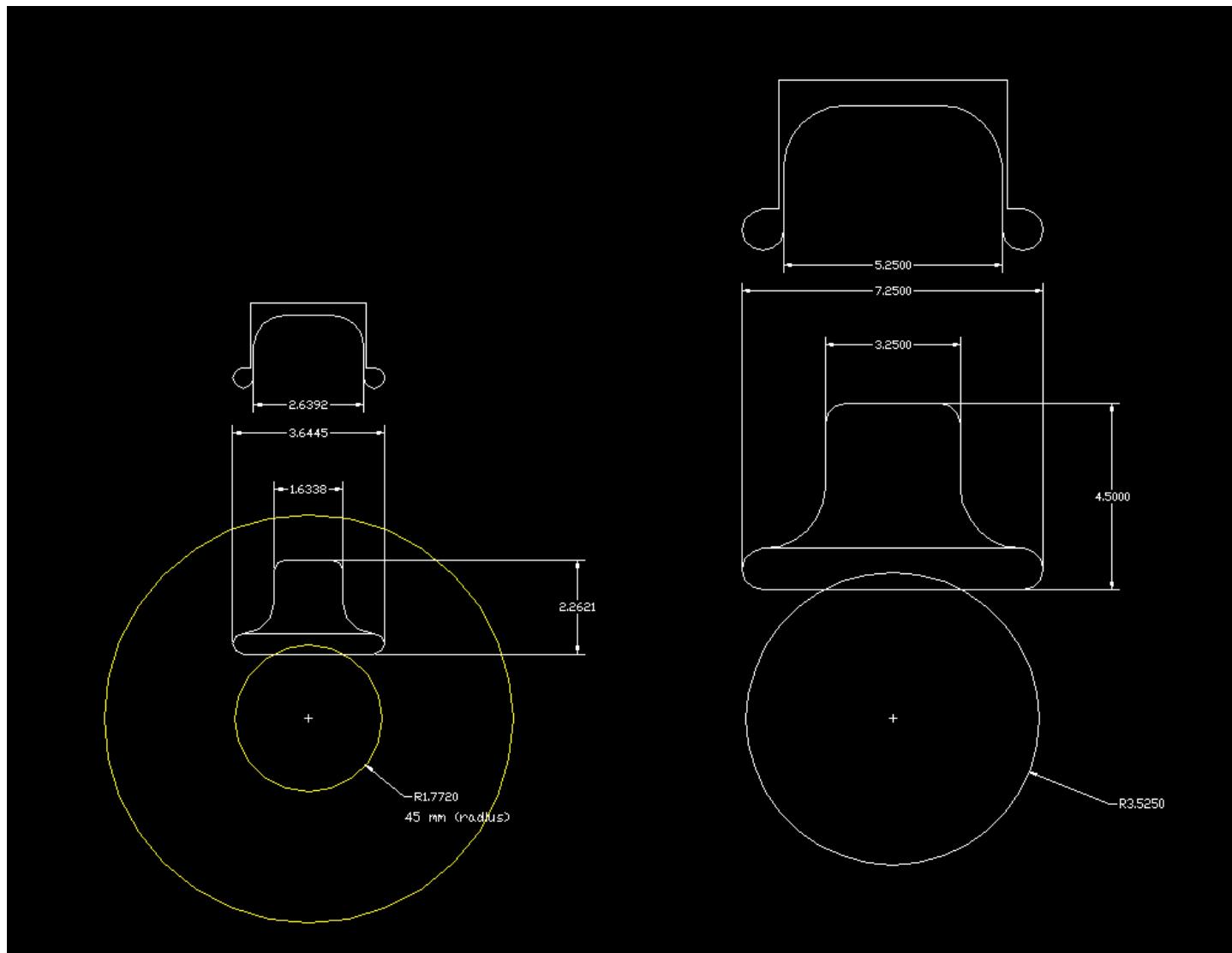
# Coupler

C.Y. Tan  
11 June 2015



| REF ID   | CLASSIFICATION OF DRAWING   | DRAWING NO. 95Y225182 D-1 |             |                  |
|--|---|---------------------------|-------------|------------------|
|  |   | DATE ISSUED               | REVISION    | APPROVED         |
| B  | TOTAL SWIS WAS 26 - ADDED DETAILS 67 & 68 & OPTION VIEW   | 7/1/85                    | LCN         |                  |
| A  | ADDED VACUUM POET ON DETAIL 4<br>REMOVED DETAILS 28, 29, 430 - ADDED DETAILS 69, 64, 65, 66 - TOTAL SWIS ARE 26 | 7/1/85                    | LCN         |                  |
| -  | ORIGINAL ISSUE  | -                         | -           | -                |
| REVISION LETTER  | CLAS. CONTROL   | REVISION                  | DATE ISSUED | CHECKED APPROVED |
| Los Alamos National Laboratory<br>Los Alamos, New Mexico 87545 |   |                           |             |                  |
| PERPENDICULAR BIASED<br>RF CAVITY COLD MODEL                   |   |                           |             |                  |
| LAMPF II   |   |                           |             |                  |
| TOLERANCE - UNLESS OTHERWISE NOTED                             |   |                           |             |                  |
| X-1 O.K. + ANGLES - 1  |   |                           |             |                  |
| O.K. + 1 O.K. + FINISH -                                       |   |                           |             |                  |
| SCALE DRAWING NO. 26 95Y225182 D-1                             |   |                           |             |                  |
| WEST NORTH MIL.  |   |                           |             |                  |

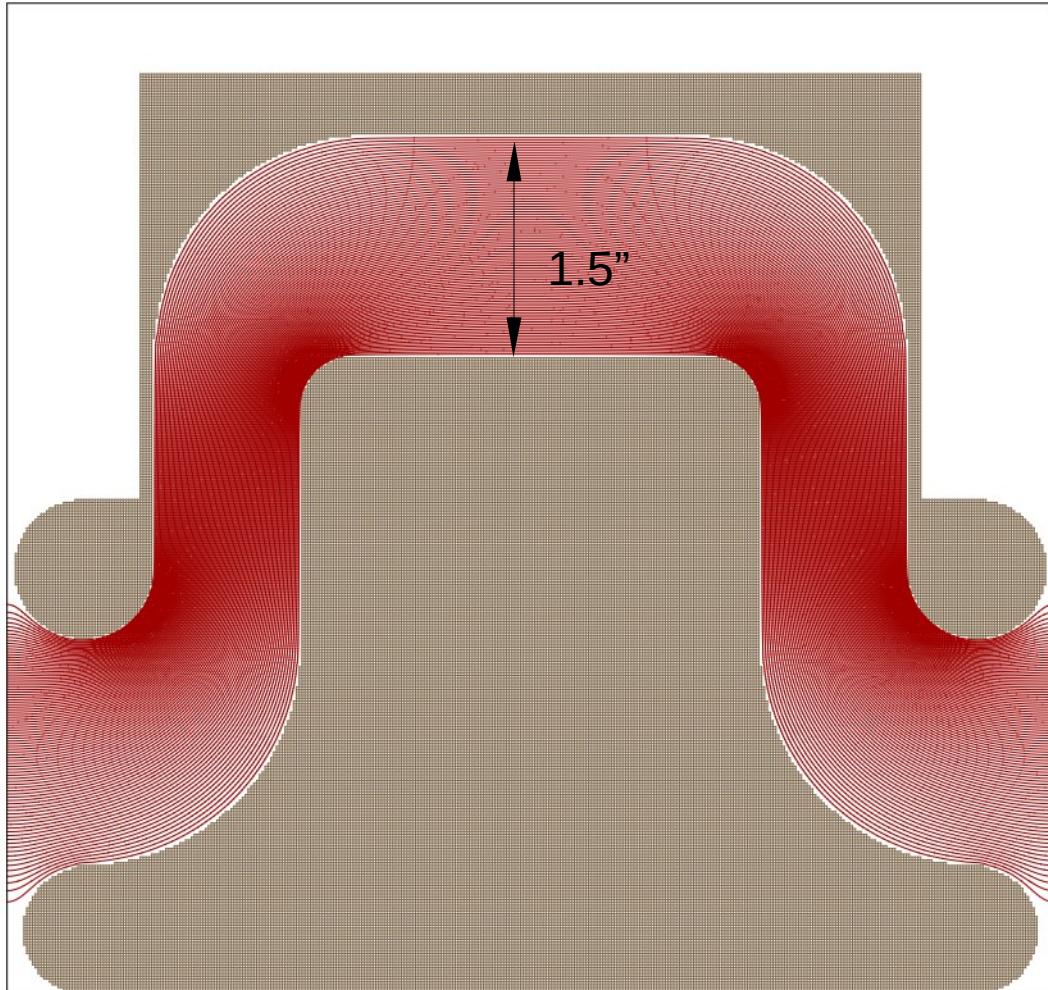
# LANL Coupler



Rescaled for our cavity

LANL design

# Simion calculation

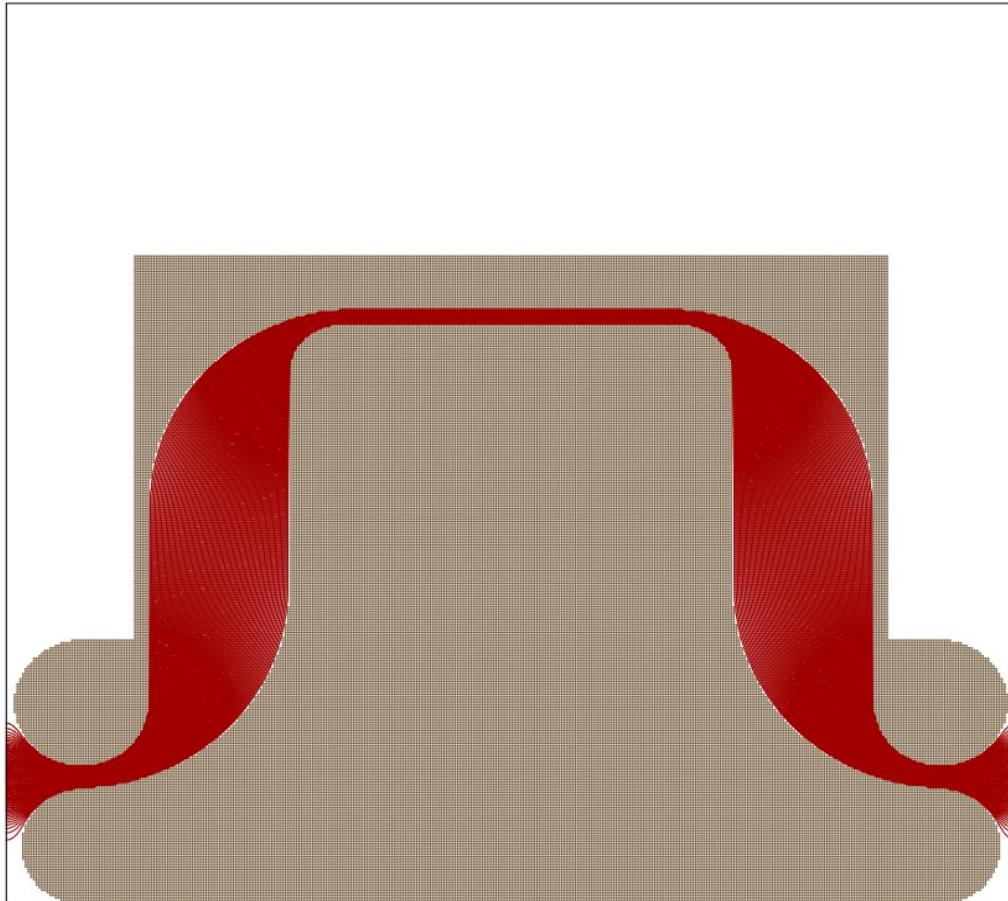


Simion calculation says LANL capacitance is 8 pF. But ssl-preprint-067.pdf claims their cap is 13 pF, so it is about 50% off. Needs to be checked with a better program. I am going to assume that SIMION is correct for now.

Using the reduced size for our cavity, the scale reduction is  $\alpha=0.5$   
And  $C \sim A/d$ , rescaling everything by  $\alpha$ ,  
 $A \rightarrow \alpha^2 A$  and  $d \rightarrow \alpha d$ , thus  $C \rightarrow \alpha C$ .

Thus the capacitance of the reduced size coupling capacitor is 4 pF. **This is too small for us.**

# Move cap closer



I checked two cases:

- 0.2" gap: 28 pF
- 0.1" gap: 42 pF

These are probably underestimates.

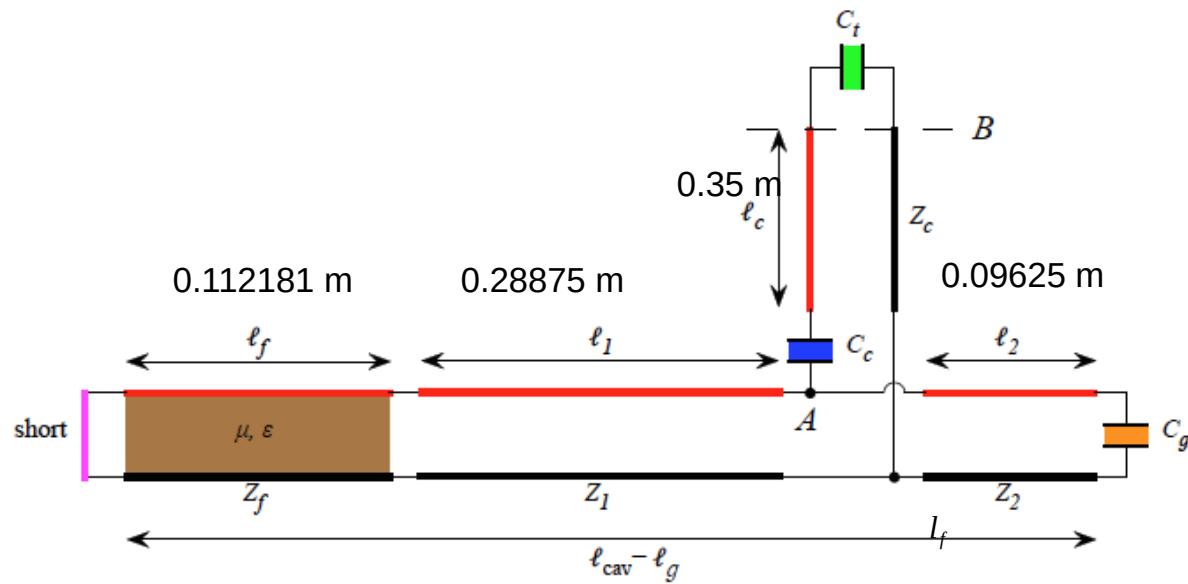
Again, for our smaller sized cap, it will be:

- 0.1" gap: 14 pF
- 0.05" gap: 21 pF

Therefore, the gap for us is probably going to be between 0.1" and 0.2" to get 10 pF, if we leave the gaps on the sides as is.

0.1" – 0.2" is probably too small a gap to hold off 100 kV.

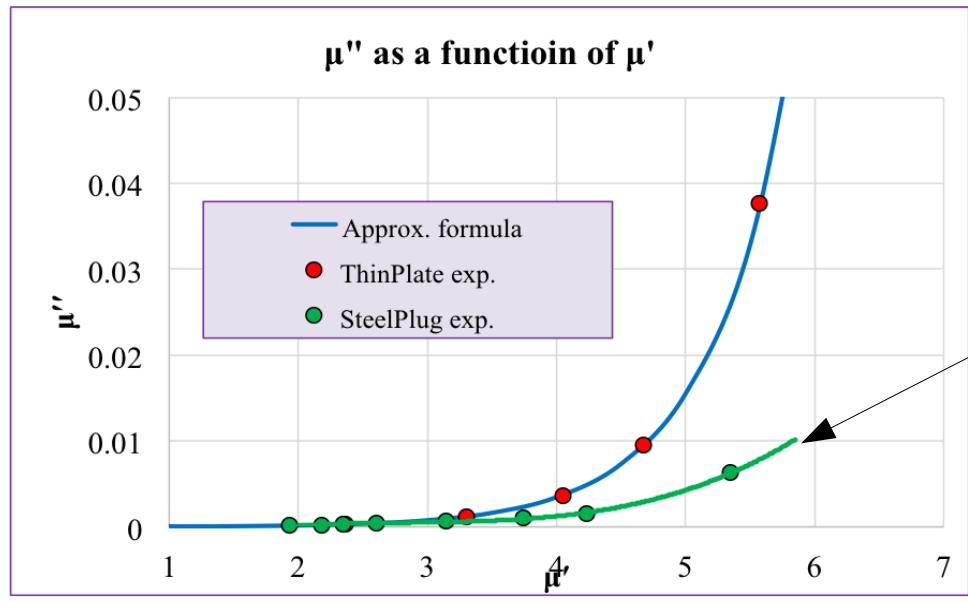
# Summary of dimensions used in calculations



| Parameter | value   | comments                |
|-----------|---------|-------------------------|
| $C_c$     | 10 pF   | Coupling capacitance    |
| $C_t$     | 60 pF   | Tube capacitance        |
| $C_g$     | 3.4 pF  | Gap capacitance         |
| $r_{fi}$  | 0.105 m | Inner radius of ferrite |
| $r_{fo}$  | 0.170 m | Outer radius of ferrite |

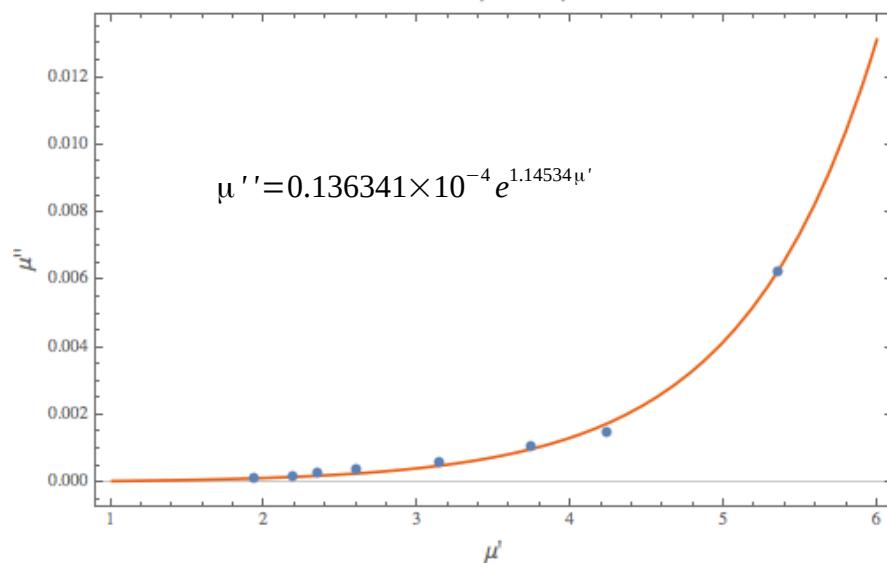
| Parameter  | value            | comments                             |
|------------|------------------|--------------------------------------|
| $Z_1, Z_2$ | $61.299 \Omega$  | Non-ferrite characteristic impedance |
| $Z_c$      | $20.9269 \Omega$ | Characteristic impedance of coupler  |
|            |                  |                                      |

# The new measurement of $\mu''$ vs $\mu'$ used in calculation

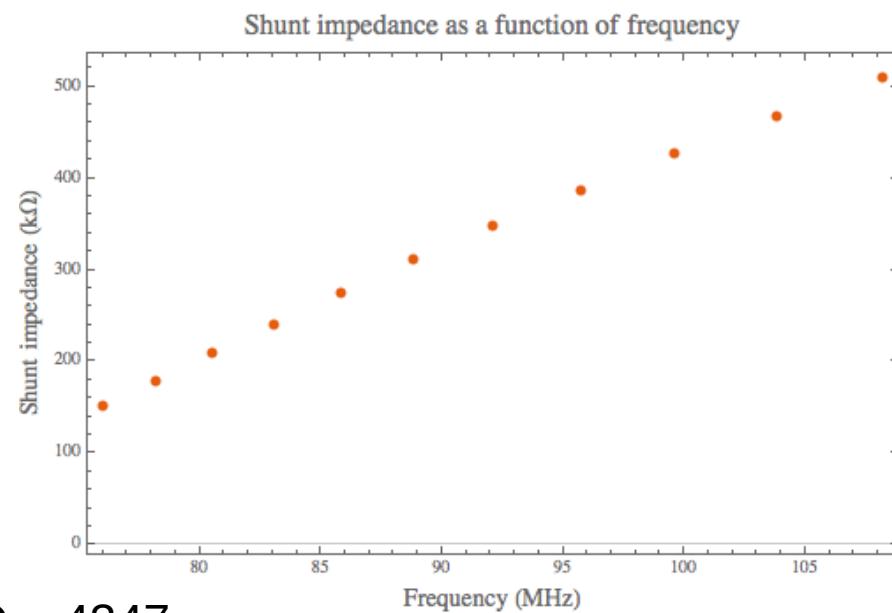
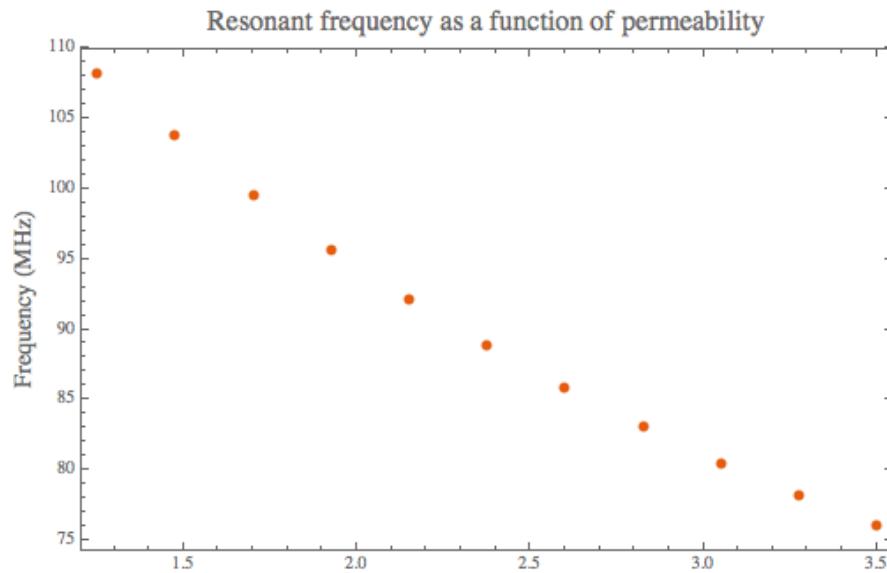


The data is from 21 May 2015 talk.

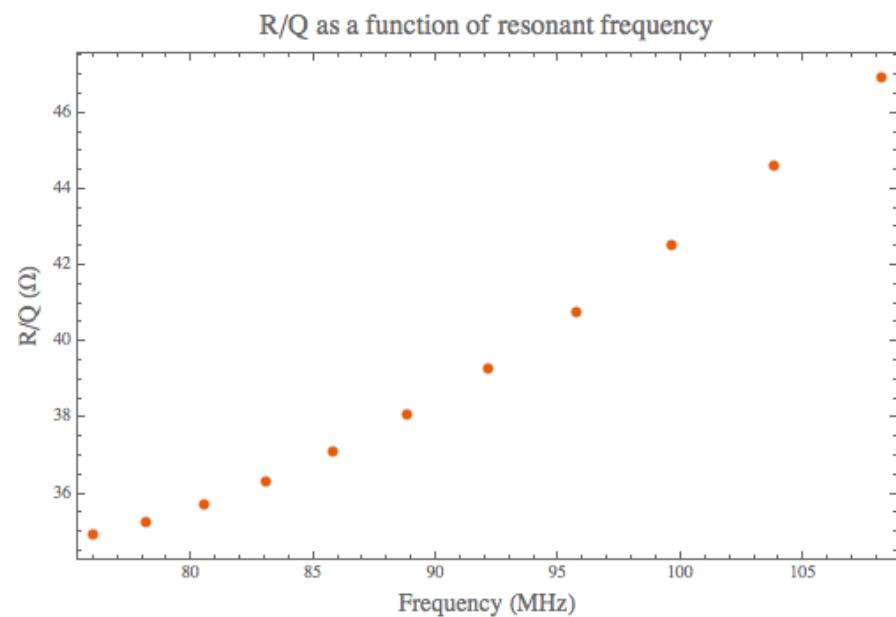
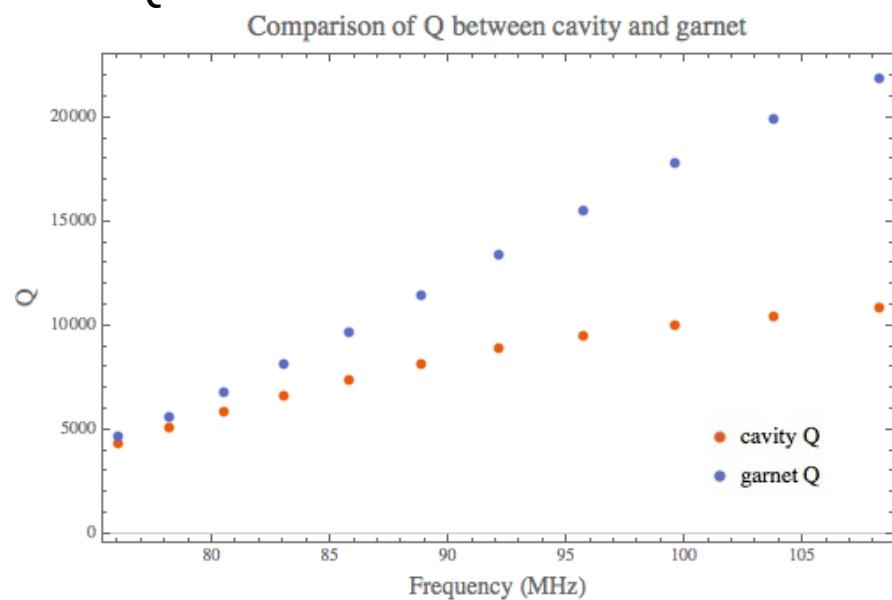
Steel Plug experiment data is used in Mathematica calculation.



# The following are results from Mathematica

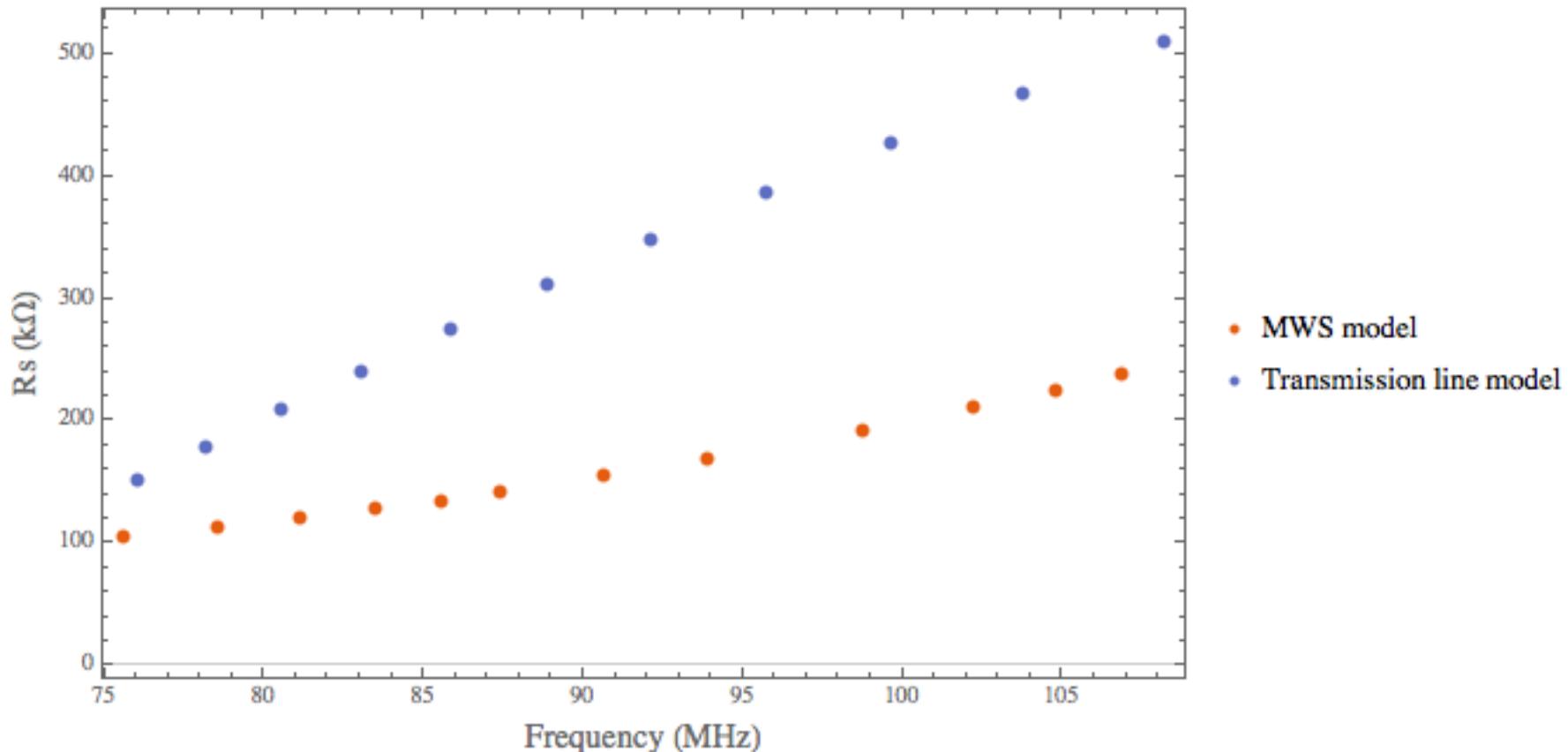


Shunt impedance at  $76 \text{ MHz} = 151 \text{ k}\Omega$ ,  $Q = 4347$   
 $R/Q$  at  $76 \text{ MHz} = 35 \Omega$



# Rs comparison

Comparing Transmission line Rs to MWS Rs



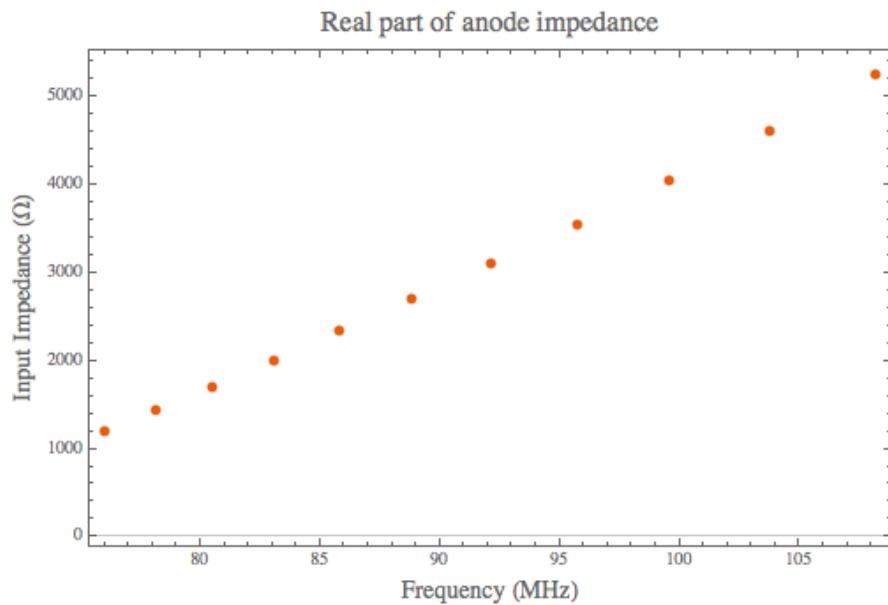
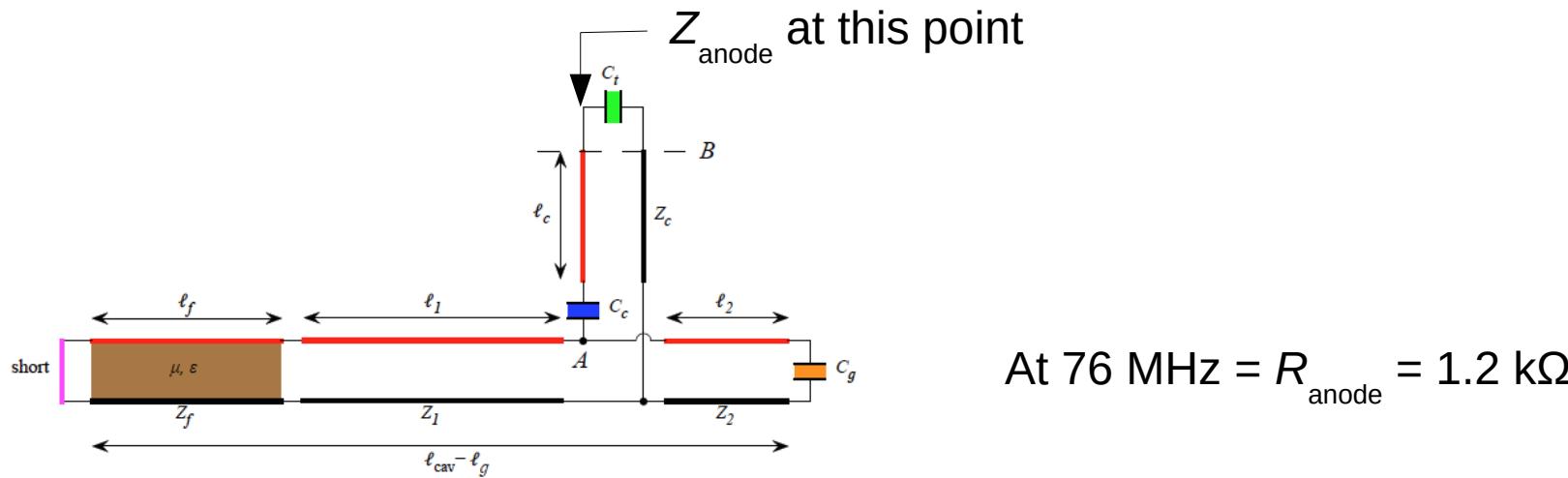
Note: I divided Gennady's results by 2 because RF group's definition is always from rms, i.e.

$$\begin{aligned} P_{\text{rms}} &= V_{\text{rms}}^2 / R_s \\ \Rightarrow R_s &= V_{\text{rms}}^2 / P_{\text{rms}} = V_{\text{peak}}^2 / 2P_{\text{rms}} = V_{\text{eff}}^2 / 2P_{\text{thermal}} = R_{\text{gennady}} / 2 \end{aligned}$$

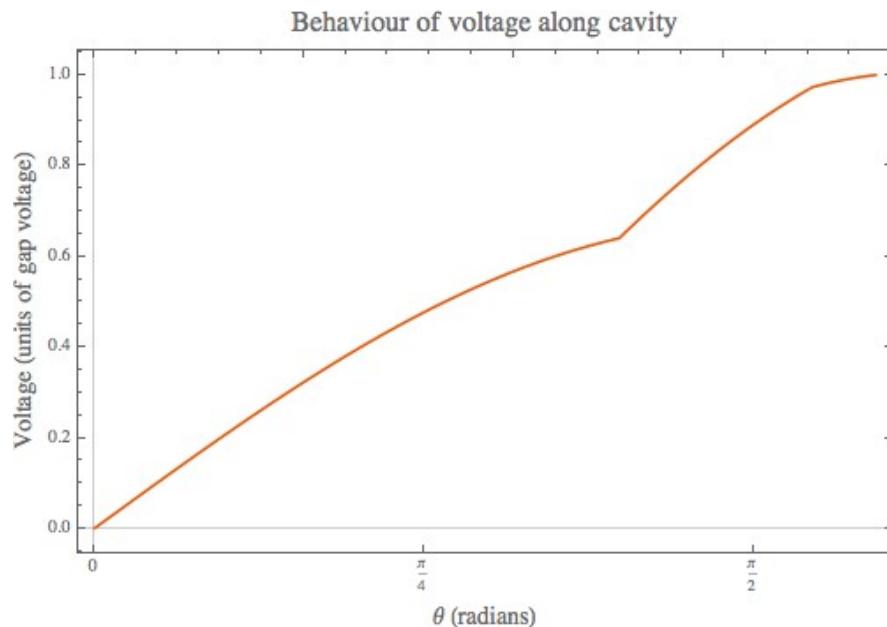
# Interesting points

- Transmission line (TL) model predicts  $R_s$  has a much higher Q at higher frequencies. Similar  $R_s$  at 76 MHz  $\sim 100 \text{ k}\Omega$ 
  - For high frequencies, does this mean that losses are much higher from non-uniform B-field in MWS than uniform B-field in TL model?
- I'm not sure this makes sense:
  - MWS has lower losses at low frequency compared to TL model and yet its  $R_s$  is smaller than TL? Note:  $R_s$  is “poportional” to Q. Recall  $R/Q = \omega_0 L$ .

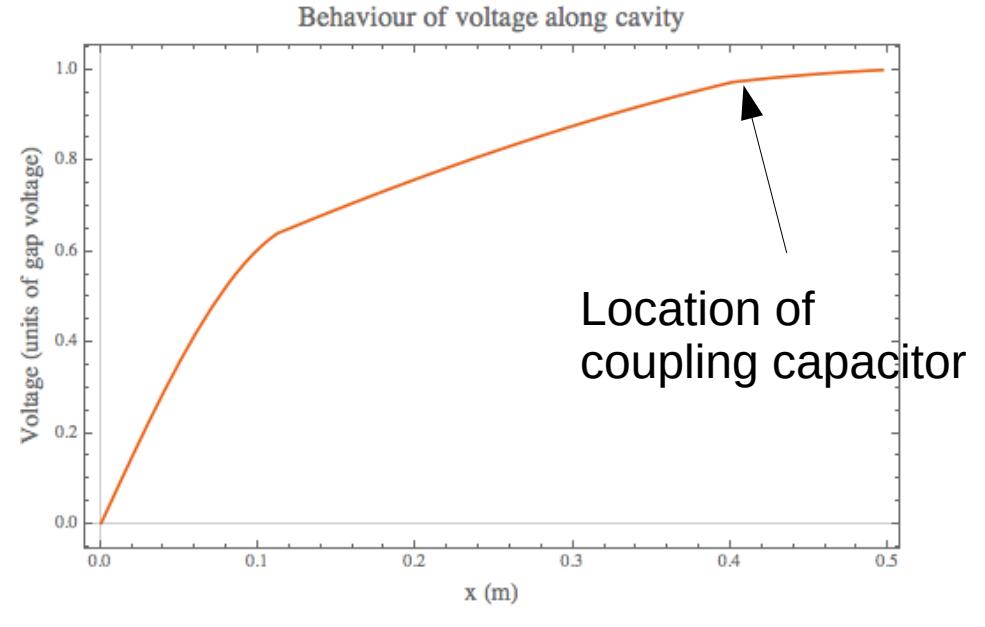
# Anode impedance



# Voltage along the cavity



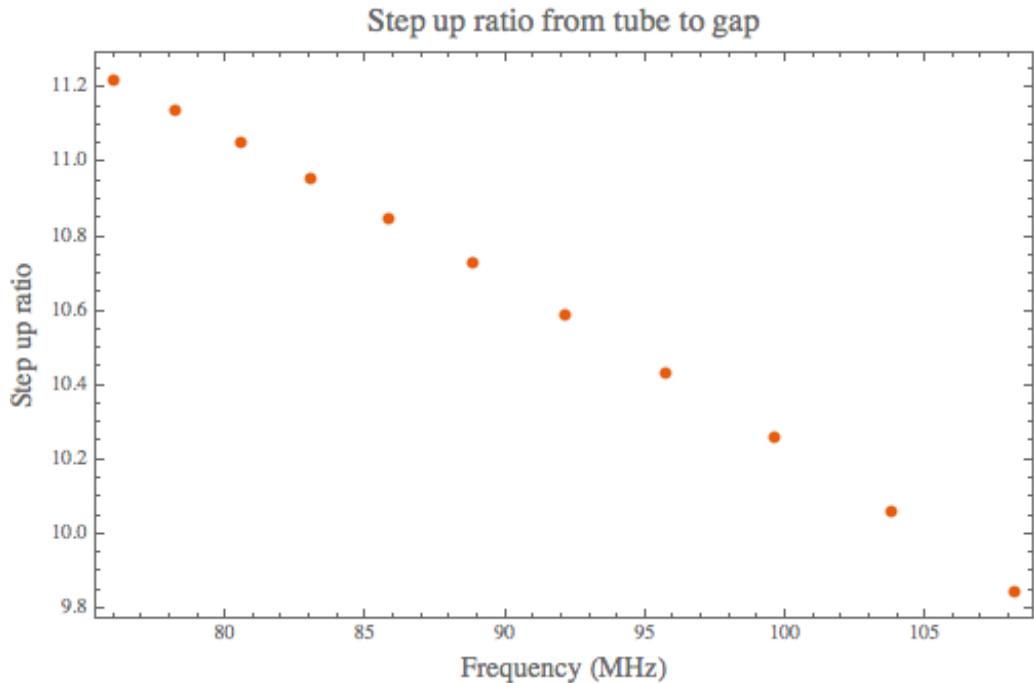
In terms of phase



In terms of position

Location of  
coupling capacitor

# Step up ratio



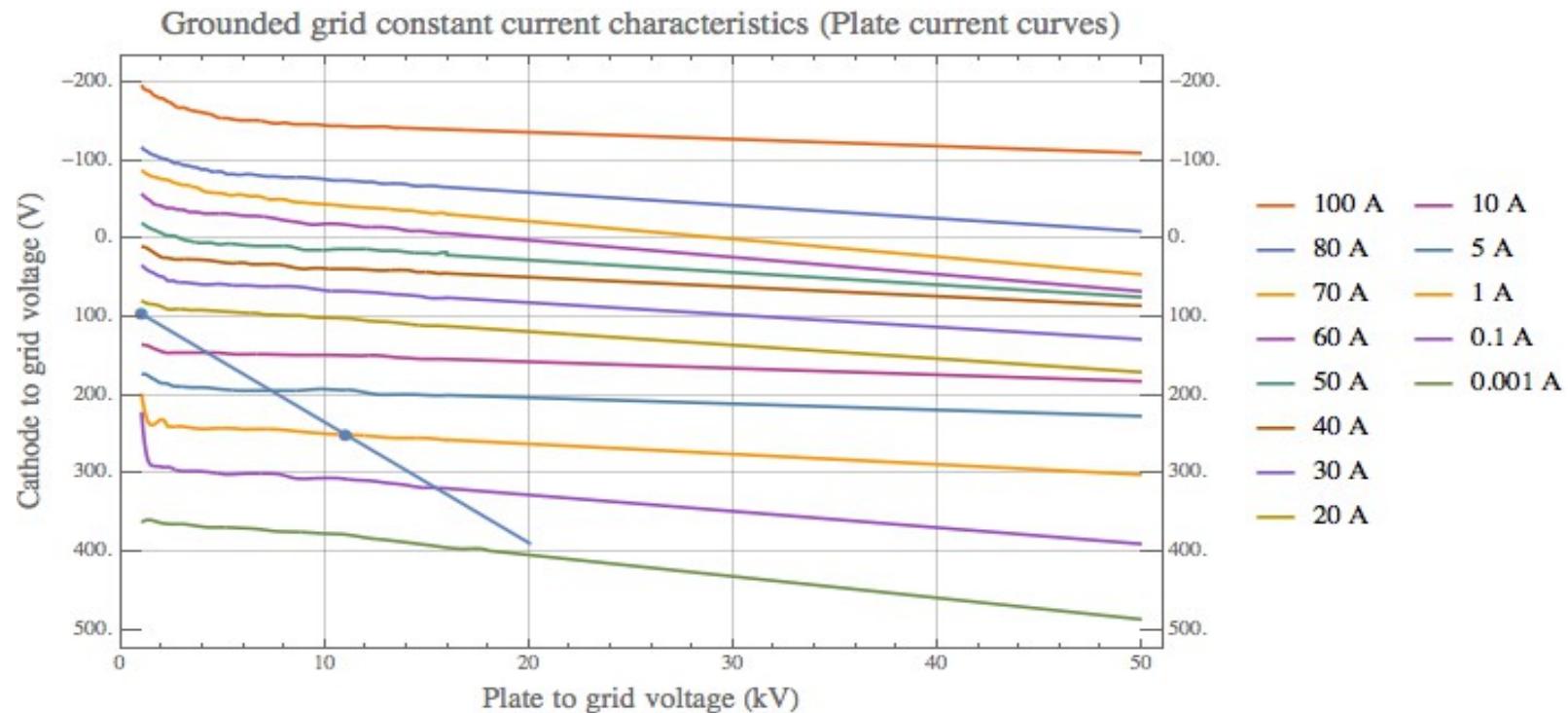
This is highly dependent on the coupling capacitance that we have chosen. Can be changed later.

Assume step up is 10 for now.

SSC etc. using step up of 8.

This choice affects the load line and thus the DC power losses.

# Y567 Grounded Grid Load line



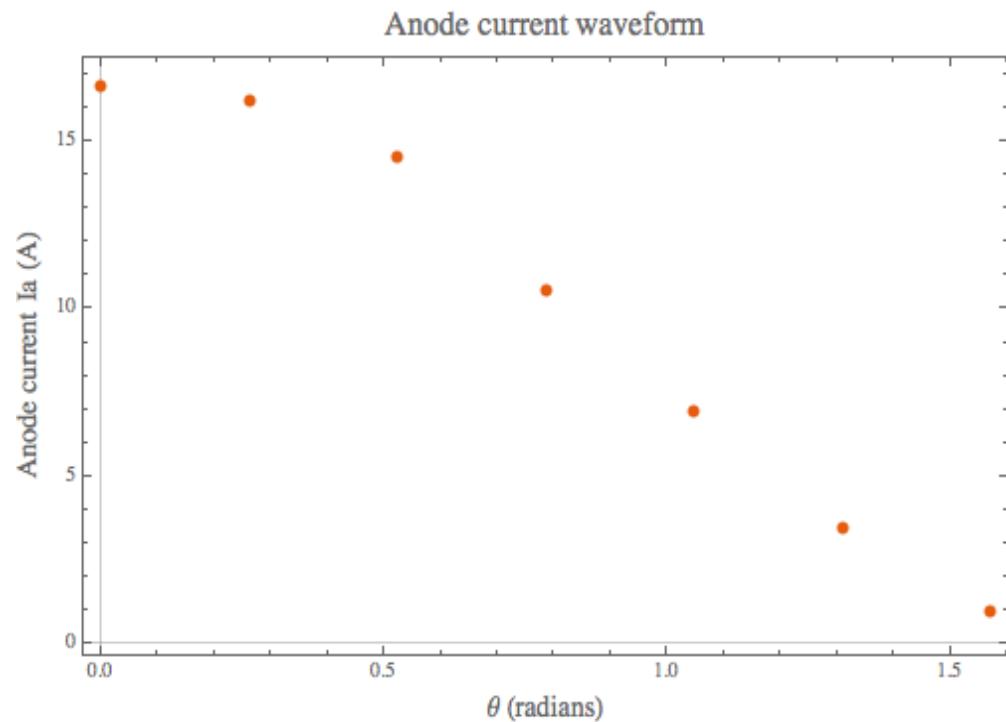
The two points that were chosen are:

Right hand point (Vanode = 11 kV – assuming stepup of 10, 1 kV screen voltage, and Vcathode = 250 V cathode to grid bias)

Left hand point (Vanode = 1.025 kV – screen to grid bias,  
I peak =  $4 * I_0 = 15.97$  A)

$I_0$  = takes into account the inefficiency of class B operation,  $V_{gap} = 100$  kV and  $R_{shunt} = 151.8$  k $\Omega$ .

# Fourier components



$I_{dc} = 4.88 \text{ A}$ , and thus DC input power  
 $V_{anode} = 11 \text{ kV} \Rightarrow P_{dc} = 54 \text{ kW}$ .

$I@76 \text{ MHz} = 7.73\text{A}$  gives  $P_{rf} = 38 \text{ kW}$

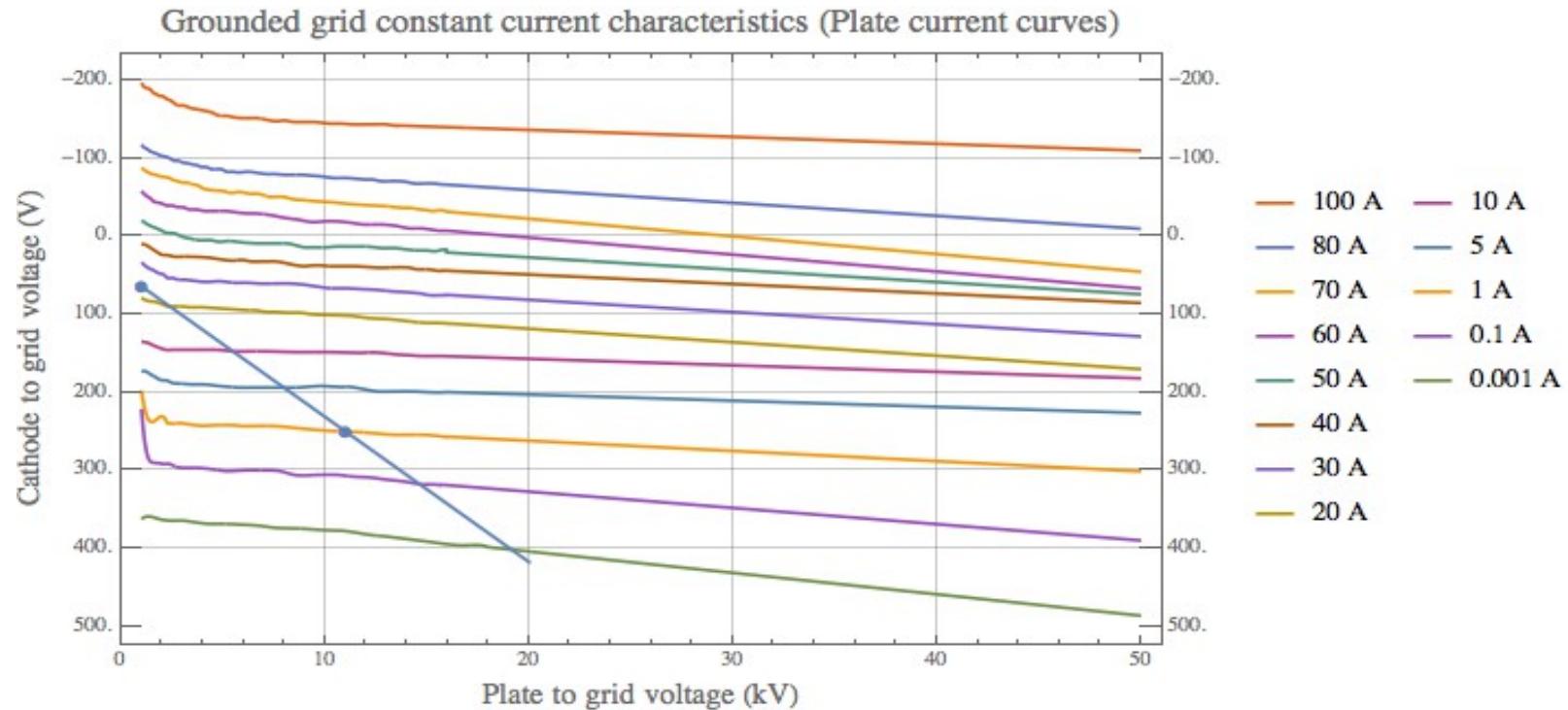
Efficiency = 0.72

The load resistance (which I assume is the anode resistance) =  $1.3 \text{ k}\Omega$

This number is consistent with the anode resistance calculated earlier =  $1.2 \text{ k}\Omega$

Latest numbers of anode resistance of  $1.2 \text{ k}\Omega$  works very well for the Y567 where we want the number to be about  $1.5 \text{ k}\Omega$ .

# Y567 Grounded Grid Load line (Gennady's Rs=105.9 kΩ)



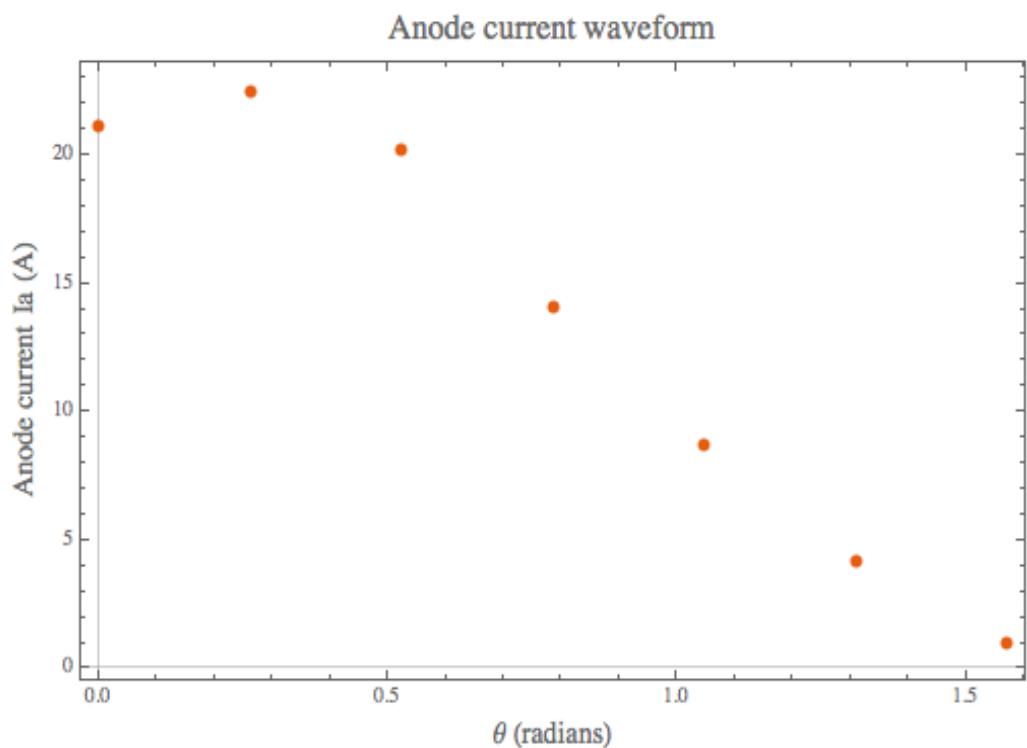
The two points that were chosen are:

Right hand point (Vanode = 11 kV – assuming stepup of 10, 1 kV screen voltage, and Vcathode = 250 V cathode to grid bias)

Left hand point (Vanode = 1.025 kV – screen to grid bias,  
I peak = 4 \* I<sub>0</sub> = 22.89 A)

I<sub>0</sub> = takes into account the inefficiency of class B operation, V<sub>gap</sub> = 100 kV and R<sub>shunt</sub> = 105.9 kΩ.

# Fourier components (MWS numbers)



$I_{dc} = 6.5 \text{ A}$ , and thus DC input power  $V_{anode} = 11 \text{ kV} \Rightarrow P_{dc} = 71.4 \text{ kW}$ .

$I@76 \text{ MHz} = 10.4 \text{ A}$  gives  $P_{rf} = 52 \text{ kW}$

Efficiency = 0.73

The load resistance (which I assume is the anode resistance) =  $0.96 \text{ k}\Omega$

Don't know anode resistance yet from MWS.  
Need to put in coupler to get value.

$0.96 \text{ k}\Omega$  works for the Y567.

# Todo/Summary

- Is 10 pF a realistic choice for the coupling capacitor?
  - Impacts design of the coupling capacitor. Simple rescaling of LANL design gives 4 pF. ( $\leftarrow$  This number needs to be checked with some other program)
  - Looks like coupling capacitor will need to hold off about 100 kV. I think this is doable. SSC cavity gap voltage is 127 kV. (PAC1993\_0753.PDF)